

# Electromagnetic energy and antennas

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SSF-workshop, Lund University, January 14, 2015

## Acknowledgments

- The Swedish Research Council.
- Swedish Foundation for Strategic Research (SSF).

Collaboration with:

- Marius Cismasu, Lund University
- Doruk Tayli, Lund University
- Sven Nordebo, Linnæus University
- Lars Jonsson, KTH



Swedish Foundation for Strategic Research

















## Optimization of antenna currents: examples

#### Gain over Q

 $\begin{array}{ll} \mbox{minimize} & \mbox{Stored energy} \\ \mbox{subject to} & \mbox{Radiation intensity} = P_0 \end{array}$ 

#### **Q** for superdirectivity $D \ge D_0$ .

minimize Stored energy

subject to Radiation intensity  $= D_0 P_{\rm rad}/(4\pi)$ Radiated power  $< P_{\rm rad}$ 

#### **Embedded structures**

minimize Stored energy

subject to Radiation intensity =  $P_0$ 

Correct induced currents

#### Need to:

- 1. Express the stored energy in the current density J.
- 2. Solve the optimization problems.





# What is (stored) EM energy?



- Time average energy density  $\epsilon_0 |\boldsymbol{E}|^2/4$  and  $\mu_0 |\boldsymbol{H}|^2/4$ .
- What is stored and radiated?
- How can we express the (stored) energy in the current density?
- ▶ Here, currents in free space.



$$W^{(E)} = \frac{C|V|^2}{4} = \frac{|I|^2}{4\omega^2 C}$$

and in inductors

$$W^{(\mathrm{M})} = \frac{L|I|^2}{4}$$















# MoM for Q and $Q_{Z'}$ (I)

Use a method of moments (MoM) formulation of the electric field integral equation (EFIE). Impedance matrix  ${\bf Z}={\bf R}+j{\bf X}$ 

$$\begin{split} \frac{Z_{mn}}{\eta} &= j \int_{S} \int_{S} \left( k^{2} \boldsymbol{\psi}_{m1} \cdot \boldsymbol{\psi}_{n2} - \nabla_{1} \cdot \boldsymbol{\psi}_{m1} \nabla_{2} \cdot \boldsymbol{\psi}_{n2} \right) \frac{\mathrm{e}^{-jkr_{12}}}{4\pi k r_{12}} \, \mathrm{dS}_{1} \, \mathrm{dS}_{2} \\ \text{where } \boldsymbol{\psi}_{n1} &= \boldsymbol{\psi}_{n}(\boldsymbol{r}_{1}), \ \boldsymbol{\psi}_{n2} &= \boldsymbol{\psi}_{n}(\boldsymbol{r}_{2}), \ n = 1, \dots, N, \text{ and} \\ r_{12} &= |\boldsymbol{r}_{1} - \boldsymbol{r}_{2}|. \\ \text{The current density is } \boldsymbol{J}(\boldsymbol{r}) &= \sum_{n=1}^{N} I_{n} \boldsymbol{\psi}_{n}(\boldsymbol{r}) \text{ with the expansion} \\ \text{coefficients determined from} \end{split}$$

$$\mathbf{ZI} = \mathbf{V}$$
 or  $\mathbf{I} = \mathbf{Z}^{-1}\mathbf{V} = \mathbf{YV}$ 

where  ${\bf V}$  is a column matrix with the excitation coefficients. The input admittance is

$$Y_{\rm in} = 1/Z_{\rm in} = \mathbf{V}^{\rm T} \mathbf{Y} \mathbf{V} / V_{\rm in}^2$$

where  $Z_{in} = R_{in} + jX_{in}$  is the input impedance.

# MoM for Q and $Q_{Z'}$ (II)

Differentiate the MoM impedance matrix

$$\frac{k \partial Z_{mn}}{\eta \partial k} = \int_{V} \int_{V} j \left( k^{2} \boldsymbol{\psi}_{m1} \cdot \boldsymbol{\psi}_{n2} + \nabla_{1} \cdot \boldsymbol{\psi}_{m1} \nabla_{2} \cdot \boldsymbol{\psi}_{n2} \right) \frac{\mathrm{e}^{-\mathrm{j}kr_{12}}}{4\pi k r_{12}} \\ + \left( k^{2} \boldsymbol{\psi}_{m1} \cdot \boldsymbol{\psi}_{n2} - \nabla_{1} \cdot \boldsymbol{\psi}_{m1} \nabla_{2} \cdot \boldsymbol{\psi}_{n2} \right) \frac{\mathrm{e}^{-\mathrm{j}kr_{12}}}{4\pi} \, \mathrm{dS}_{1} \, \mathrm{dS}_{2}$$

Differentiated input admittance (frequency independent  $\mathbf{V}~(\mathsf{MoM}))$ 

$$V_{in}^2 Y_{in}' = (\mathbf{V}^T \mathbf{Y} \mathbf{V})' = \mathbf{V}^T \mathbf{Y}' \mathbf{V} = -\mathbf{I}^T \mathbf{Z}' \mathbf{I}.$$

The stored energy determined from  $\mathbf{X}' = \operatorname{Im} \mathbf{Z}'$ 

$$W_{\mathbf{X}'}^{(\mathrm{E})} + W_{\mathbf{X}'}^{(\mathrm{M})} = \frac{1}{4} \mathbf{I}^{\mathsf{H}} \mathbf{X}' \mathbf{I}$$

is identical to the stored energy expression (free space) introduced by Vandenbosch (IEEE-TAP 2010), see • ?? and already considered by Harrington (1968,1971,1975).

#### ${\it Q}$ and ${\it Q}_{Z'}$ for free-space self-resonant antennas

Assume for simplicity a **self-resonant** antenna (circuit)

$$Q_{\mathbf{Z}'} = \frac{\omega |Z'_{\mathrm{in}}|}{2R_{\mathrm{in}}} = \frac{\omega |\mathbf{I}^{\mathrm{T}} \mathbf{Z}' \mathbf{I}|}{2\mathbf{I}^{\mathsf{H}} \mathbf{R} \mathbf{I}}$$

and using MoM with the stored energy by Vandenbosch

$$Q = \frac{2\omega \max\{W_{\mathbf{X}'}^{(\mathrm{E})}, W_{\mathbf{X}'}^{(\mathrm{M})}\}}{P_{\mathrm{d}}} = \frac{\omega \mathbf{I}^{\mathsf{H}} \mathbf{X}' \mathbf{I}}{2\mathbf{I}^{\mathsf{H}} \mathbf{R} \mathbf{I}}$$

Transpose for  $Q_{\mathbf{Z}'}$  and Hermitian transpose for Q

- ► I<sup>H</sup>X'I ≥ 0 for positive semidefinite matrices X'.
- $|\mathbf{I}^{\mathrm{T}}\mathbf{Z}'\mathbf{I}| = 0$  for some  $\mathbf{I}$  (rank > 1).

See also Capek+*etal.* IEEE-TAP 2014 for  $Q_{Z'}$  using  $\mathbf{I}^{H}$  and  $\mathbf{I}'$  and  $\mathbf{P}_{Z'}$  for stored energy and Q-factors in lumped circuits.

### Convex optimization

minimize  $f_0(\mathbf{x})$ subject to  $f_i(\mathbf{x}) \le 0, \ i = 1, ..., N_1$  $\mathbf{A}\mathbf{x} = \mathbf{b}$ 



where  $f_i(x)$  are convex, *i.e.*,  $f_i(\alpha \mathbf{x} + \beta \mathbf{y}) \leq \alpha f_i(\mathbf{x}) + \beta f_i(\mathbf{y})$  for  $\alpha, \beta \in \mathbb{R}, \alpha + \beta = 1, \alpha, \beta \geq 0.$ 

Solved with efficient standard algorithms. No risk of getting trapped in a local minimum. A problem is 'solved' if formulated as a convex optimization problem.

Antenna performance expressed in the current density J, e.g.,

- ▶ Radiated field  $F(\hat{k}) = -\hat{k} \times \hat{k} \times \int_{V} J(r) e^{jk\hat{k} \cdot r} dV$  is affine.
- Radiated power, stored electric and magnetic energies, and Ohmic losses are positive semi-definite quadratic forms in J.

## Finite ground plane with $\{6, 10, 25, 100\}\%$ antenna region



## Finite ground plane with $\{6, 10, 25, 100\}\%$ antenna region



## Why convex optimization?

Solved if formulated as a convex optimization problem.

Consider the  ${\cal G}/{\cal Q}$  problem

```
 \begin{split} \mathrm{minimize} & \max\{\mathbf{I}^{\mathsf{H}}\mathbf{X}_{\mathrm{e}}\mathbf{I},\mathbf{I}^{\mathsf{H}}\mathbf{X}_{\mathrm{m}}\mathbf{I}\} \\ \mathrm{subject to} & \mathbf{F}^{\mathsf{H}}\mathbf{I}=1 \end{split}
```

Many (optimization) algorithms can be used to solve this problem.

- ▶ Can e.g., use any of the solvers included in CVX.
  - Very simple to use.
  - ► Good for small problems but less efficient for larger problems.
- A dedicated solver for quadratic programs.
  - More efficient for larger problems.
- ▶ Random search, eg genetic algorithms (GA), particle swarms,....
  - ▶ Very inefficient. Note you do not (should not) use (GA, ...) to solve e.g., Ax = b (min.||Ax b||).
- We also use a dual formulation
  - Computational efficient for large problems.
  - Illustrates dual problems and posteriori error estimates.

### Why convex optimization: illustration

The upper bound on  $G/Q|_{\rm ub}$  is obtained by solving the dual (relaxed) problem, *i.e.*, finding the minimum of the (blue) curve

$$\left.\frac{G}{Q}\right|_{\rm ub} \leq \frac{G_\alpha}{\alpha Q_{\rm e\alpha} + (1-\alpha) Q_{\rm m\alpha}}$$

This is efficiently solved by golden section search and parabolic interpolation.



#### Why convex optimization: illustration

 $G(\hat{\boldsymbol{z}}, \hat{\boldsymbol{x}})/Q$ 0.10.05 $G_{\alpha}/Q_{\mathrm{e}\alpha}$ 0  $\alpha$ 0 0.20.40.60.81  $G(\hat{\boldsymbol{y}}, \hat{\boldsymbol{x}})/Q$ 0.10.05 $G_{\alpha}/Q_{\mathrm{e}\alpha}$  $G_{lpha}/Q_{
m mo}$ 0 0 0.20.40.60.81  $\ell/\lambda \approx 0.1$  or  $ka \approx 0.35$ 

We also compute the actual G/Q for the current  $\mathbf{I}_{\alpha}$  to get

$$\frac{G_{\alpha}}{\max\{Q_{\mathrm{e}\alpha}, Q_{\mathrm{m}\alpha}\}} \le \left.\frac{G}{Q}\right|_{\mathrm{ub}}$$

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## Why: simple optimization formulations

#### Super directivity:

minimize 
$$\max{\{\mathbf{I}^{\mathsf{H}}\mathbf{X}_{e}\mathbf{I}, \mathbf{I}^{\mathsf{H}}\mathbf{X}_{m}\mathbf{I}\}}$$
  
subject to  $\mathbf{F}^{\mathsf{H}}\mathbf{I} = 1$   
 $\mathbf{I}^{\mathsf{H}}\mathbf{R}_{r}\mathbf{I} \le 4\pi/(\eta_{0}D_{0})$ 

#### Prescribed far field:

 $\begin{array}{ll} \text{minimize} & \max\{\mathbf{I}^{\mathsf{H}}\mathbf{X}_{\mathrm{e}}\mathbf{I},\mathbf{I}^{\mathsf{H}}\mathbf{X}_{\mathrm{m}}\mathbf{I}\}\\ \text{subject to} & \int_{\Omega}|\boldsymbol{F}(\hat{\boldsymbol{k}})-\boldsymbol{F}_{0}(\hat{\boldsymbol{k}})|^{2}\,\mathrm{d}\Omega_{\hat{\boldsymbol{k}}}<\delta \end{array}$ 

#### **Embedded** antennas:

minimize 
$$\max{\{\mathbf{I}^{\mathsf{H}}\mathbf{X}_{e}\mathbf{I}, \mathbf{I}^{\mathsf{H}}\mathbf{X}_{m}\mathbf{I}\}}$$
  
subject to  $\mathbf{F}^{\mathsf{H}}\mathbf{I} = 1$   
 $\mathbf{I}_{2} = \mathbf{C}\mathbf{I}_{1}$ 



# Summary

- Stored energy expressed in the current density, fields, and input impedance.
- Can we define stored energy for a Herglotz function?
- Energy in complex media?
- Optimization of the antenna structure (global optimization) and the antenna currents (convex optimization).
- Convex optimization for bounds and optimal currents: G/Q, superdirective, embedded, ....
- Closed form solutions for small antennas.

Initial results for efficiency, more realistic geometries (phones), SAR, MIMO. Investigating dielectrics, volume currents, magnetic currents, ...

#### Slides at: http://www.eit.lth.se/staff/mats.gustafsson



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