

Sum Rules and Physical Bounds in Electromagnetics

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Outline

1 Sum rules in EM Finite objects

2 Conclusions



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Sum rules and physical bounds on passive systems General simple approach

- 1. Identify a linear and passive system.
- 2. Construct a Herglotz (or similarly a positive real) function h(z) that models the parameter of interest.
- 3. Investigate the asymptotic expansions of h(z) as $z \rightarrow 0$ and $z \rightarrow \infty$.
- Use integral identities for Herglotz functions to relate the dynamic properties to the asymptotic expansions.
- 5. Bound the integral.

Examples: Matching networks [2, 3], Radar absorbers [11], Antennas [6, 7, 4], Scattering [12, 1], High-impedance surfaces [9], Metamaterials [5], Extraordinary transmission [8], Periodic structures [10]



Integral identities for Herglotz functions

Herglotz functions with the symmetry $h(z)=-h^*(-z^*)$ (real-valued in the time domain) have asymptotic expansions ($N_0\geq 0$ and $N_\infty\geq 0$)

$$\begin{cases} h(z) = \sum_{n=0}^{N_0} a_{2n-1} z^{2n-1} + o(z^{2N_0-1}) & \text{as } z \hat{\to} 0 \\ h(z) = \sum_{n=0}^{N_\infty} b_{1-2n} z^{1-2n} + o(z^{1-2N_\infty}) & \text{as } z \hat{\to} \infty \end{cases} \xrightarrow{\text{Im}} e^{-\frac{1}{2N_0} - \frac{1}{2N_0} - \frac{1}{2N_0} - \frac{1}{2N_0} e^{-\frac{1}{2N_0} - \frac{1}{2N_0} - \frac{$$

where $\hat{\rightarrow}$ denotes limits in the Stoltz domain $0 < \theta \leq \arg(z) \leq \pi - \theta$???. They satisfy the identities $(1 - N_{\infty} \leq n \leq N_0)$

$$\lim_{\varepsilon \to 0^+} \lim_{y \to 0^+} \frac{2}{\pi} \int_{\varepsilon}^{\frac{1}{\varepsilon}} \frac{\operatorname{Im} h(x+\mathrm{i}y)}{x^{2n}} \, \mathrm{d}x = a_{2n-1} - b_{2n-1} = \begin{cases} -b_{2n-1} & n < 0\\ a_{-1} - b_{-1} & n = 0\\ a_1 - b_1 & n = 1\\ a_{2n-1} & n > 1 \end{cases}$$

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Bernland, Luger, Gustafsson, Sum rules and constraints on passive systems. J. Phys. A: Math. Theor., 2011. Mats Gustafsson, Department of Electrical and Information Technology, Lund University, Sweden (4)

Integral identities for Herglotz functions

Common cases

Known low-frequency expansion $(a_1 \ge 0)$:

$$h(z) \sim egin{cases} a_1 z & \mbox{as } z \hat{
ightarrow} 0 \ b_1 z & \mbox{as } z \hat{
ightarrow} \infty \end{cases}$$

that gives the n=1 identity (we drop the limits for simplicity)

$$\lim_{\varepsilon \to 0^+} \lim_{y \to 0^+} \frac{2}{\pi} \int_{\varepsilon}^{1/\varepsilon} \frac{\operatorname{Im} h(x + \mathrm{i}y)}{x^2} \, \mathrm{d}x \stackrel{\text{def}}{=} \frac{2}{\pi} \int_0^\infty \frac{\operatorname{Im} h(x)}{x^2} \, \mathrm{d}x = a_1 - b_1 \le a_1$$

Known high-frequency expansion (short times) $(b_{-1} \leq 0)$:

$$h(z) \sim egin{cases} a_{-1}/z & \mbox{as } z \hat{
ightarrow} 0 \ b_{-1}/z & \mbox{as } z \hat{
ightarrow} \infty \end{cases}$$

that gives the n = 0 identity

$$\frac{2}{\pi} \int_0^\infty \operatorname{Im} h(x) \, \mathrm{d}x = a_{-1} - b_{-1} \le -b_{-1}.$$

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Forward scattering sum rule



- Finite scattering object composed of a linear, passive, and time translational invariant medium.
- Incident linearly polarized plane wave.

- The propagation (wavefront) speed is limited by the speed of light.
- Optical theorem (energy conservation).
- Induced dipole moment in the static limit.
- Shadow scattering

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Forward scattering sum rule



Use the
$$n = 1$$
 identity with
 $a_1 = \gamma = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{e} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{m} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \text{ and } b_1 = 0, \text{ i.e.,}$
 $\frac{2}{\pi} \int_0^\infty \frac{\sigma_{\text{ext}}(k)}{k^2} \, \mathrm{d}k = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{e} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{m} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}})$

or written in the free-space wavelength $\lambda=2\pi/k$

$$\frac{1}{\pi^2} \int_0^\infty \sigma_{\text{ext}}(\lambda) \, \mathrm{d}\lambda = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{\text{e}} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{\text{m}} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}})$$

Propagation speed limited by the speed of light



- Causality in the sense that the scattered field cannot precede the incident field in the forward direction.
- Causal impulse response $h_t(t)$.
- ► Analytic transfer function h(k), (Fourier, Laplace transform of h_t(t), where k = 2πf/c₀ denotes the free-space wavenumber.).

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Energy conservation (passivity)



$$W_{\text{ext},\tau} = -\int_{-\infty}^{T} \int_{\partial V} \left(\boldsymbol{E}_{\text{i}}(t,\boldsymbol{r}) \times \boldsymbol{H}_{\text{s}}(t,\boldsymbol{r}) + \boldsymbol{E}_{\text{s}}(t,\boldsymbol{r}) \times \boldsymbol{H}_{\text{i}}(t,\boldsymbol{r}) \right) \cdot \hat{\boldsymbol{n}}(\boldsymbol{r}) \, \mathrm{dS} \, \mathrm{d}t$$

simplify to

$$W_{\text{ext}} = \int_{-\infty}^{T} \int_{\mathbb{R}} E(t) h_{\text{t}}(\tau - t) E(\tau) \, \mathrm{d}t \, \mathrm{d}\tau \ge 0$$

for all E implying

$$\operatorname{Im} h(k) = \sigma_{\mathrm{ext}}(k) \ge 0 \quad \text{for } \operatorname{Im} k > 0$$

cf., the optical theorem.

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Low-frequency asymptotic expansion



► $h(k) = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{e} \hat{\boldsymbol{e}} k + \mathcal{O}(k^{2})$ as $k \to 0$ (Kleinman&Senior 1986).

- Polarizability dyadic $\gamma_{
 m e}$.
- Induced dipole moment $\boldsymbol{p} = \epsilon_0 \boldsymbol{\gamma}_{\mathrm{e}} E_0 \hat{\boldsymbol{e}}.$
- ▶ Variational principles $\gamma_{\rm e} \leq \gamma_{\infty}$ (Jones 1985, Sjöberg 2009).
- High contrast polarizability dyadic γ_{∞} .

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High-frequency asymptote



Shadow scattering (Peierls 1979, Gustafsson etal 2008).

- ▶ Im $h(k) = \sigma_{\text{ext}}(k) \le 2A$ on average as $k \to \infty$. $h(k) \rightarrow 0$, *i.e.*, for $0 < \delta < \arg k < \pi \delta$.
- the extinction paradox.

Forward scattering sum rule



Use the
$$n = 1$$
 identity with
 $a_1 = \gamma = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{e} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{m} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \text{ and } b_1 = 0, \text{ i.e.,}$
 $\frac{2}{\pi} \int_0^\infty \frac{\sigma_{\text{ext}}(k)}{k^2} \, \mathrm{d}k = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{e} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{m} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}})$

or written in the free-space wavelength $\lambda=2\pi/k$

$$\frac{1}{\pi^2} \int_0^\infty \sigma_{\text{ext}}(\lambda) \, \mathrm{d}\lambda = \hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{\text{e}} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{\text{m}} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}})$$

Polarizability dyadic and induced dipole moment

The induced dipole moment can be written

$$oldsymbol{p} = \epsilon_0 oldsymbol{\gamma}_{ ext{e}} \cdot oldsymbol{E}$$

where γ_{e} is the polarizability dyadic.

Example (Dielectric sphere)

A dielectric sphere with radius a and relative permittivity $\epsilon_{\rm r}$ has the polarizability dyadic

$$\boldsymbol{\gamma}_{\mathrm{e}} = 4\pi a^{3} \frac{\epsilon_{\mathrm{r}} - 1}{\epsilon_{\mathrm{r}} + 2} \mathbf{I} \rightarrow \boldsymbol{\gamma}_{\infty} = 4\pi a^{3} \mathbf{I}$$

as $\epsilon_{\mathrm{r}}
ightarrow \infty$.

Analytic expressions for spheroids, elliptic discs, half spheres, hollow half spheres, touching spheres,...



Extinction cross sections for $a = 50 \,\mathrm{nm}$ spheres

$$\sigma_{\rm ext} = \sigma_{\rm a} + \sigma_{\rm s} = \frac{P_{\rm a} + P_{\rm s}}{|\boldsymbol{E}_{\rm i}|^2 / 2\eta_0}$$

Sum of the scattered and absorbed powers divided by the incident power flux. Integrate over the free-space wavelength $\lambda=2\pi/k$





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Forward scattering of antennas



- Forward scattering measurement of a dipole antenna.
- Loaded, short, and open circuit.
- ► Length 15 cm and 0.5 GHz to 6 GHz.

	in cm^3	loaded	short	open
sim:	γ	661	661	291
sim:	$\frac{2}{\pi} \int_0^{k_2} \frac{\sigma_{\text{ext}}(k)}{k^2} \mathrm{d}k$	644	644	265
meas:	$\frac{2}{\pi} \int_{k_1}^{k_2} \frac{\sigma_{\text{ext}}(k)}{k^2} \mathrm{d}k$	605	670	322

Forward scattering of loaded and unloaded antennas. IEEE-TAP. 2012.



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Given a geometry, *e.g.*, sphere, rectangle, spheroid, or cylinder. How does D/Q (directivity bandwidth product) depend on the geometry for optimal antennas?

$$\frac{D}{Q} \leq \frac{\eta k_0^3}{2\pi} \left(\hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{\mathrm{e}} \cdot \hat{\boldsymbol{e}} + (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \cdot \boldsymbol{\gamma}_{\mathrm{m}} \cdot (\hat{\boldsymbol{k}} \times \hat{\boldsymbol{e}}) \right)$$

based on

- Passive materials
- Antenna forward scattering
- Identities for Herglotz function

Physical limitations on antennas of arbitrary shape Proc R. Soc. A, 463. 2589-2607, 2007. Illustrations of new physical bounds on linearly polarized antennas, IEEE Trans. Antennas Propagat., 2009. Absorption Efficiency and Physical Bounds on Antennas, Int. J. of Antennas and Propagat., 946746, 2010 http://www.mathworks.com/matlabcentral/fileexchange/26806-antennag

Circumscribing rectangles



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Circumscribing rectangles



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How can we measure the polarizability?

- Change of capacitance in a parallel plate capacitor.
- The polarizability in a parallel plate waveguide.
- The periodic polarizability for symmetric objects.







Objects with increasing distance between the coins.

Large separation of charge give a large polarizability.

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High-contrast polarizability dyadics: γ_∞

 γ_∞ is determined from the induced normalized surface charge density, $\rho,$ as

$$\hat{\boldsymbol{e}} \cdot \boldsymbol{\gamma}_{\infty} \cdot \hat{\boldsymbol{e}} = \frac{1}{E_0} \int_{\partial V} \hat{\boldsymbol{e}} \cdot \boldsymbol{r} \rho(\boldsymbol{r}) \, \mathrm{dS}$$

where ρ satisfies the integral equation

$$\int_{\partial V} \frac{\rho(\boldsymbol{r}')}{4\pi |\boldsymbol{r} - \boldsymbol{r}'|} \, \mathrm{dS}' = E_0 \boldsymbol{r} \cdot \hat{\boldsymbol{e}} - V_n$$

with the constraints of zero total charge

$$\int_{\partial V_n} \rho(\boldsymbol{r}) \, \mathrm{dS} = 0$$

Can also use FEM (Laplace equation).



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Removal of metal from a square plate



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Removal of metal from a square plate and circular disk



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Sum rules for passive systems



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Sum rules for passive systems



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admittance want

large $\operatorname{Re} Y(\omega_0)$ with Y(0) = 0. forward scattering (cross section) Use identity

 $\begin{array}{l} \textbf{S-parameter} \text{ want} \\ |S(\omega_0)| \leq \delta \text{ with } |S(0)| = 1. \\ \text{absorber, matching, blockage,} \\ \text{modes, } \ldots \end{array}$

Use log+identity

admittance want

small $|Y(\omega_0)| \leq \delta$ with $Y(0) = \infty$. high impedance surface, temporal dispersion.

Use pulse+identity

S-parameter want $S(\omega_0) \approx 1$ with S(0) = -1. high impedance surface, extra ordinary transmission Use Cayley+pulse+identity

Outline

1 Sum rules in EM Finite objects





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Conclusions

- Sum rules derived from integral identities for Herglotz functions.
- Physical bounds.
- Extinction cross section, antennas, extra ordinary transmission, transmission blockage, high-impedance surfaces, radar absorbers, temporal dispersion, perfect lenses, artificial magnetism.

Why physical bounds for passive systems?

- ► Realistic expectations. Possible/impossible.
- Possible design improvements. Is it worth it?
- Figure of merit for a design.
- Use of active and/or non-linear systems.







High-impedance surface.

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Outline

1 Sum rules in EM Finite objects

2 Conclusions



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