



# STORED ELECTROMAGNETIC ENERGY

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## Stored EM energy (definitions and interpretations)

Stored electric  $W_e$  and magnetic  $W_m$  energies are instrumental in the analysis of small antennas. The stored energy is used to estimate the bandwidth [7, 22], determine physical bounds [4, 12, 21], antenna optimization [5]. Unfortunately, the stored energy is not uniquely defined and there are many different proposals in the literature [3, 10, 17, 22]:

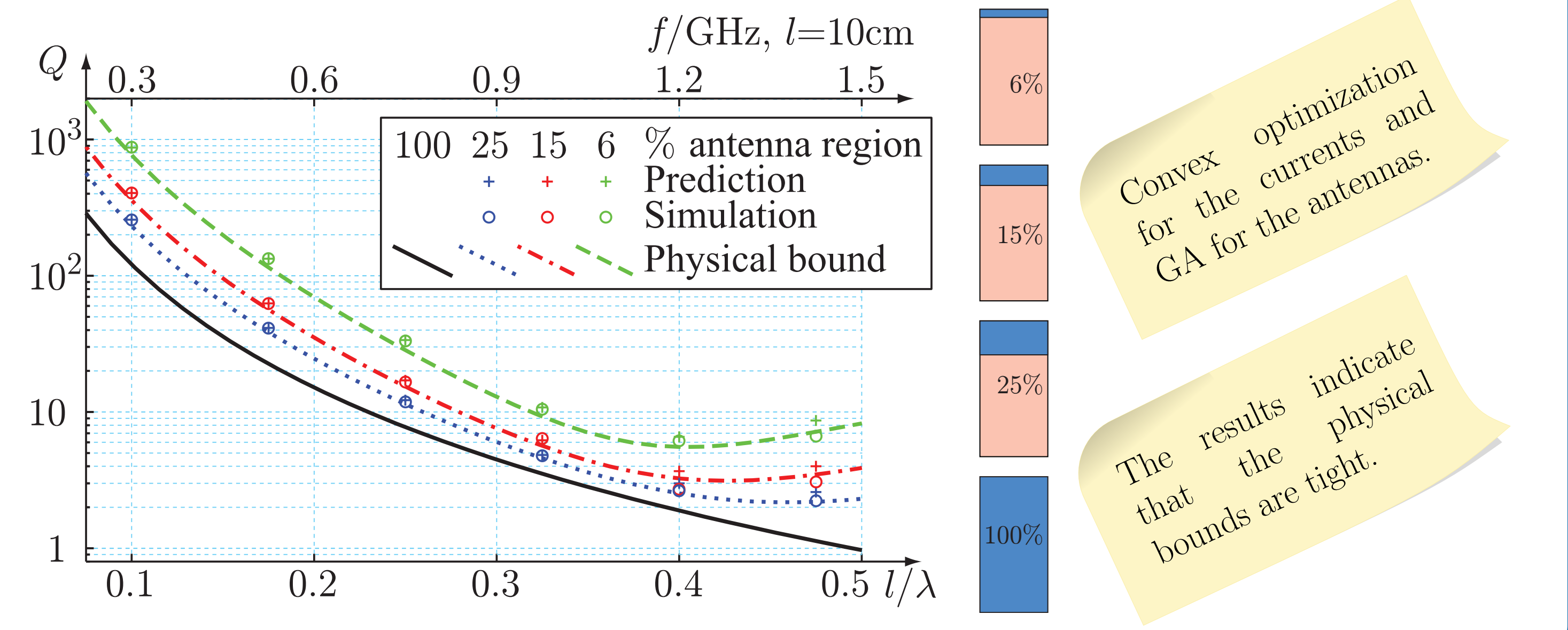
**Fields** with the difference between the energy density and the energy density of the radiated field as proposed by Collin & Rothschild 1964 [6, 7, 20, 22], see also [16]. The main problem is to define and evaluate the energy density of the radiated field and to perform the integration over  $\mathbb{R}^3$ .

**Currents** suggested by Harrington [15] based on MoM matrices, Geyi [8] for  $ka \ll 1$ , and Vandebosch [18]. More general as a state-space representations [14]. Useful for antenna current optimization [12] and an intuitive interpretation of the stored energy (in the states). Problems with the time (phase) delay.

**Input impedance** and circuit models by Chu [4] for spherical modes and in general [10]. Also local approximations using differentiation [2, 13, 22]. Well defined stored energy for a rational input impedance but needs all frequencies. Differentiation is a good approximation for single resonances.

## Physical bounds and optimization using stored energy expressed in the current density

Lower bounds on  $Q$  for a planar PEC rectangular plate with length  $\ell = 10$  cm and width  $\ell/2 = 5$  cm, where the antenna region is constrained to 100, 25, 15, 6% of the rectangle [5, 12]. Optimized antennas using single frequency optimization [5].



## Stored energy and Q-factor expressions based on fields, currents, and input impedance

### Subtraction of the far field (FF)

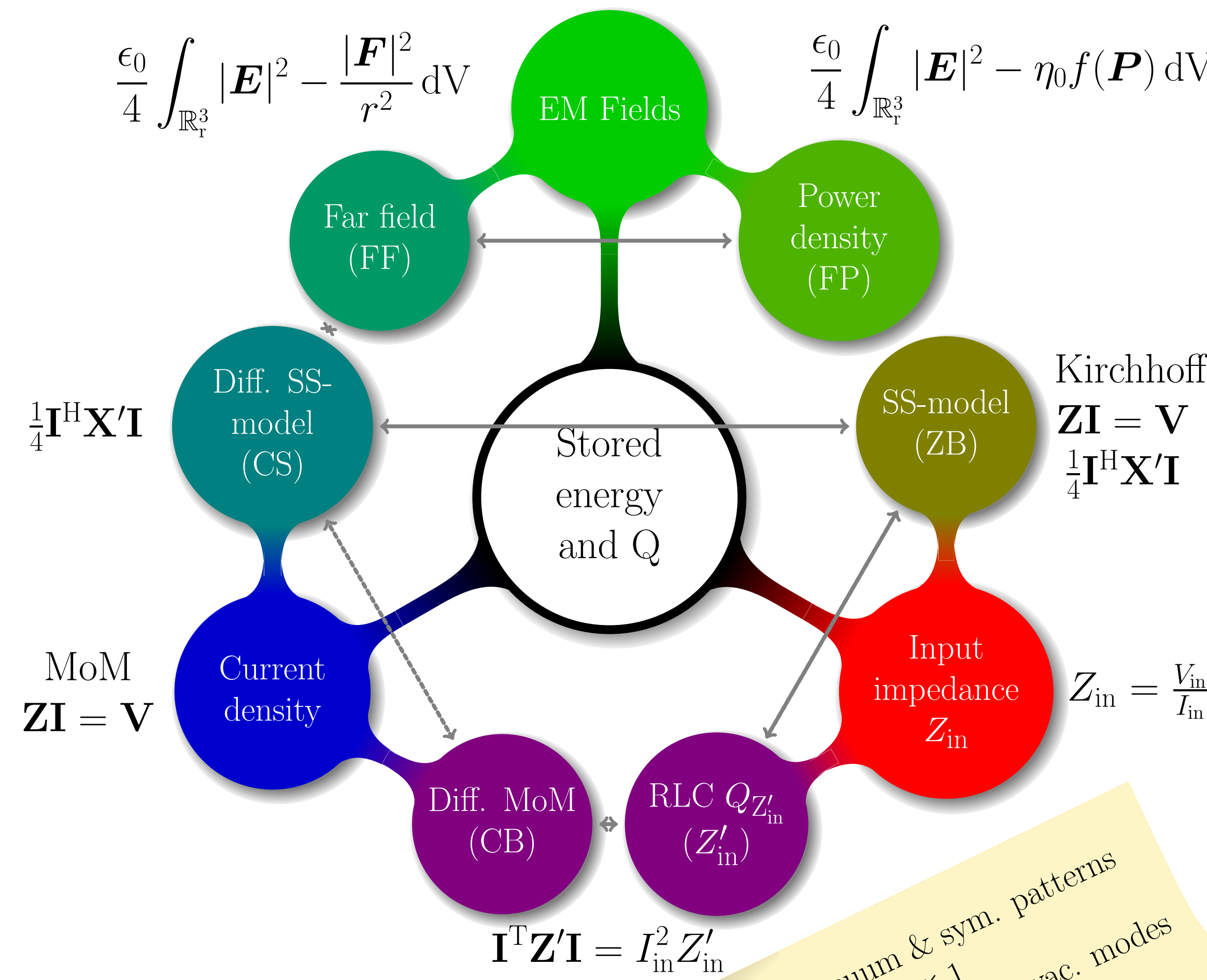
Subtraction of the energy density corresponding to the far field is common [7, 8, 22]

- Can be coordinate dependent [11, 22].
- Can be negative for large structures [10].
- Hard to generalize to lossy and dielectric media.
- Also expressed in the input reactance and far field.

### Current density and state-space models

Differentiated EFIE MoM impedance matrix (vacuum) and state-space models (dispersive media)

- Sesquilinear form in  $\mathbf{I}$  for the stored energy (CS).
  - Can be negative for large structures [10].
  - Useful for antenna current optimization [5, 12].
  - Vacuum case [15, 18] is identical to the subtracted far field (FF) for coordinate independent cases [10].
- Bilinear form in  $\mathbf{I}$  (analytic in  $s$ ) for  $Z'_{in}$  (CB).
  - Indefinite and hence  $\min Q_{Z'_{in}} = 0$  [13].
  - $Q_{Z'_{in}} \leq Q$  for small antennas [14].



### Subtraction of power density (FP)

Subtraction of the energy related to the power density (Poynting vector) [6, 11, 16]. Use  $\mathbf{P} = \text{Re}\{\mathbf{E} \times \mathbf{H}^*\}$

- $f(\mathbf{P}) = \hat{\mathbf{r}} \cdot \mathbf{P}$  coordinate dependent.  $Q$  differ with  $ka$  from the subtracted far field for spherical modes [11].
- $f(\mathbf{P}) = |\mathbf{P}|$  hard to evaluate but can be generalized to lossy and dielectric media.

### Input impedance ( $Z_B, Z'_{in}$ )

- Circuit synthesis (Brune) or state-space model (ZB)
  - Rational PR input impedance (approximation).
  - Energy stored in the constructed states [19] (or lumped circuit elements [1]).
- Differentiated input impedance [22] ( $Z'_{in}$ )
  - Easy to evaluate.
  - Padé approximation with a resonance model [13].
  - Inversely proportionality to the fractional bandwidth (\*) as  $\Gamma_0 \rightarrow 0$  if  $Q_{Z'_{in}} > 0$ .

FF=CS, vacuum & sym. patterns  
CS ≈ ZB for  $ka \ll 1$   
 $Q_{FF} = Q_{FPa} - ka$ , sph. vac. modes  
CS=CB for real-valued states

## Q-factor $Q$ and fractional bandwidth $B$

The Q-factor is defined by the stored energy and inversely proportional to  $B$ , i.e.,

$$Q = \frac{2\omega \max\{W_e, W_m\}}{P_d}, \quad B \approx \frac{2}{Q} \frac{\Gamma_0}{\sqrt{1 - \Gamma_0^2}} \quad \text{single self-resonance } Q \approx Q_{Z'_{in}} = \frac{\omega |Z'_{in}|}{2R_{in}} \quad (*)$$

## Stored energy and state-space models

The EFIE impedance matrix is

$$\mathbf{Z} = s\mu_r \mathbf{L} + \frac{1}{s\epsilon_r} \mathbf{C}_i \quad \text{and} \quad \mathbf{Z}\mathbf{I} = (s\mu_r \mathbf{L} + \frac{1}{s\epsilon_r} \mathbf{C}_i) \mathbf{I} = \mathbf{V} v_{in}$$

where the matrices  $\mathbf{L}$  and  $\mathbf{C}_i$  depend on the frequency  $s = j\omega$  and

$$L_{mn} = \int_V \int_V \psi_{m1} \cdot \psi_{n2} \frac{e^{-sn\epsilon_0 R_{12}}}{4\pi R_{12}} dV_1 dV_2, \quad C_{imn} = \int_V \nabla_1 \cdot \psi_{m1} \nabla_2 \cdot \psi_{n2} \frac{e^{-sn\epsilon_0 R_{12}}}{4\pi R_{12}} dV_1 dV_2,$$

The Lorentz model

$$\epsilon_r(s) = \epsilon_\infty + \frac{\alpha^2}{\beta^2 + \gamma s + \delta s^2}$$

gives the state-space representation [14]

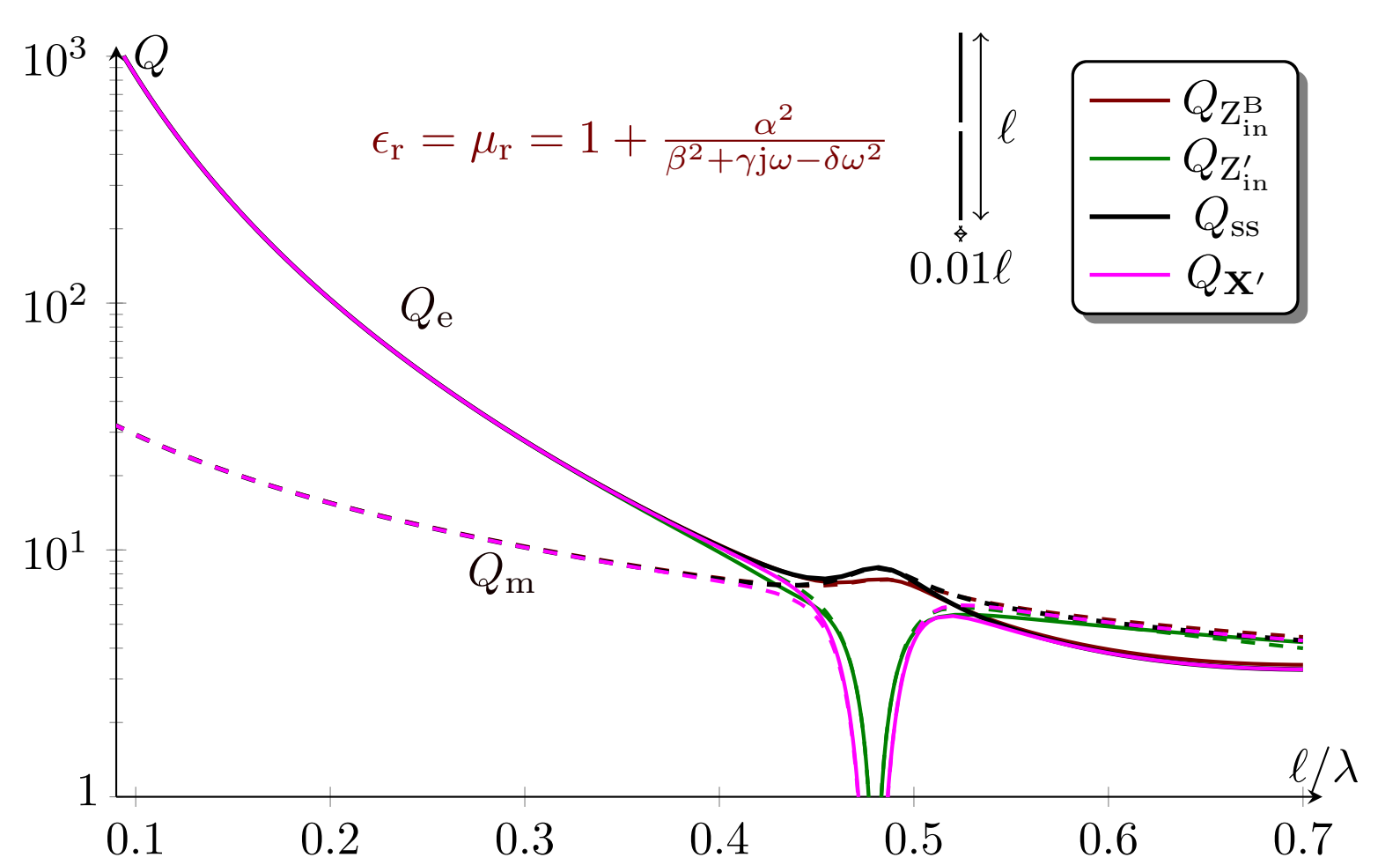
$$s \begin{pmatrix} \mu_r \mathbf{L} & \mathbf{0} & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \epsilon_\infty \mathbf{C} & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{C}_i & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{0} & \beta^2 \delta \mathbf{C}_i \end{pmatrix} \begin{pmatrix} \mathbf{I} \\ \mathbf{U} \\ \mathbf{P} \\ \dot{\mathbf{P}} \end{pmatrix} = \begin{pmatrix} \mathbf{0} & -\mathbf{1} & \mathbf{0} & \mathbf{0} \\ \mathbf{1} & \mathbf{0} & \mathbf{0} & -\alpha \mathbf{1} \\ \mathbf{0} & \mathbf{0} & \mathbf{0} & \beta^2 \mathbf{C}_i \\ \mathbf{0} & \alpha \mathbf{1} & -\beta^2 \mathbf{C}_i & -\gamma \beta^2 \mathbf{C}_i \end{pmatrix} \begin{pmatrix} \mathbf{I} \\ \mathbf{U} \\ \mathbf{P} \\ \dot{\mathbf{P}} \end{pmatrix} + \begin{pmatrix} \mathbf{V} \\ \mathbf{0} \\ \mathbf{0} \\ \mathbf{0} \end{pmatrix} v_{in}$$

with  $i_{in} = \mathbf{Y}\mathbf{I}$  and  $y_{in} = \mathbf{Y}\mathbf{I}/v_{in}$ . Reciprocal system (with internal symmetry  $\text{diag}(1, -1, -1, 1)$  if  $\mathbf{V} = \mathbf{Y}^T$ , see [19]).

## Strip dipole in $\epsilon, \mu$ Lorentz media

Stored energy expressions are compared for a strip dipole in an electric and magnetic Lorentz medium [14]. The resulting Q-factors are depicted.

- State-space results  $Q_{ss}$  (currents and polarizations) agree with the circuit synthesized values [14].
- The results from the differentiated MoM matrices vanish at the resonance wavelength  $\ell \approx 0.48\lambda$ .



Also physical bounds using current optimization with near-field constraints for antennas in lossy and temporally dispersive media [9].

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