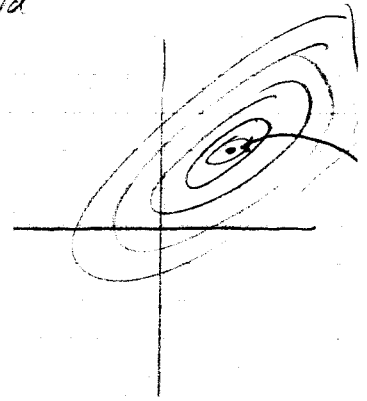
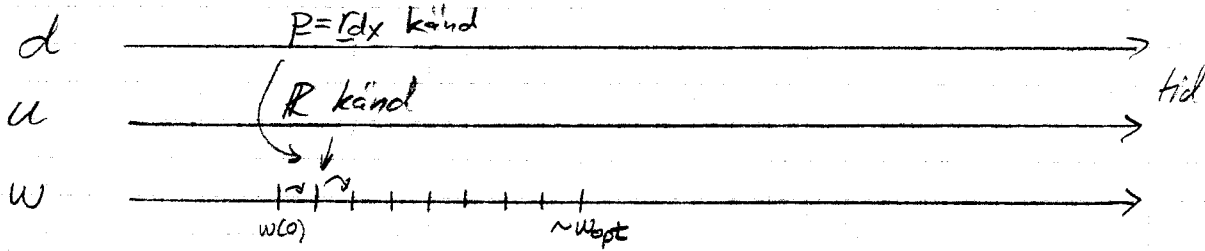
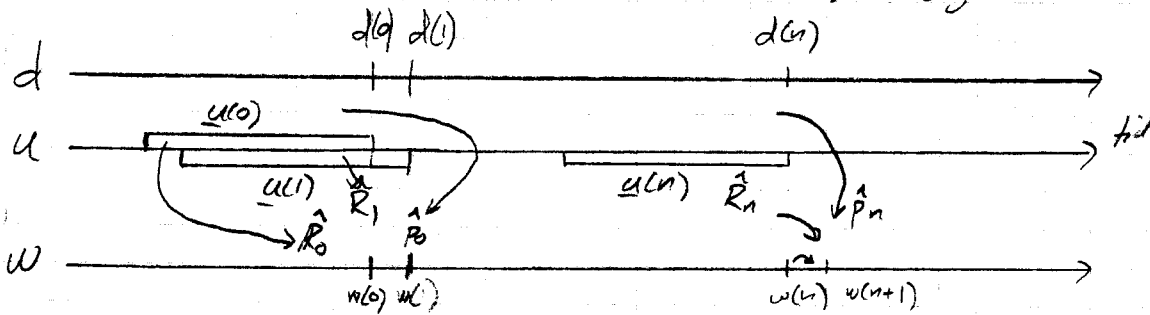


# Steepest descent metoden (WSS $\rightarrow$ $R$ finns + känd)



# LMS (adaptiv filter) (WSS $\rightarrow$ $R$ finns men skänd) Måste skapas från signaler



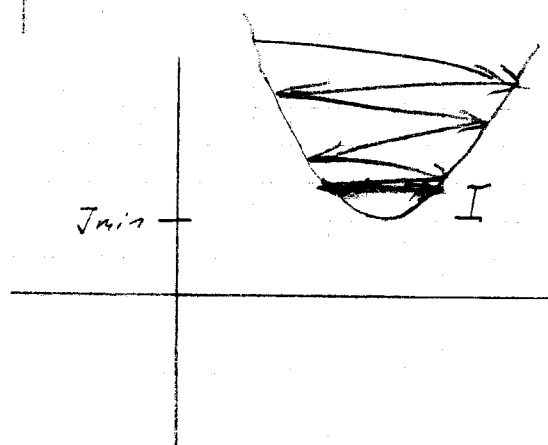
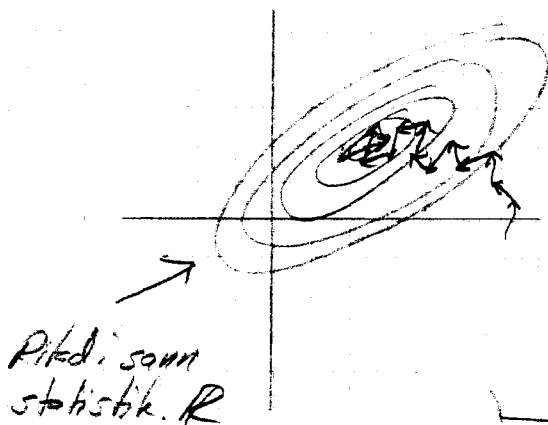
$$\hat{R}_0 = u(0)u^H(0)$$

$$\hat{R}_1 = u(1)u^H(1)$$

$$R_i \neq R_j$$

- Även om statistikerna i medel ges av  $R$  och  $p$  så behöver inte  $\hat{R}_i = R$  och  $\hat{p}_i = p$

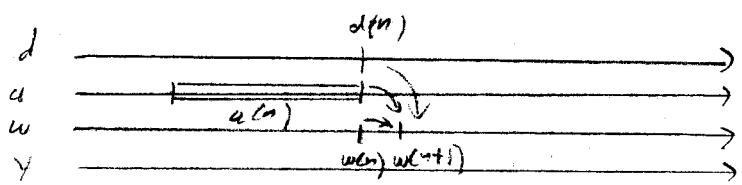
- Varje ny insignalblock har en egen uppteckning om var skälens botten är.
- Eftersom det finns en medelstatistik så är skälorna tyfset lika
- Kan anpassa sig till förändrad statistik
- Innan iterations index, men för LMS tidsindex.



Enset i  $\hat{R}_i$  och  $\hat{p}_i$  gör att metoden blir viktig när botten

# Verktyg

## Independence theory:



### Antaganden

1. Datavektorerna statistiskt oberoende (ny info i varje  $u(n)$ )  
 $u(n) \perp u(m) \quad n \neq m$
2. Datavektorerna vid tidpunkt  $n$  stat obero av önskade signalen fram till  $n-1$   
 $u(n) \perp d(m) \quad m = 0 \dots n-1$
3. Den önskade signalen vid  $n$  beror av datavektorerna fram till  $n$  men inte av den önskade signalen fram till  $n-1$   
 $d(n) \perp d(m) \quad m = 0 \dots n-1$

- Konsekvenser:
1.  $\hat{w}(n) = \hat{w}(n-1) + \mu u(n-1)(d(n-1) - w^H(n-1)u(n-1))$   
 $\Rightarrow \hat{w}(n) \perp u(n) \quad (I_a)$
  2.  $e_0(n) = d(n) - w_0^H u(n)$   $e_0(n)$  är den del av  $d(n)$  som inte kan repr. av  $u(n)$   
 $\Rightarrow e_0(n) \perp u(n) \quad (I_b)$   
 $e_0(n) \perp \hat{w}(n) \quad (I_c)$  gäller även  $e(n)$   
 $\tilde{d}(n) \quad \tilde{u}(n), \tilde{d}(n) \quad m = 0 \dots n-1$

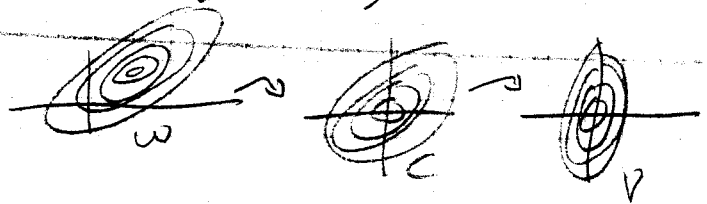
### Matristeori

$$\text{tr}(AB) = \text{tr}(BA) \quad (M_a)$$

$$E[\text{tr}A] = \text{tr}(E[A]) \quad (M_b)$$

### Konvergensanalys SD (WSS)

1. Introducera  $c(n) = w(n) - w_0$   
 $v(n) = Q^H(w(n) - w_0)$



2. Uttryck uppdelningen i  $v(n)$

$$w(n+1) = w(n) + \mu(p - R w(n))$$

$$\Rightarrow v(n) = (I - \mu R) v(n)$$

oberoende rader  
 homogen  
 kopplad till  $\lambda_i$

forts.

2 forts. Eftersom raderna är oberoende så kan insvängningar för de olika egenmoderna separeras:

$$v_k(n+1) = (1 - \mu \lambda_k) v_k(n)$$

Lösning

$$v_k(n) = (1 - \mu \lambda_k)^n v_k(0)$$

stabilitet

$$0 < \mu < \frac{2}{\lambda_{\max}}$$

3. Uttryck  $J(n)$  i  $v(n)$  för att se hur medelkvadratfelet konvergerar mot noll.

$$\begin{aligned} J(n) &= J_{\min} + \sum_k \lambda_k |v_k(n)|^2 \\ &= J_{\min} + \sum_k \lambda_k (1 - \mu \lambda_k)^{2n} |v_k(0)|^2 \end{aligned}$$

# Konvergensanalys LMS (WSS, R känd men $\lambda_i$ okänd)



1. Introducera koefficientvektorn  $\epsilon(n) = \hat{w}(n) - w_0$

2. Uttryck uppdatering i  $\epsilon(n)$

$$\hat{w}(n+1) = \hat{w}(n) + \mu u(n) e^*(n)$$

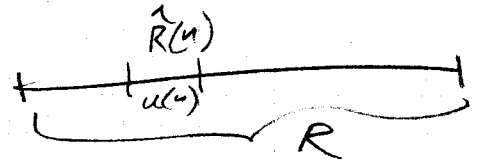
$$\epsilon(n+1) = \epsilon(n) + \mu u(n) (d^*(n) - u^H(n) \epsilon(n) - u^H(n) w_0)$$

$$\epsilon(n+1) = (I - \mu u(n) u^H(n)) \epsilon(n) + \mu u(n) e_0^*(n)$$

- ej ober rader

- ej homogen

- ej naturligt kopplad till  $\lambda_i$



Direct Averaging: Ersätt  $\hat{R}$  med  $R$ . Om  $\mu$  litet kommer dessa båda diff. ekv att bete sig lika.

$$\Rightarrow \epsilon(n+1) = (I - \mu R) \epsilon(n) + \mu u(n) e_0^*(n) \quad *$$

- kopplad till  $\lambda_i$  genom  $R$  men också till  $u(n)$

3. Uttryck  $J(n)$  i  $\epsilon(n)$

$$J(n) = E\{|e(n)|^2\} = E\{|d(n) - \hat{w}^H(n) u(n)|^2\}$$

$$= E\{|d(n) - \epsilon^H(n) u(n) - w_0^H u(n)|^2\}$$

$$= E\{|e_0(n) - \epsilon^H(n) u(n)|^2\} = [I_0, I_c]$$

$$= E\{|e_0(n)|^2\} + E\{\underbrace{\epsilon^H(n) u(n) u^H(n) \epsilon(n)}_{\text{skalär}}\}$$

$$= [M_0, M_c] = E\{|e_0(n)|^2\} + \text{tr}(E\{u(n) u^H(n) \epsilon(n) \epsilon^H(n)\})$$

$$= [I_0] = E\{|e_0(n)|^2\} + \text{tr}(\underbrace{E\{u(n) u^H(n)\}}_R E\{\epsilon(n) \epsilon^H(n)\})$$

$$= J_{\min} + \text{tr}(R K(n)) \text{ där } K(n) = E\{\epsilon(n) \epsilon^H(n)\}$$

4. Diff ekv i  $K(n)$

$$K(n+1) = E\{\epsilon(n+1) \epsilon^H(n+1)\} = [I_0, I_c] K(n) = E\{\epsilon(n) \epsilon^H(n)\}$$

$$K(n+1) = (I - \mu R) K(n) (I - \mu R) + \mu^2 R J_{\min}$$

- markstors

- rad och kolonnberoende

- inhomogen

- naturligt kopplad till  $\lambda_i$

## 5. Separera egenmoder

$K(n)$  kovariansmatris för filterfelvektorn.

$$Q^T R Q = A$$

$$Q^T K(n) Q = X(n) \quad (\text{applikera } Q \text{ på } K(n))$$

$$J(n) = J_{\min} + \text{tr}(R K(n))$$

$$= J_{\min} + \text{tr}(Q A Q^T Q X(n) Q^T)$$

$$= J_{\min} + \text{tr}(A X(n)) \quad X(n) \text{ likformig med } K(n)$$

$$= J_{\min} + \sum_i \lambda_i x_i(n)$$

där  $x_i(n)$  = diagonalelement i den likformiga kovariansmatrisen för filterfelvektorn

Diff. eku i  $X(n)$

$$\begin{aligned} X(n+1) &= Q^T K(n+1) Q = Q^T (I - \mu R) Q X(n) Q^T (I - \mu R) Q + \mu^2 Q^T R Q J_{\min} \\ &= (I - \mu A) X(n) (I - \mu A) + \mu^2 A J_{\min} \end{aligned}$$

- inhomogen

- oberoende rader

För varje diagonal element gäller eku

$$x_i(n+1) = (1 - \mu \lambda_i)^2 x_i(n) + \mu^2 \lambda_i J_{\min}$$

## Slutsats

$$J(n) = J_{\min} + \underbrace{\sum_i \lambda_i x_i(n)}_{J_{\text{ex}}(n)}$$

där  $x_i(n)$  är lösningen till

$$x_i(n+1) = (1 - \mu \lambda_i)^2 x_i(n) + \mu^2 \lambda_i J_{\min} > 0$$

## Observationer

1/ Konvergerar om  $|1 - \mu \lambda_i| < 1$   
dus  $0 < \mu < \frac{2}{\lambda_{\max}}$

2 Inhomogen: Konv inte mot noll

$$J(n) = J_{\min} + \overset{\rightarrow 0}{J_{\text{transient}}(n)} + J_{\text{ex}}(\infty) \sim \mu, \lambda_i, J_{\min}$$

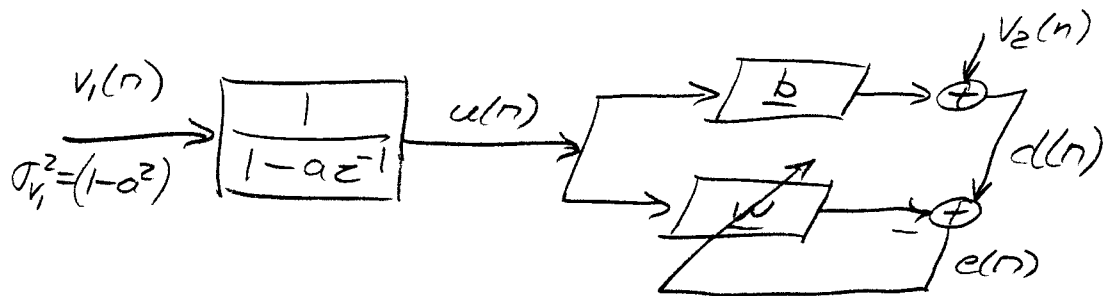
Skatta egenvärdena med hjälp av totala energin i  
insignalvektorn

$$\begin{aligned} \underline{u}^H(n) \underline{u}(n) &= \sum_{k=0}^{M-1} |u(n-k)|^2 \approx \sum_{k=0}^{M-1} \underbrace{E[|u(n-k)|^2]}_{r(0)} = M r(0) \\ &= \text{tr}(R) = \text{tr}(Q \Lambda Q^H) = \text{tr} \Lambda = \sum_{k=0}^{M-1} \lambda_k \end{aligned}$$

Byt  $0 < \nu < \frac{2}{\lambda_{\max}}$  (som gäller i medel)

mot  $0 < \nu < \frac{2}{\sum_{k=0}^{M-1} \lambda_k}$  (strävtare gräns)

# Exempel: Egenvärdesproblem



$$w_0 = R_u^{-1} p$$

$$P_u(z) = (1-a^2) \left( \frac{1}{1-az^{-1}} \right) \left( \frac{1}{1-az} \right) \Leftrightarrow r_u(k) = a^{|k|} \quad (r_u(0)=1)$$

$$\Rightarrow R_u = \begin{bmatrix} 1 & a \\ a & 1 \end{bmatrix}$$

$$p(-k) = E\{u(n-k)d(n)\} = E\{u(n-k)(u(n)*b(n) + v_2(n))\}$$

$$= E\{u(n-k)(b_0 u(n) + b_1 u(n-1) + v_2(n))\}$$

$$= b_0 r_u(-k) + b_1 r_u(-k+1) \quad k=0,1$$

$$\Rightarrow p = \begin{bmatrix} b_0 + b_1 a \\ b_0 a + b_1 \end{bmatrix} \quad w_0 = R_u^{-1} p = \frac{1}{1-a^2} \begin{bmatrix} 1 & -a \\ -a & 1 \end{bmatrix} \begin{bmatrix} b_0 + b_1 a \\ b_0 a + b_1 \end{bmatrix} = \begin{bmatrix} b_0 \\ b_1 \end{bmatrix}$$

$$J_{\min} = E\left\{ \underbrace{(d(n) - \hat{d}(n))}_{e(n)} d(n) \right\} = E\{d(n)d(n)\} - E\{d(n)\hat{d}(n)\}$$

$$= E\left\{ (b_0 u(n) + b_1 u(n-1) + v_2(n))(b_0 u(n) + b_1 u(n-1) + v_2(n)) \right\}$$

$$- E\left\{ (b_0 u(n) + b_1 u(n-1) + v_2(n))(b_0 u(n) + b_1 u(n-1)) \right\} = r_{v_2}(0) = \sigma_{v_2}^2$$

$$|\lambda I - R_u| = 0 \Rightarrow (\lambda - 1)^2 - a^2 = 0 \Rightarrow \lambda = 1 \pm |a|$$

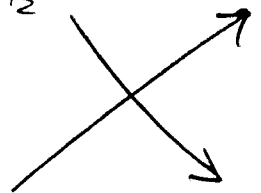
$$\kappa(R_u) = \frac{\lambda_{\max}}{\lambda_{\min}} = \frac{1+|a|}{1-|a|}$$

$$\lambda = 1+|a|: \lambda q = R_u q \Rightarrow (1+|a|) \begin{bmatrix} q_1 \\ q_2 \end{bmatrix} = \begin{bmatrix} 1 & a \\ a & 1 \end{bmatrix} \begin{bmatrix} q_1 \\ q_2 \end{bmatrix}$$

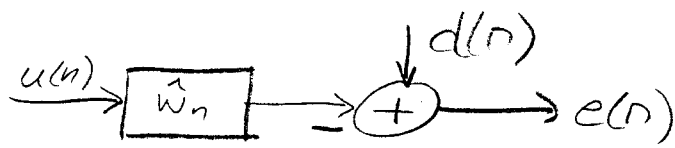
$$\begin{cases} (1+|a|)q_1 = q_1 + a q_2 \\ (1+|a|)q_2 = a q_1 + q_2 \end{cases} \Rightarrow q_1 = \frac{a}{|a|} q_2$$

$$\lambda = 1-|a|: q_1 = -\frac{a}{|a|} q_2$$

$$Q = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

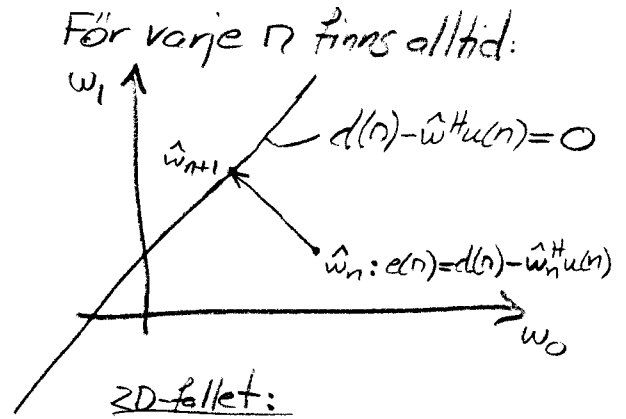


# Normalized LMS (tolkning)



Strategi: Uppdatera filterkoef så att vi uppfyller  $d(n) - \hat{w}_{n+1}^H u(n) = 0$  genom att hoppa ett så kort steg som möjligt.

Nackdel: Ingen medelvärdesbildning.  
Ej robust.



2D-fallet:

$$d(n) - w_0 u(n) - w_1 u(n-1) = 0$$

$$w_1 = \frac{d(n)}{u(n-1)} - w_0 \frac{u(n)}{u(n-1)}$$

$$\begin{aligned} J(n) &= \|\hat{w}_{n+1} - w_n\|^2 + \lambda (d(n) - \hat{w}_{n+1}^H u(n)) + \lambda^* (d(n) - u(n)^H \hat{w}_{n+1}) \\ &= \hat{w}_{n+1}^H \hat{w}_{n+1} - \hat{w}_{n+1}^H \hat{w}_n - \hat{w}_n^H \hat{w}_{n+1} + \hat{w}_n^H \hat{w}_n + \lambda (d(n) - \hat{w}_{n+1}^H u(n)) + \lambda^* (d(n) - u(n)^H \hat{w}_{n+1}) \end{aligned}$$

$$\begin{cases} \frac{\partial J(n)}{\partial \hat{w}_{n+1}^H} = (\hat{w}_{n+1} - \hat{w}_n) - \lambda u(n) = 0 & (1) \\ \frac{\partial J(n)}{\partial \lambda^*} = d(n) - u(n)^H \hat{w}_{n+1} = 0 & (2) \end{cases}$$

(1) Mult  $u(n)^H$  fr. vån.

$$\lambda = \frac{u(n)^H (\hat{w}_{n+1} - \hat{w}_n)}{\|u(n)\|^2} \stackrel{(2)}{=} \frac{d(n) - u(n)^H \hat{w}_n}{\|u(n)\|^2} = \frac{e(n)^*}{\|u(n)\|^2}$$

(1) ger

$$\hat{w}_{n+1} = \hat{w}_n + \frac{1}{\|u(n)\|^2} u(n) e(n)^*$$

Fix 1:  $\tilde{\mu}$  steglängd. Ett kortare steg ökar medelvärdesbildning

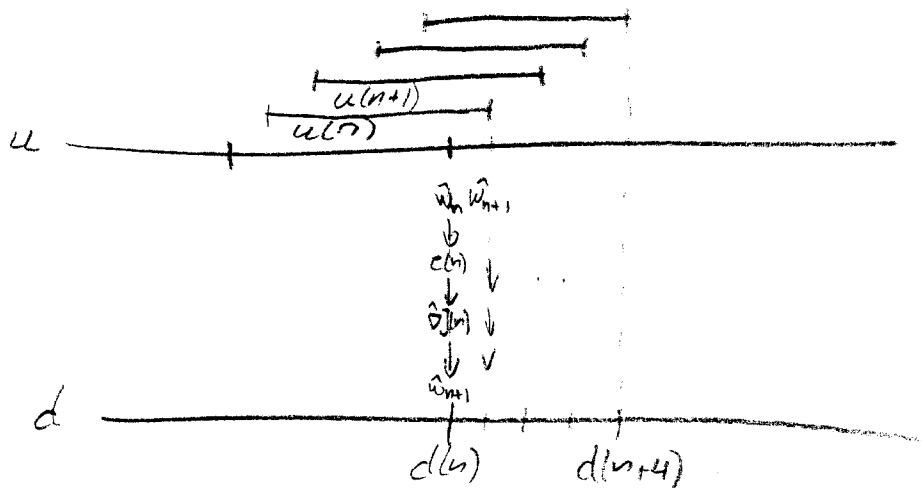
Fix 2:  $a$  om  $\|u(n)\| \rightarrow 0$

$\Rightarrow$

$$\hat{w}_{n+1} = \hat{w}_n + \frac{\tilde{\mu}}{a + \|u(n)\|^2} u(n) e(n)^*$$

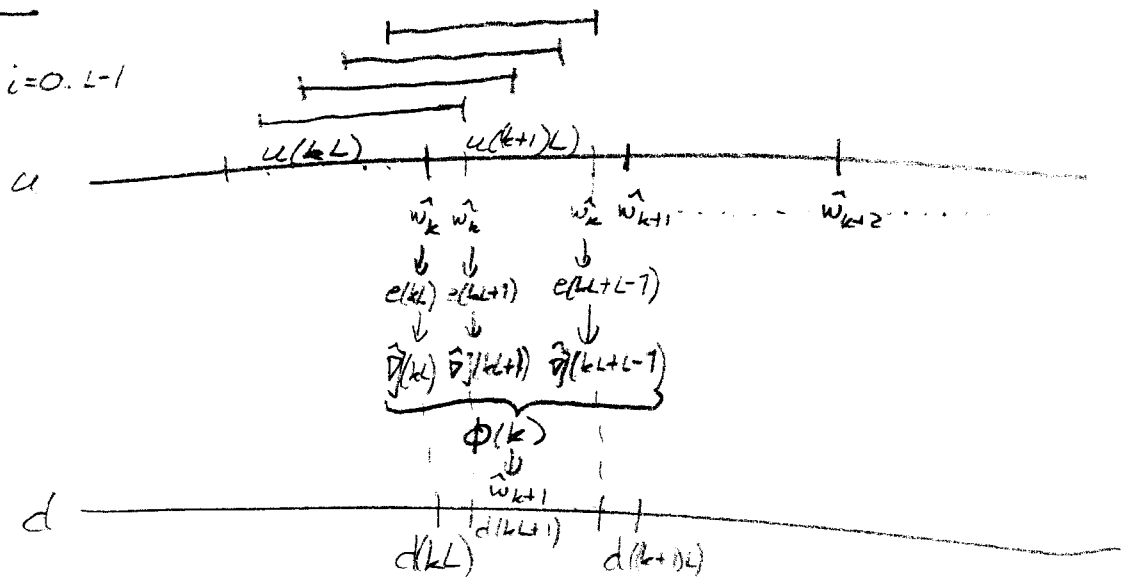


LMS

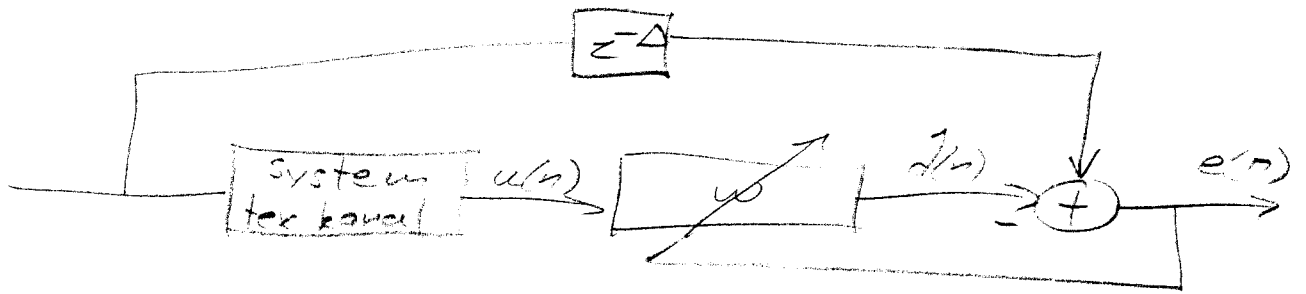


Block LMS

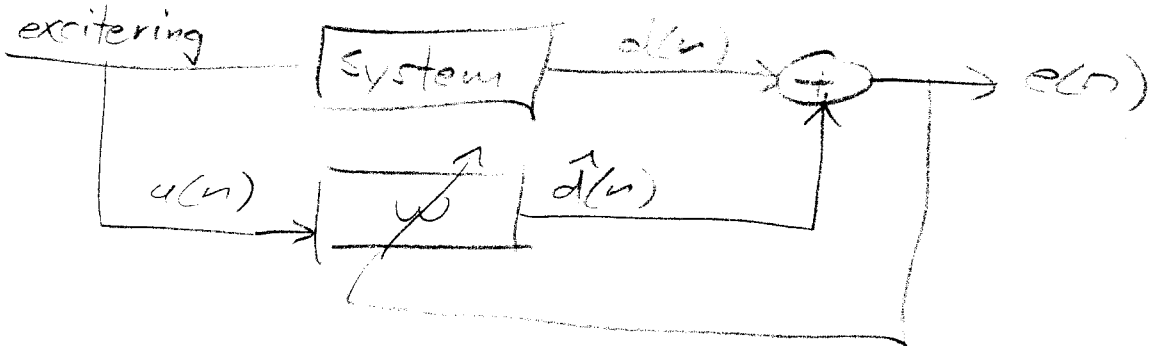
$n = kL + i, i = 0, L-1$



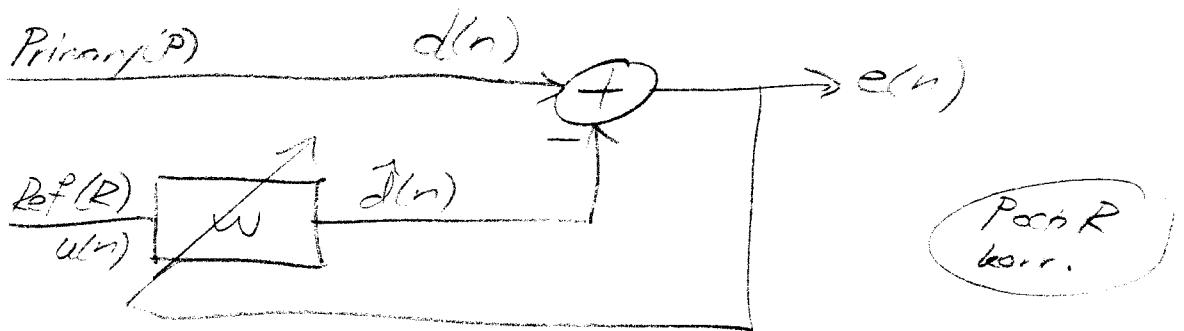
# Invers adaptiv modellering, utvärdering, avfäring



# Adaptiv modellering, systemidentifiering



# Adaptiv störningsundertryckning, kancellering



Poch R  
korr.

Ex Brusundertryckning i bil, flygplan, mobil, etc

50Hz undertryckning i EKG

Ekosläckning i tex telefon

Line enhancer

Notch filter

Högpass filter

Flera R gör också bra.

P: Tal + brus  
R: brus (korr med brus)

P: EKG inkl. 50Hz  
R: 50Hz signal från vägg

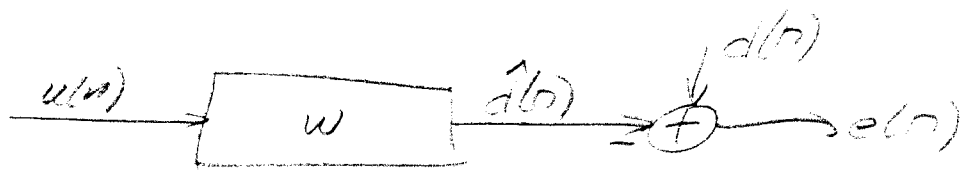
P: T2 + T1 retur  
R: T1

P: Färgat brus + periodisk sig  
R: Som P men delat  $\Delta$

P: Signal  
R:  $c \cdot \cos \omega t$

P: Signal  
R: 1

# Filtering



## Wiener filter (MS)

$$J = E\{e^2(n)\} = E\left\{\left(d(n) - \sum_{k=0}^{M-1} w(k)u(n-k)\right)^2\right\}$$

$$= E\{d^2(n)\} - 2 \sum_{k=0}^{M-1} w(k) E\{d(n)u(n-k)\} + \sum_{k_1=0}^{M-1} \sum_{k_2=0}^{M-1} w(k_1)w(k_2) E\{u(n-k_1)u(n-k_2)\}$$

$r_d(0)$                        $r_{du}(k)$                        $r_u(k_2-k_1)$

$$\frac{\partial J}{\partial w(l)} = -2 E\{d(n)u(n-l)\} + 2 \sum_{k=0}^{M-1} w(k) E\{u(n-k)u(n-l)\} = 0 \quad l=0, \dots, M-1$$

$$\Rightarrow -2 r_{du}(l) + 2 \sum_{k=0}^{M-1} w(k) r_u(l-k) = 0 \quad l=0, \dots, M-1$$

$$\sum_{k=0}^{M-1} w(k) r_u(l-k) = r_{du}(l) \quad l=0, \dots, M-1$$

$$\underline{R_u w = r_{du} \quad (\text{ott } p)}$$

där  $r_u(l-k) = E\{u(n-k)u(n-l)\}$

och  $r_{du}(l) = E\{d(n)u(n-l)\} = E\{u(n-l)d(n)\} = p(l)$

## LS-filter

$$E = \sum_{i=1}^{i_2} e^2(n) = \sum_{i=1}^{i_2} \left(d(n) - \sum_{k=0}^{M-1} w(k)u(n-k)\right)^2$$

$$= \sum_{i=1}^{i_2} d^2(n) - 2 \sum_{k=0}^{M-1} w(k) \sum_{i=1}^{i_2} d(n)u(n-k) + \sum_{k_1=0}^{M-1} \sum_{k_2=0}^{M-1} w(k_1)w(k_2) \sum_{i=1}^{i_2} u(n-k_1)u(n-k_2)$$

$\Phi_d(0)$                        $\Phi_{du}(k)$                        $\Phi_u(k_2, k_1)$

$$\frac{\partial E}{\partial w(l)} = -2 \sum_{i=1}^{i_2} d(n)u(n-l) + 2 \sum_{k=0}^{M-1} w(k) \sum_{i=1}^{i_2} u(n-k)u(n-l) = 0 \quad l=0, \dots, M-1$$

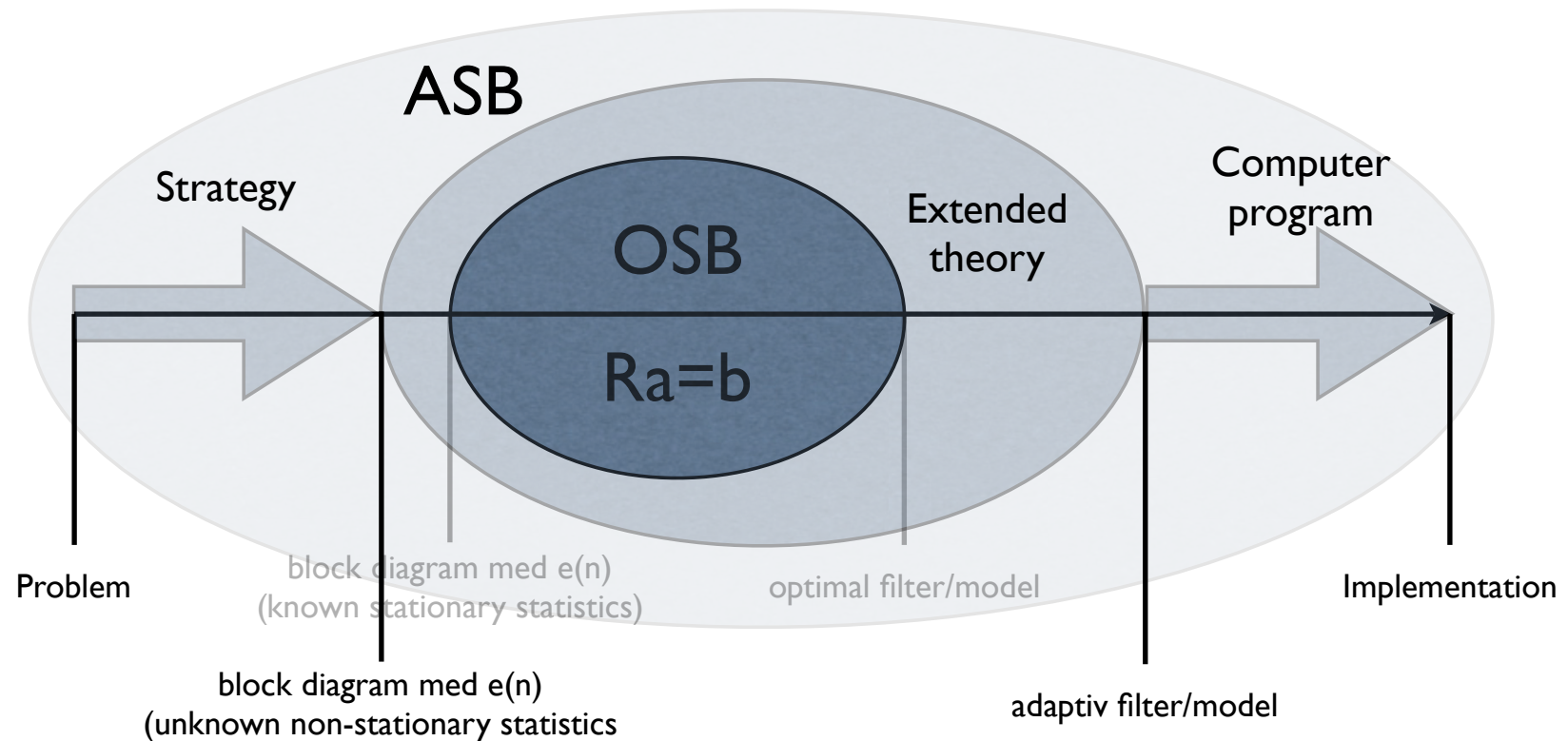
$$\Rightarrow \sum_{k=0}^{M-1} w(k) \Phi_u(l, k) = \Phi_{du}(l) \quad l=0, \dots, M-1$$

$$\underline{\Phi_u w = \Phi_{du} \quad (\text{ott } z)}$$

där  $\Phi(k, l) = \sum_{i=1}^{i_2} u(n-k)u(n-l)$

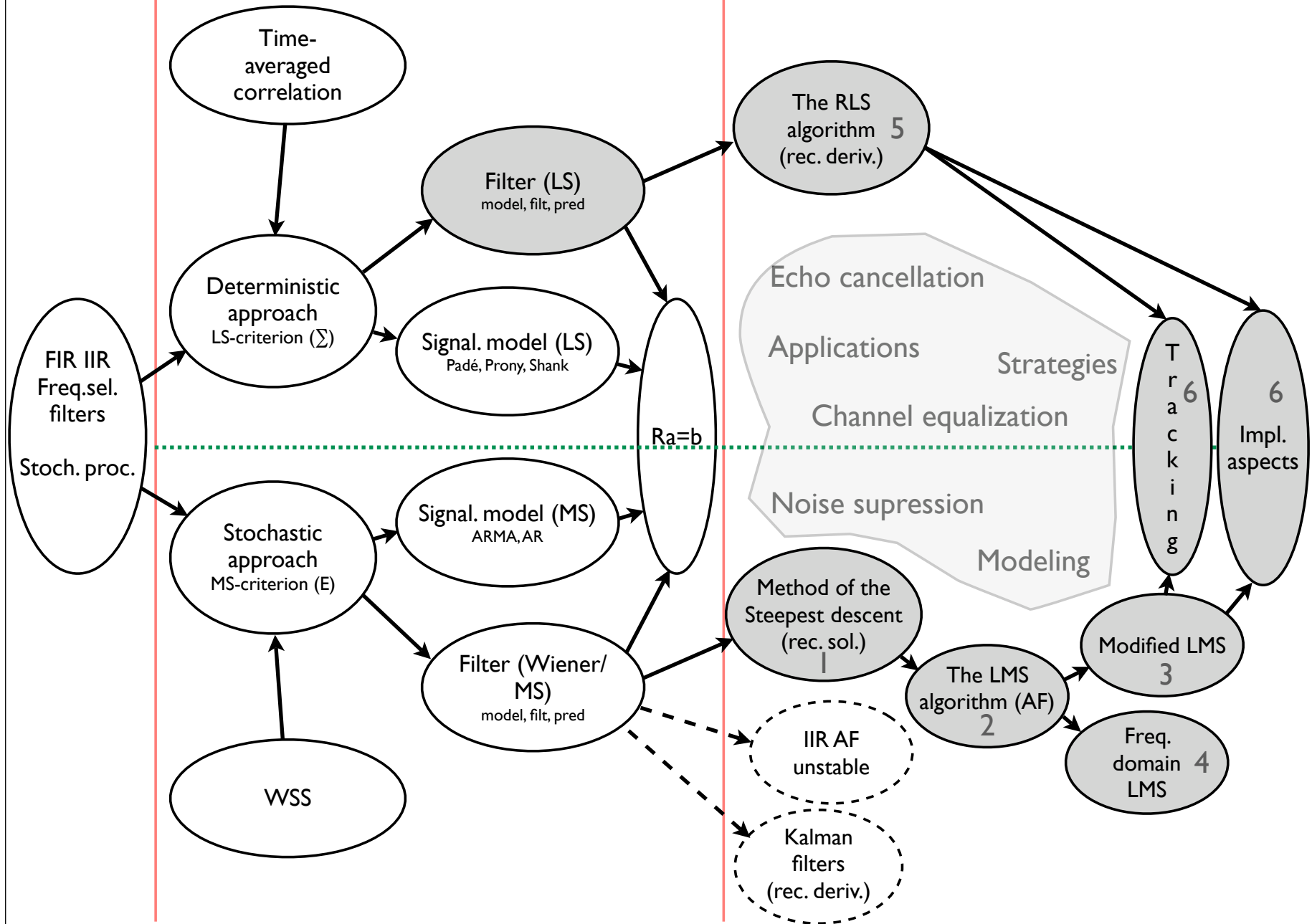
och  $\Phi_{du}(l) = \sum_{i=1}^{i_2} d(n)u(n-l) = \sum_{i=1}^{i_2} u(n-l)d(n) = z(l)$

# What do you learn in ASB?

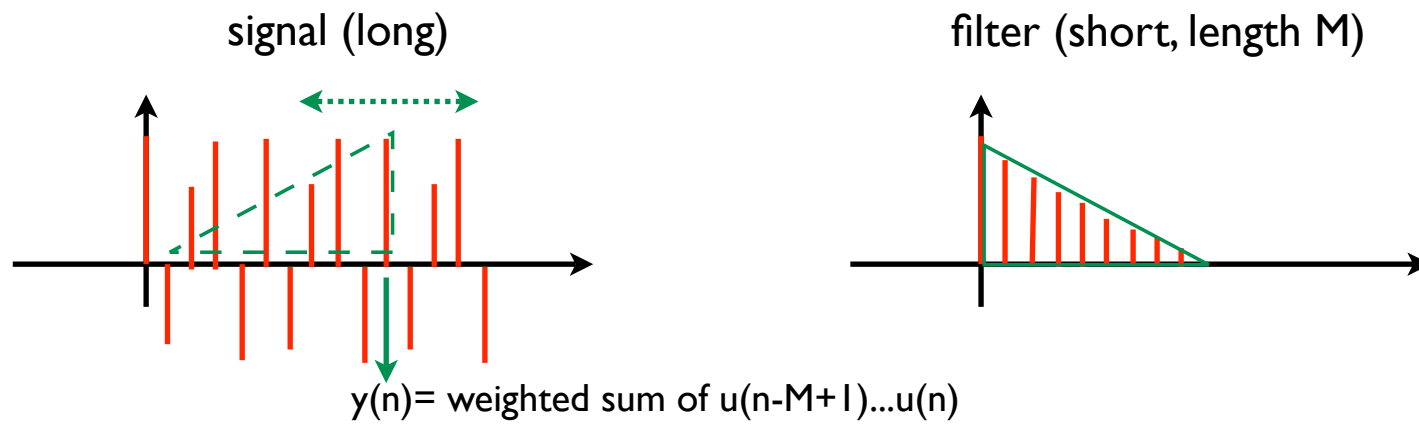


## OSB (stationary known statistics)

## ASB (self-tuning, nonstationarity)

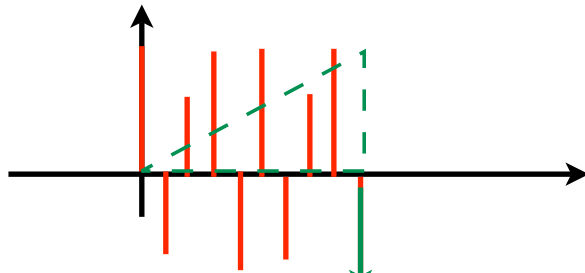


# Linear convolution



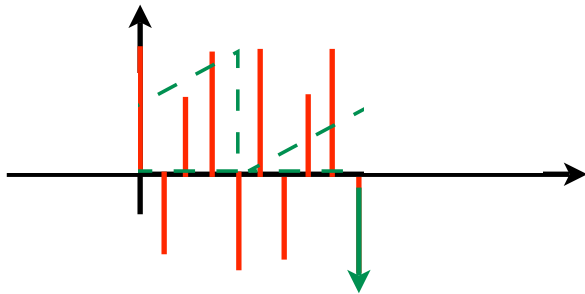
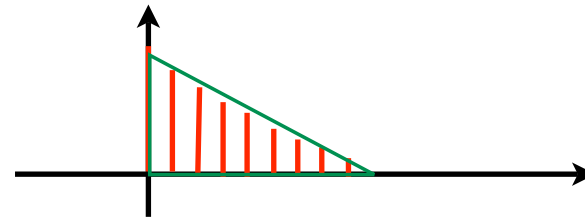
# Circular convolution

signal 1



$$y(n) = \text{weighted sum of } u(n-M+1) \dots u(n)$$

signal2/filter

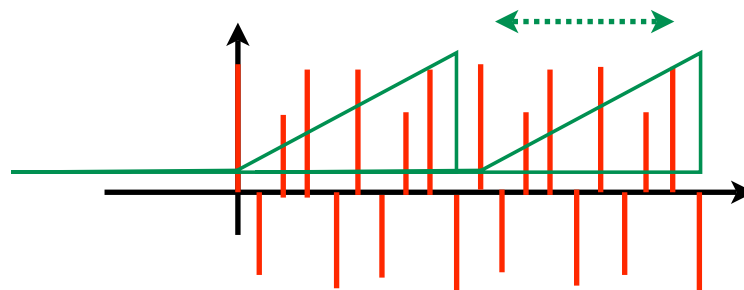
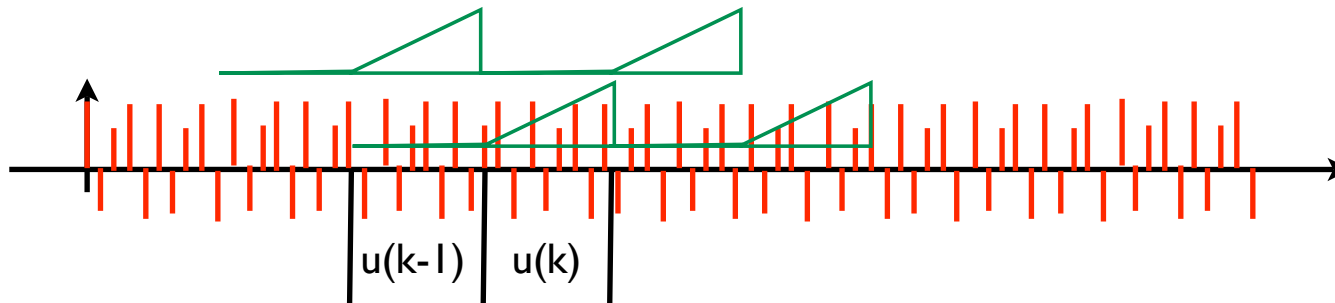


$$u(n) * w(n) = [ \dots y(n) ]$$

# Linear convolution using circular convolution



Use filter of length  $2M$  (block length  $M$ )



First  $M-1$  convolutions not OK  
Last  $M+1$  convolutions OK



# Conclusions



Similarity

