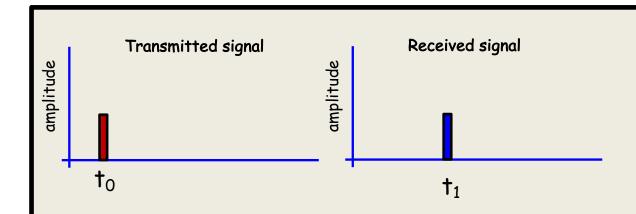


$$h(\tau, t) = \sum_{n} \alpha_n(t)\delta(t - \tau_n(t))$$

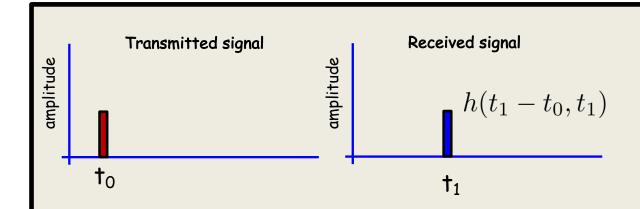
Meaning: Output at time t for input at time t- au



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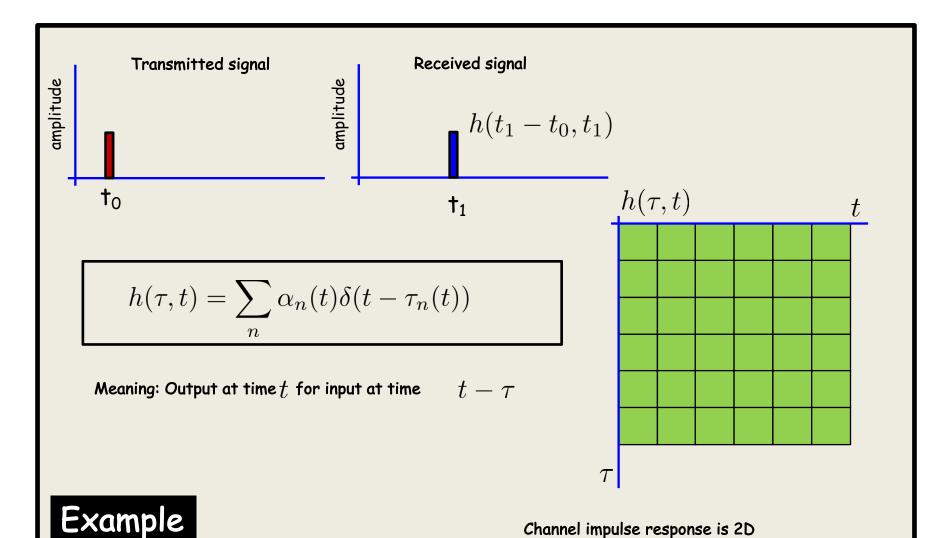
Example



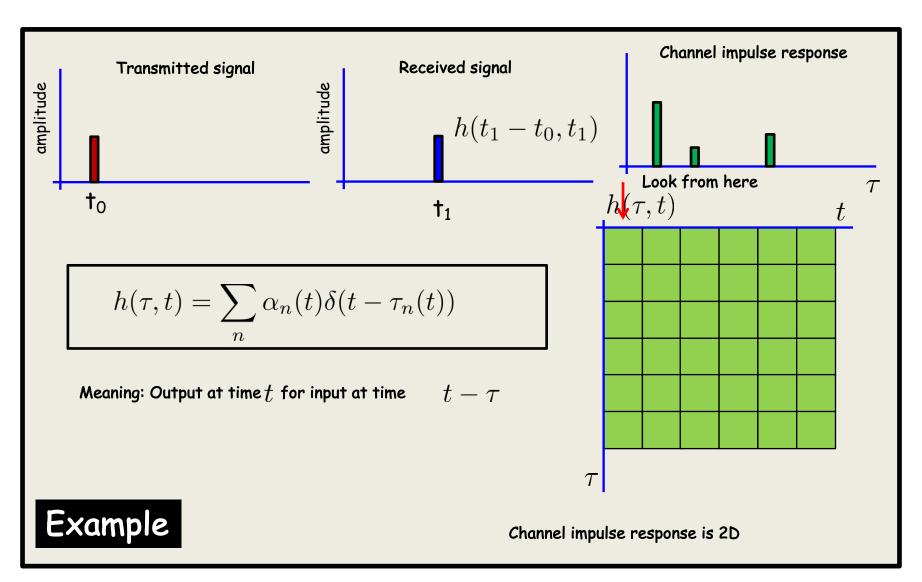
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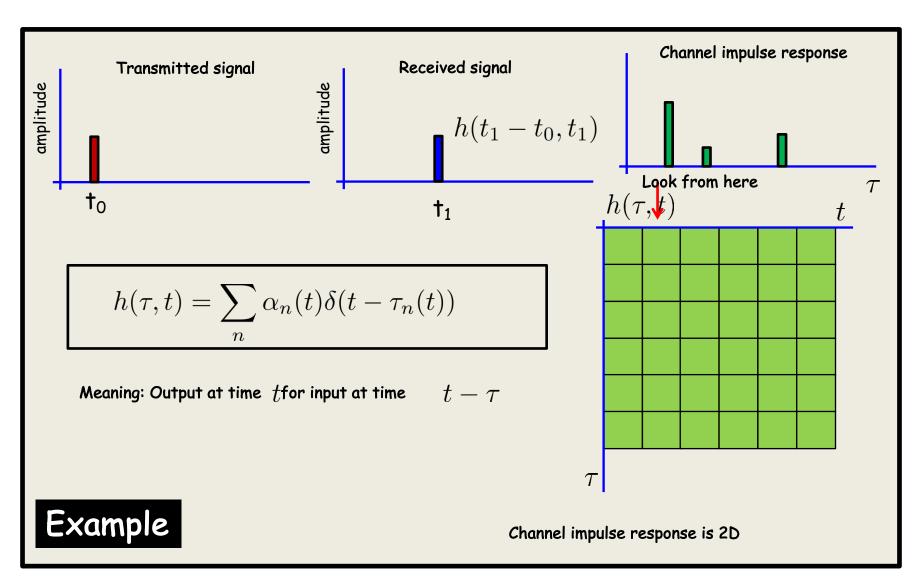
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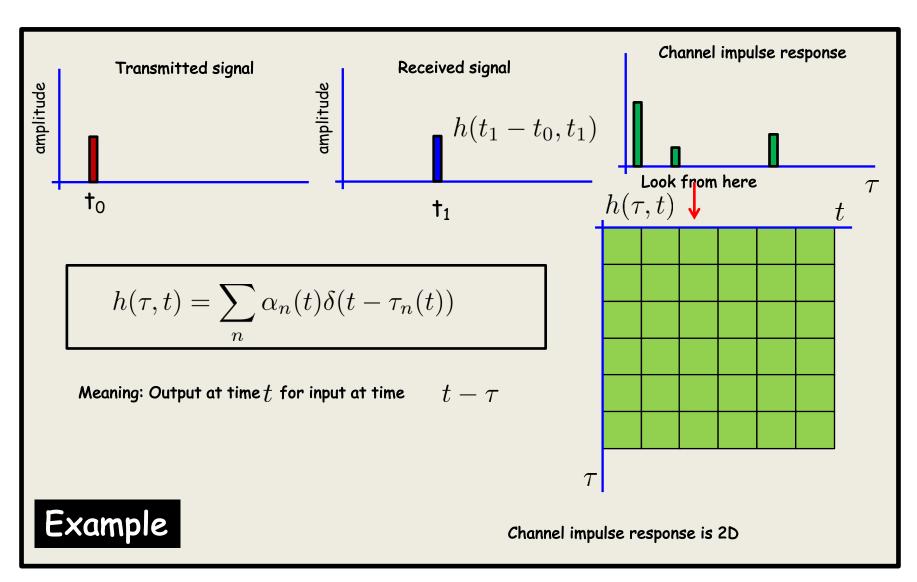
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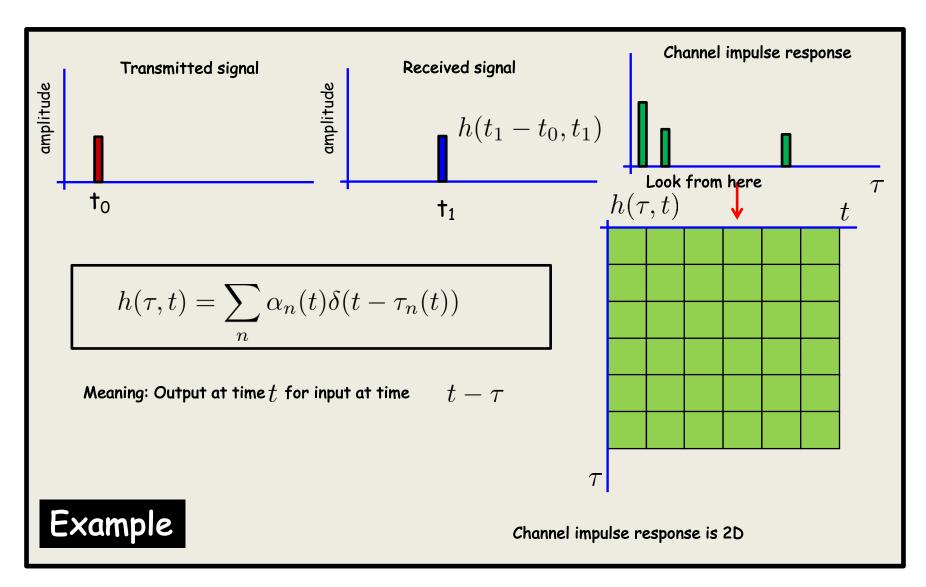


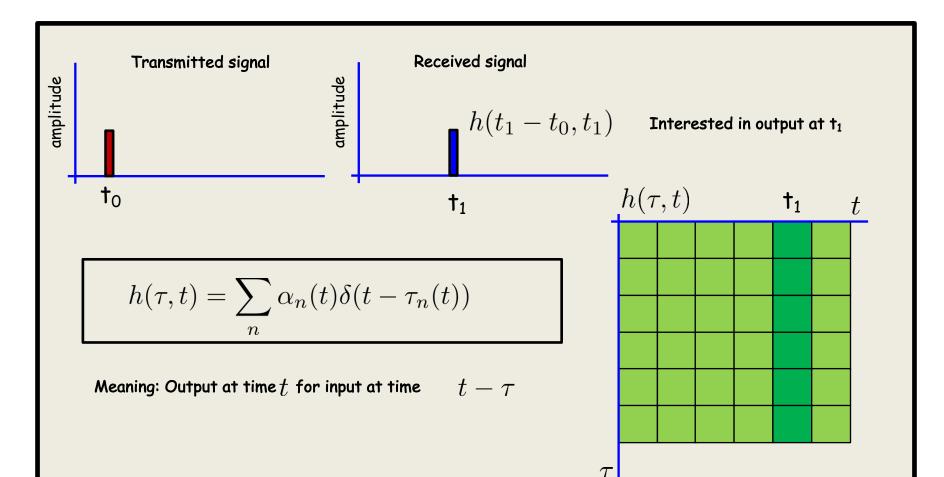
Channel impulse response is 2D



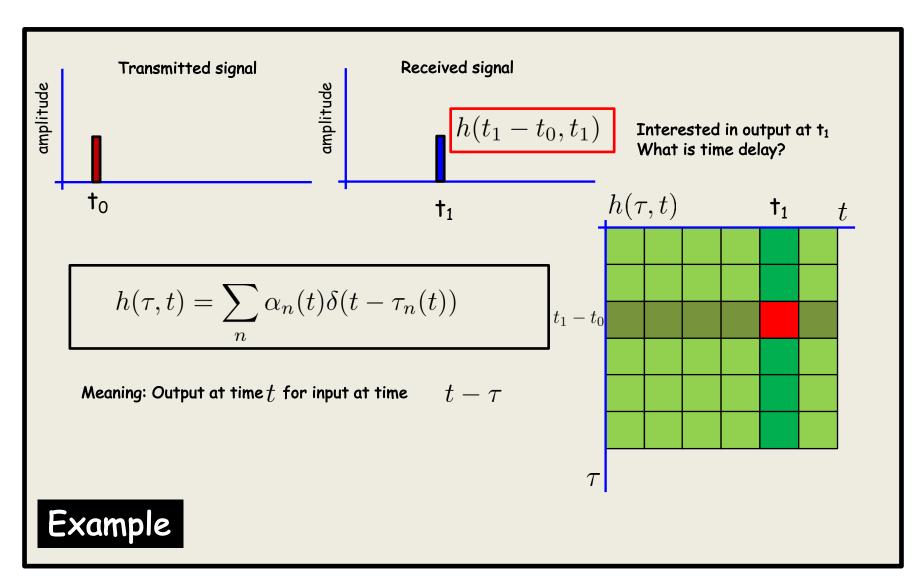


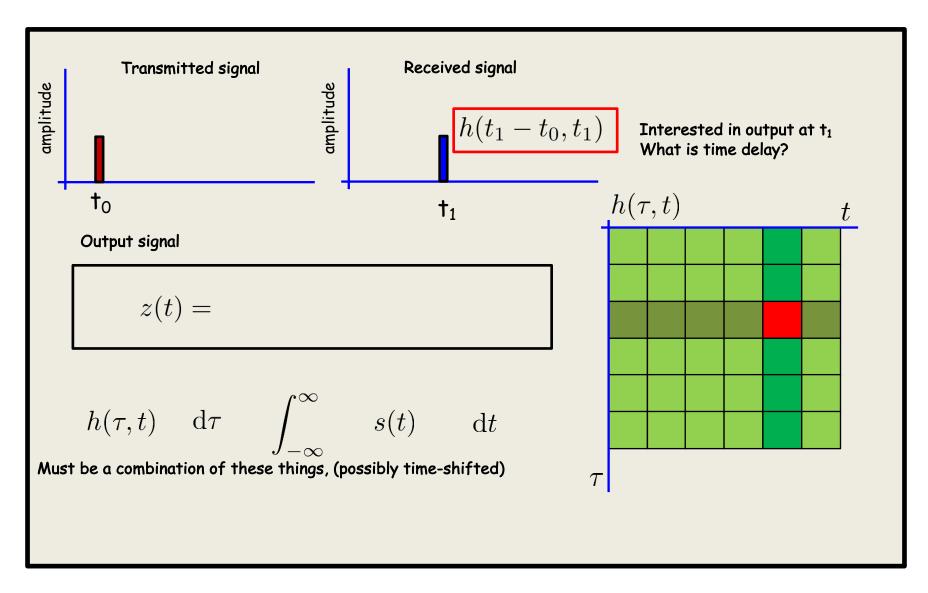


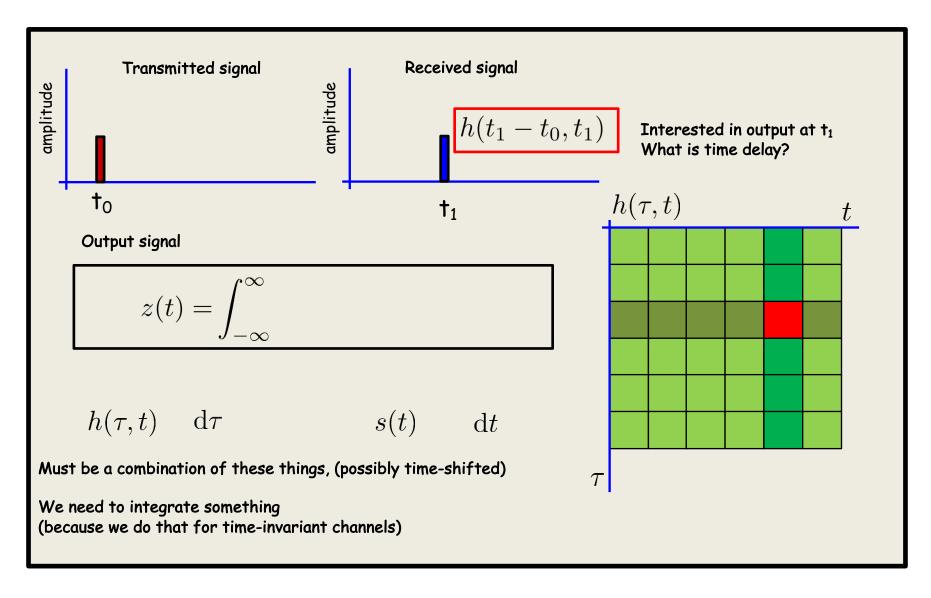


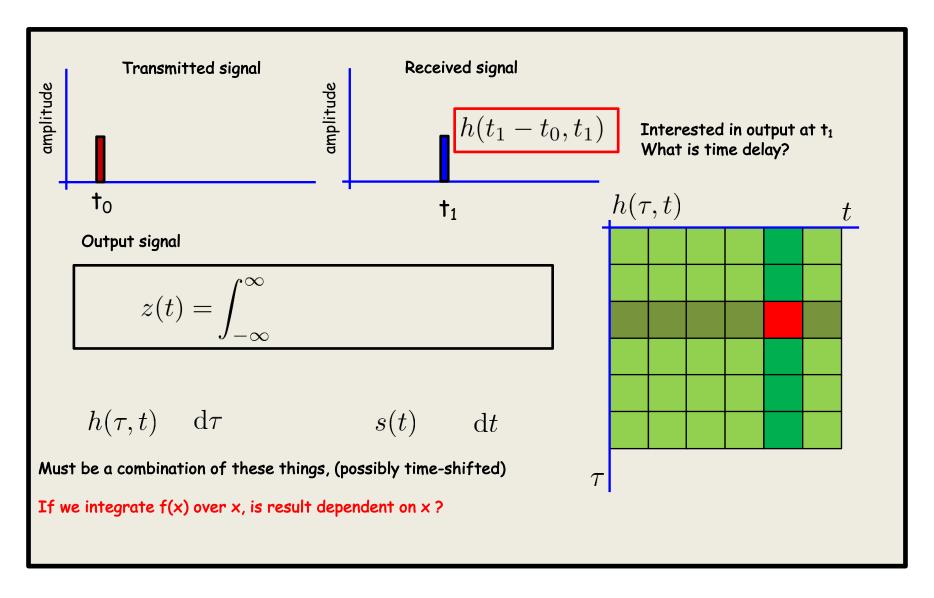


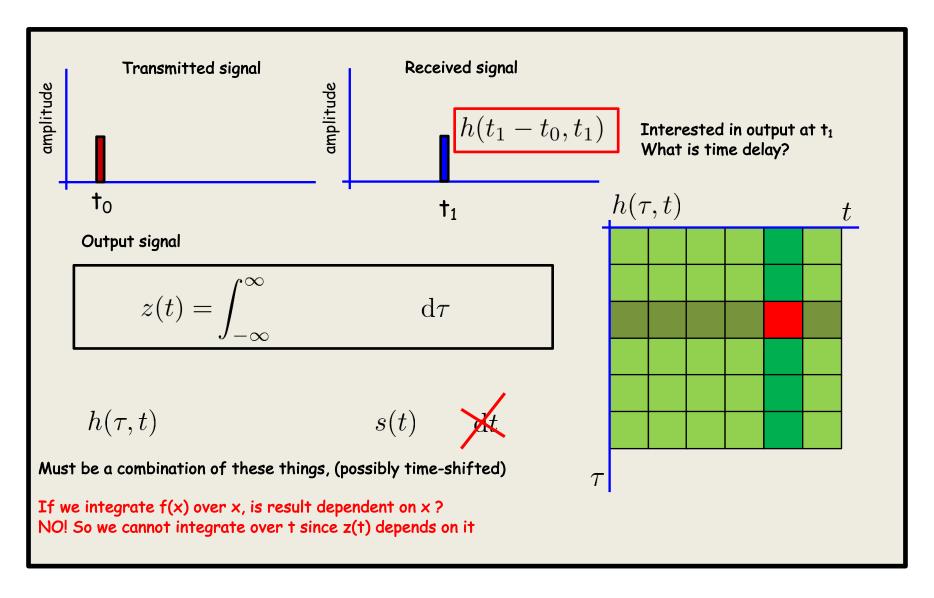
Example

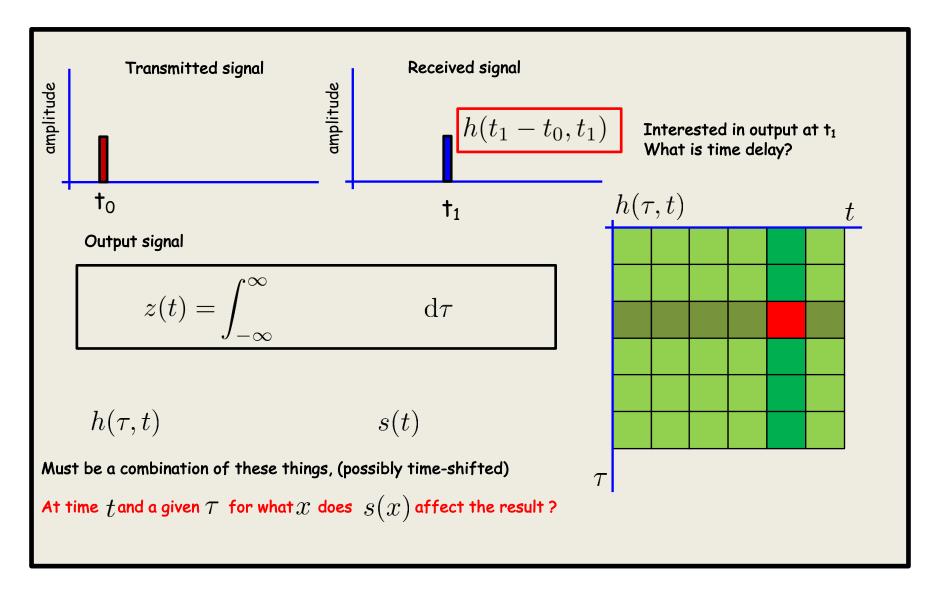


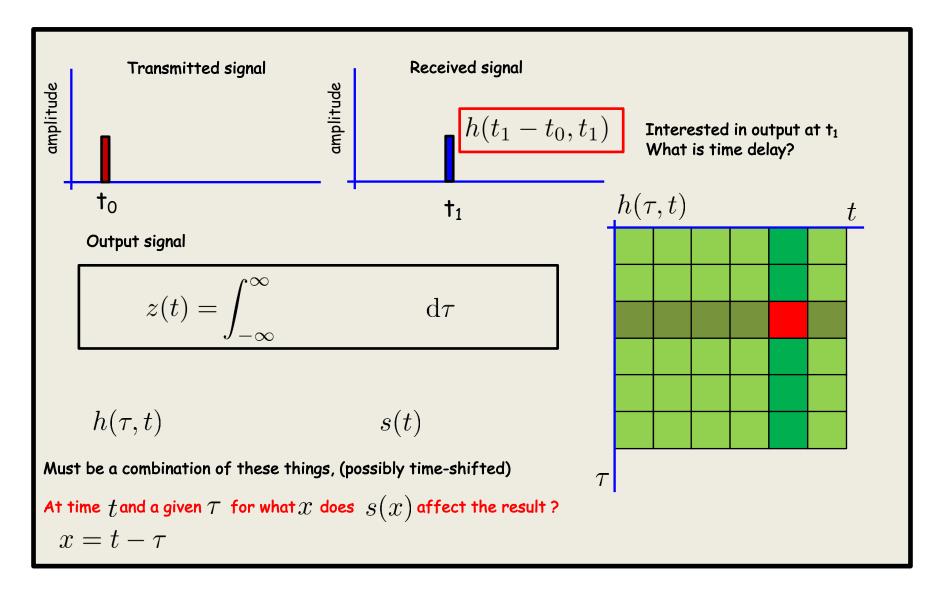


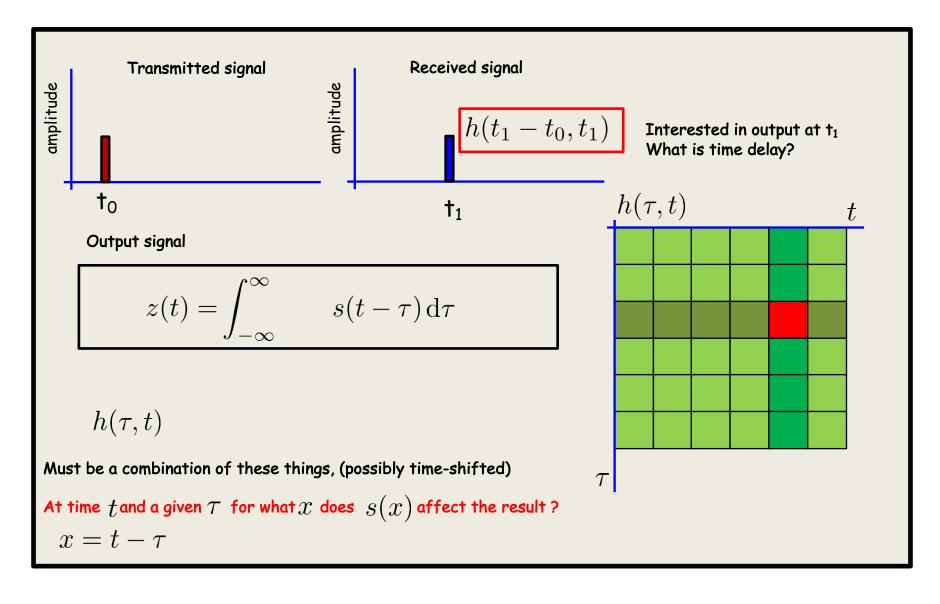


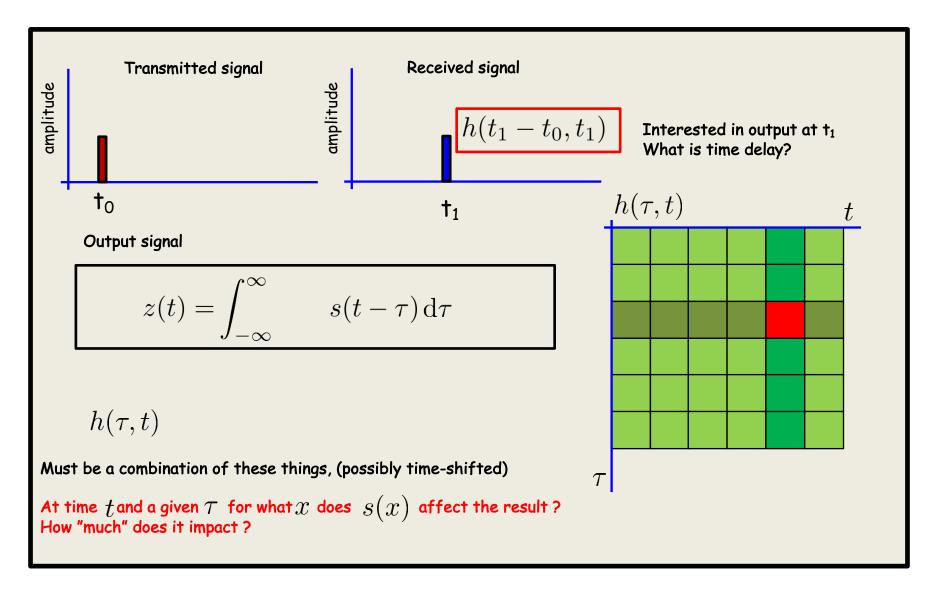


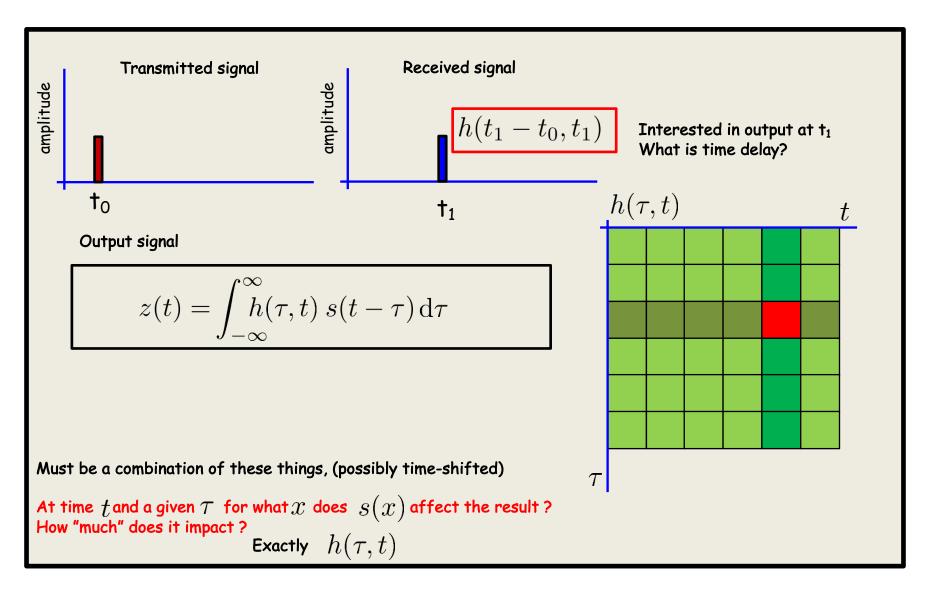


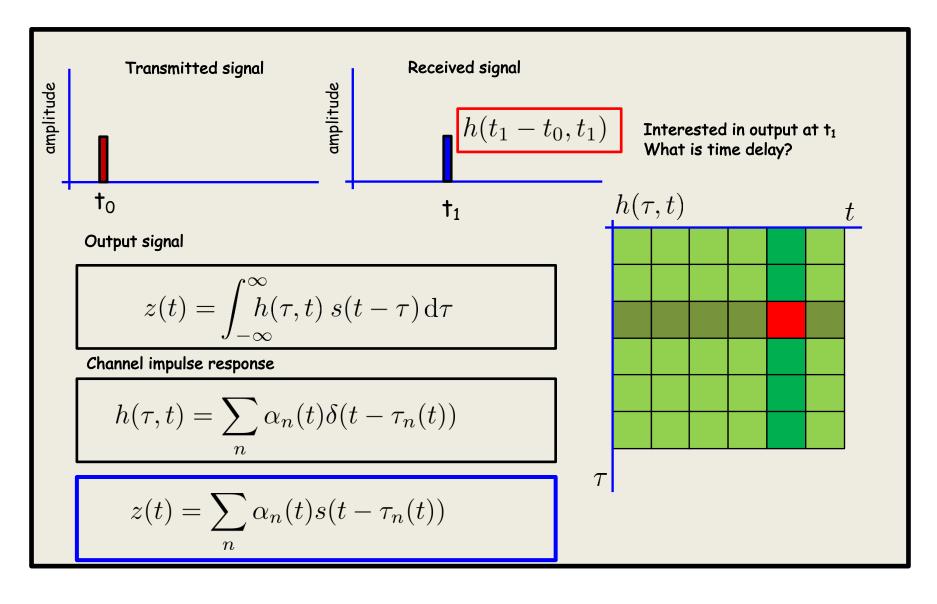












#### Assume pure cosine input

$$s(t) = \cos((\omega_c + \omega_1)t), \quad -\infty < t < \infty$$

If time-invariant channel, we get cosine out at same frequency

$$z(t) = \sum_{n} \alpha_n(t) s(t - \tau_n(t))$$

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\*\*

 $s(t) = \cos((\omega_c + \omega_1)t), \quad -\infty < t < \infty$ 

$$\cos(x - y) = \cos(x)\cos(y) + \sin(x)\sin(y)$$

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$$= \left[ \sum_{n} \alpha_n(t) \cos((\omega_c + \omega_1) \tau_n(t)) \right] \cos((\omega_c + \omega_1)t)$$

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# Assume pure cosine input If time-invariant channel, we get cosine out at same frequency $s(t) = \cos((\omega_c + \omega_1)t), \quad -\infty < t < \infty$ $z(t) = \sum \alpha_n(t) \cos((\omega_c + \omega_1)(t - \tau_n(t)))$ $= \sum_{n=0}^{\infty} \alpha_n(t) \cos((\omega_c + \omega_1)t - (\omega_c + \omega_1)\tau_n(t))$ $= \sum_{n} \alpha_n(t) \cos((\omega_c + \omega_1)\tau_n(t)) \cos((\omega_c + \omega_1)t)$ + $\sum \alpha_n(t) \sin((\omega_c + \omega_1)\tau_n(t)) \sin((\omega_c + \omega_1)t)$ $= z_I(t)\cos((\omega_c + \omega_1)t) - z_O(t)\sin((\omega_c + \omega_1)t)$ $= e_z(t)\cos((\omega_c + \omega_1)t + \theta_z(t))$

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 $z_{I}(t)$   $z_{Q}(t)$ 

Baseband signals are time-variant

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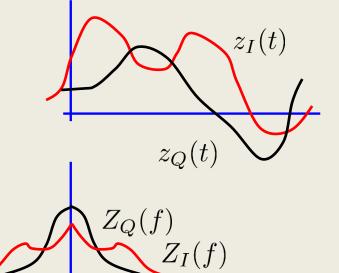
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Baseband signals are time-variant

Fourier transforms have have spread



A pure cosine has spread to other frequencies

#### Gaussian assumption

$$z_I(t) = \sum_n \alpha_n(t) \cos((\omega_c + \omega_1)\tau_n(t))$$

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We assume that both  $z_I(t)$  and  $z_Q(t)$  are Gaussian distributed with mean 0 and variance  $\sigma^2$ 

Envelope is Rayleigh distributed 
$$\ e_z(t) = \sqrt{z_I^2(t) + z_Q^2(t)}$$

We would like to understand how severe the spectral broadening is

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We would like to understand how severe the spectral broadening is

Intuatively, if the channel changes fast, there is a lot of broadening

How to measure "how fast something changes"

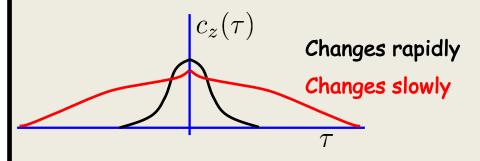
#### Gaussian assumption

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#### Covariance function

$$c_z(\tau) = \frac{1}{4} E \{ [z_I(t+\tau) + j z_Q(t+\tau)] [z_I(t) - j z_Q(t)] \}$$



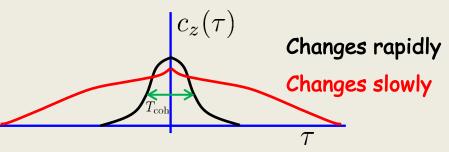
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 $c_z(\tau)$ 

Basic engineering: inverse relationship btw time and frequency
Changes slowly

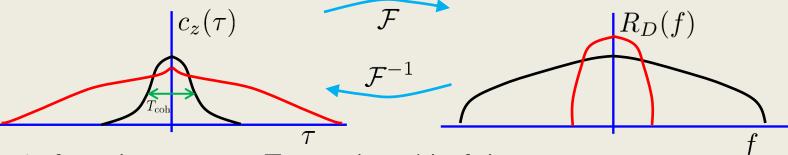
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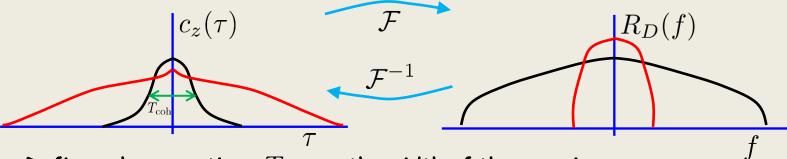
From Dig.com 1:

Fourier transform of covariance function is

Power Spectral Density (PSD)

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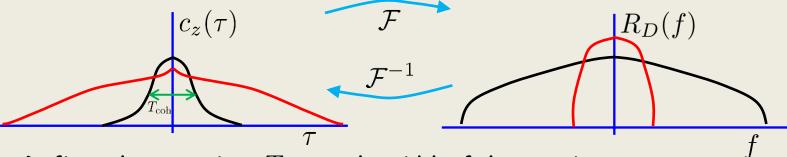
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Right plot tell us how power is being spread due to time-variance

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What causes time-variance: Doppler

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Fourier transform of covariance

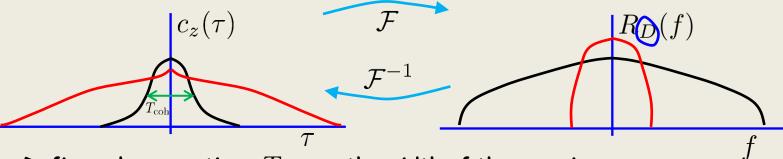
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What causes time-variance: Doppler

Width is called Doppler spread  $B_{\mathcal{D}}$ 

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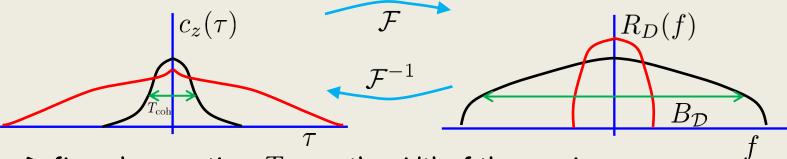
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We have, roughly,  $t_{
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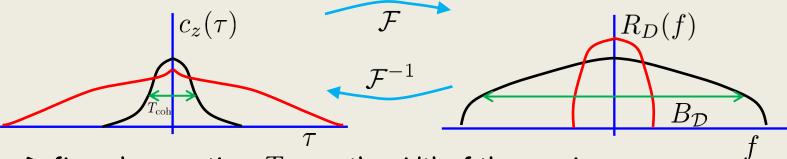
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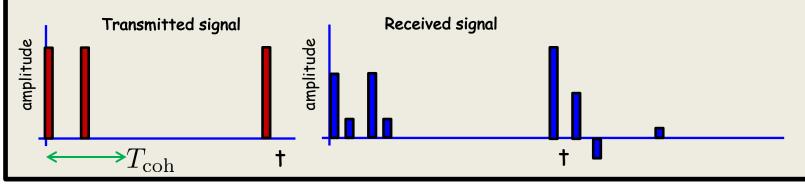
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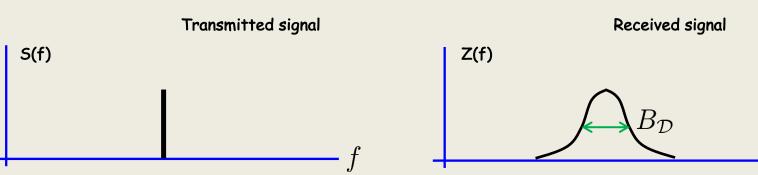
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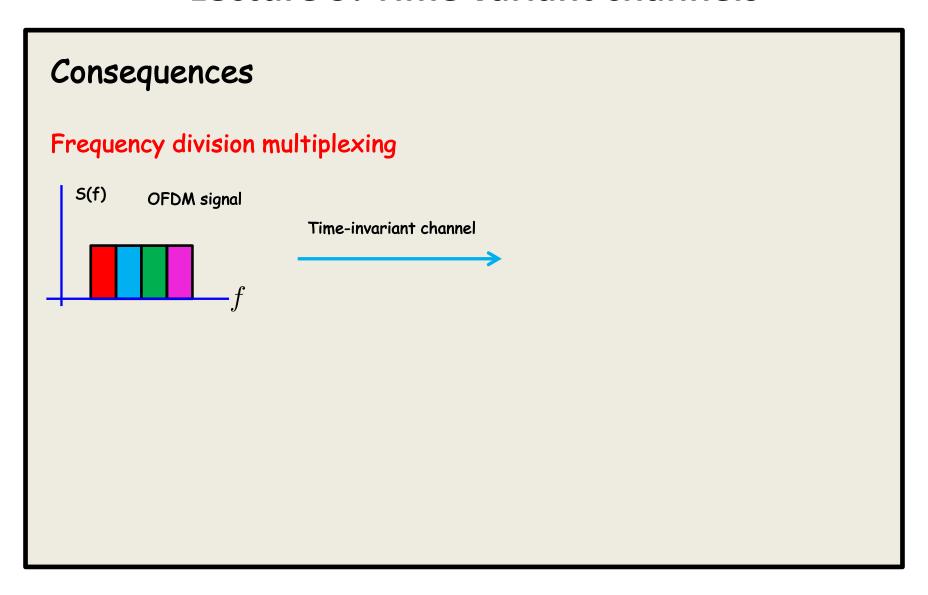
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- $\bullet$  We measure how fast the channel changes by coherence time  $t_{\text{coh}}$

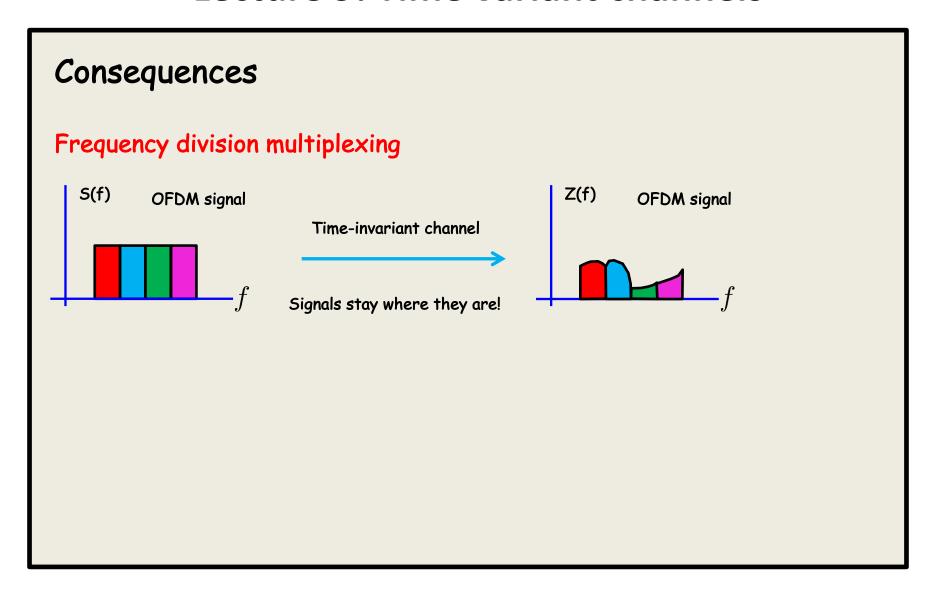


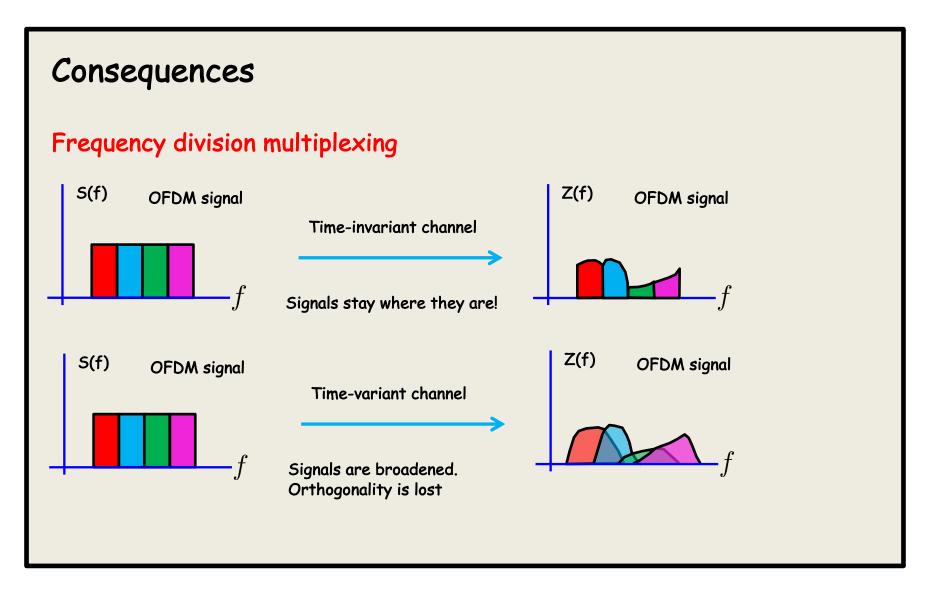
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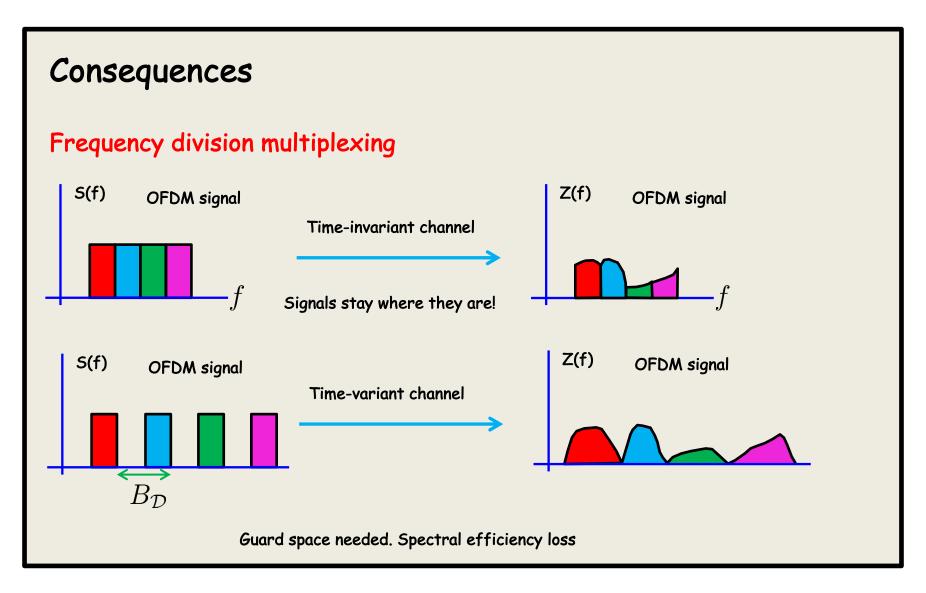


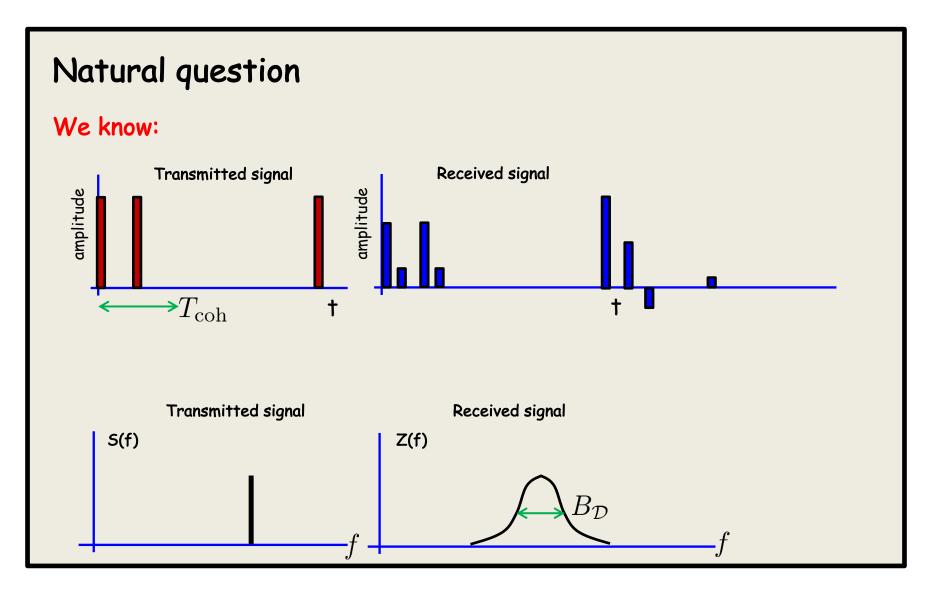
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- We have  $\,t_{
  m coh}pprox rac{1}{B_{\mathcal{D}}}$
- In industrial simulations,  $B_{\mathcal{D}}$  is varied from low to high, thus it is an input parameter to a system

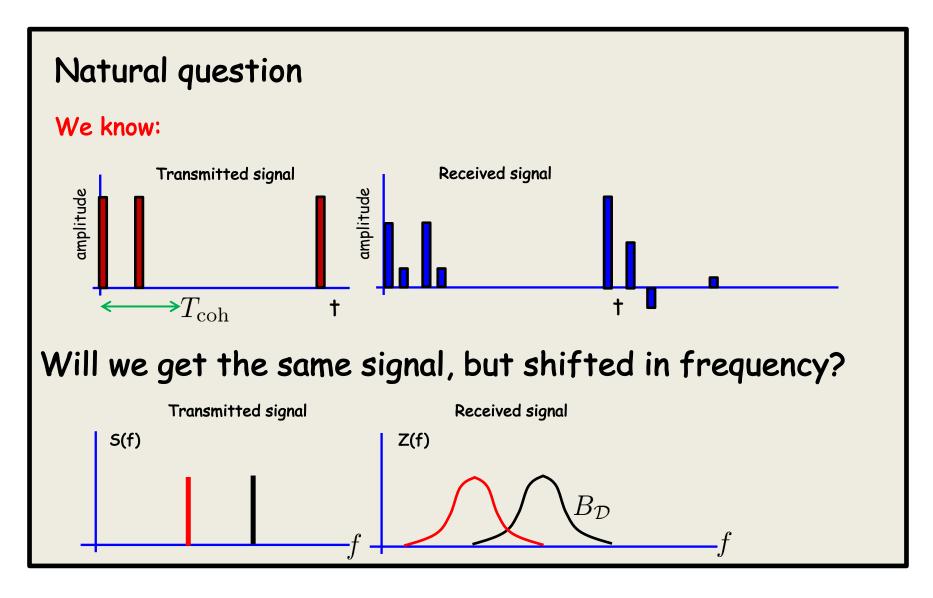


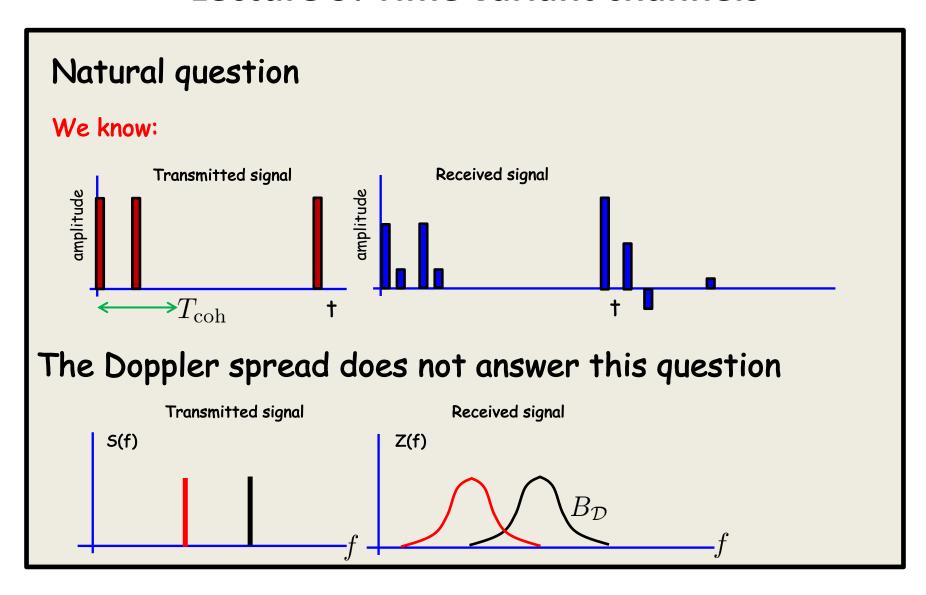








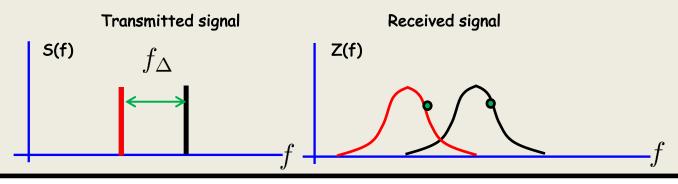




## New concept: Coherence bandwidth

Compute covariance between the points generated from 2 signals at distance  $f_{\Delta}$ 

$$\tilde{c}_z(f_\Delta) = E\left\{z(f, t)z^*(f + f_\Delta, t)\right\}$$

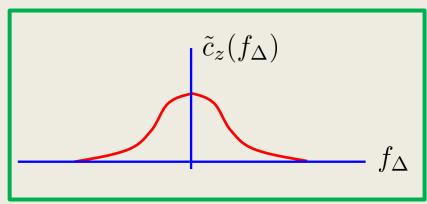


$$f f + f_{\Delta}$$

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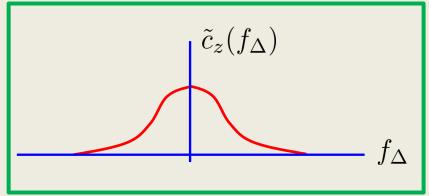
Frequency autocorrelation function

## New concept: Coherence bandwidth

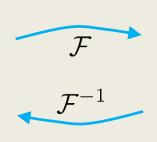
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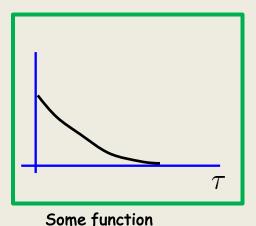
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Frequency autocorrelation function



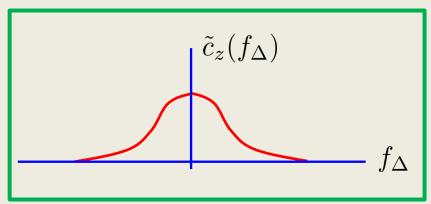


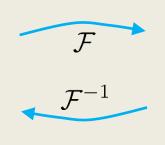
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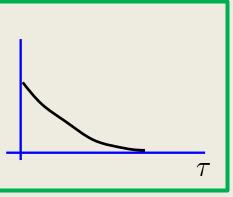
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Frequency autocorrelation function

delay power spectrum

Can be shown to be the delay power spectrum

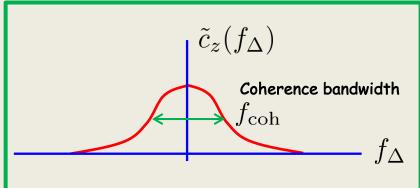
$$c_h(\tau) = E\left\{h^2(\tau, t)\right\}$$

## New concept: Coherence bandwidth

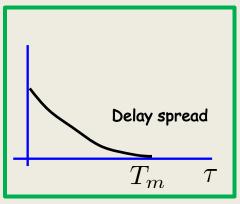
Compute covariance between the points generated from 2 signals at distance

$$f_{\Delta}$$

$$\tilde{c}_z(f_\Delta) = E\left\{z(f, t)z^*(f + f_\Delta, t)\right\}$$







delay power spectrum

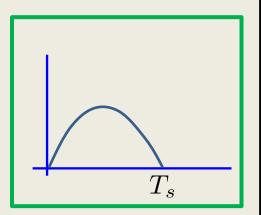
Frequency autocorrelation function

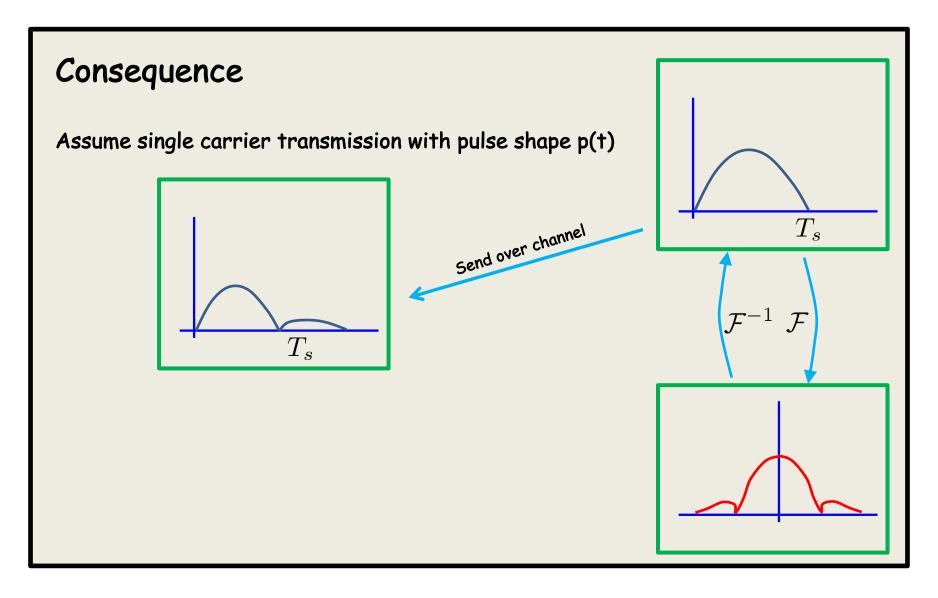
Can be shown to be the delay power spectrum

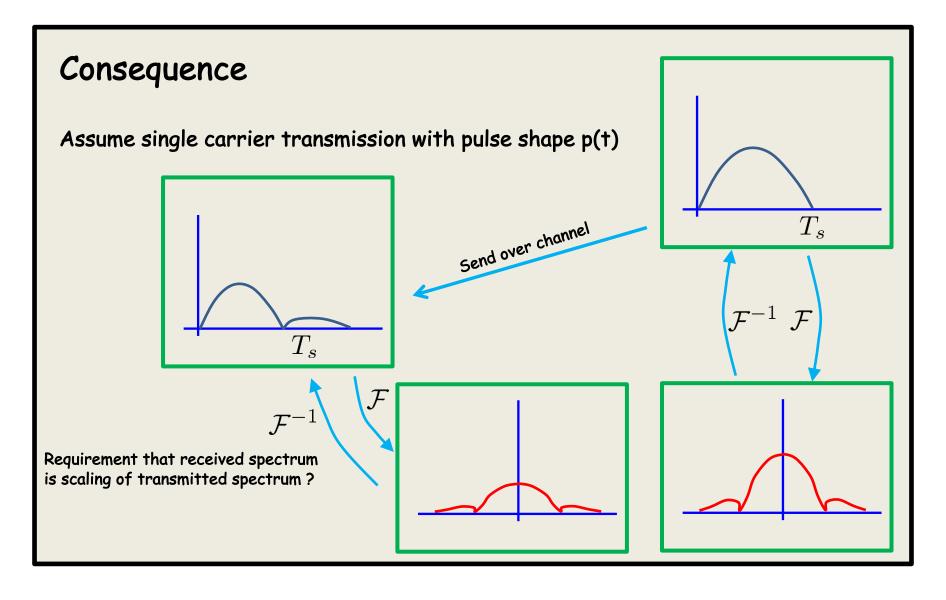
$$c_h(\tau) = E\left\{h^2(\tau, t)\right\}$$

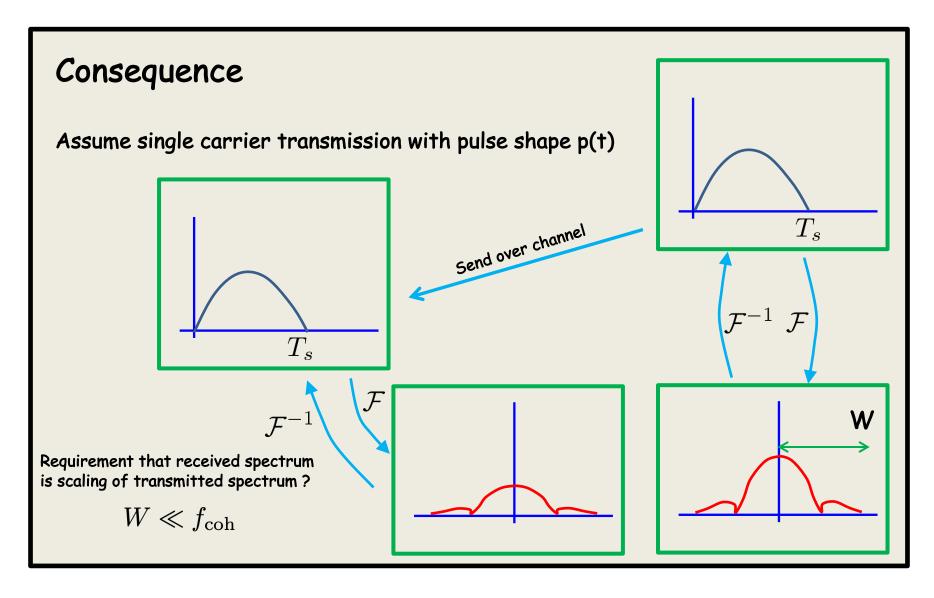
## Consequence

Assume single carrier transmission with pulse shape p(t)









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Significantly simplifed modelling. For complex basesband, signals are mutltiplied with a complex constant

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NO. Note that  $B_{\mathcal{D}}T_m$  is a channel parameter, out of our control

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a Rayleigh  $\phi$  Uniform

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