

Lecture 7: OFDM

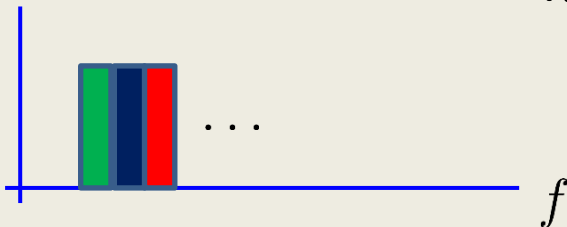
Recall: If $x(t) = A \cos(\omega_c t) - B \sin(\omega_c t)$, $0 \leq t \leq T_s$

Then $z(t) = A_z \cos(\omega_c t) - B_z \sin(\omega_c t)$, $T_h \leq t \leq T_s$

$$A_z + iB_z = (A + iB)H(\omega_c)$$

**We can transmit multiple signals at different sub-carriers:
This is OFDM**

We have



$$\text{OFDM signal} = g_{\text{rec}}(t) \sum_{k=0}^{K-1} \text{Re}\{a_k \exp(i2\pi f_k t)\}$$

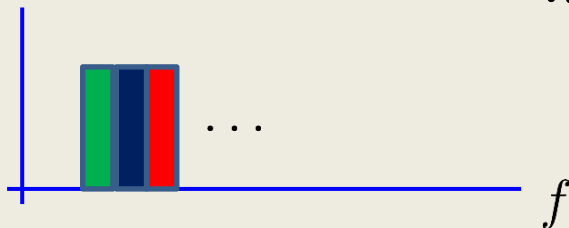
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Complex data (A+iB)

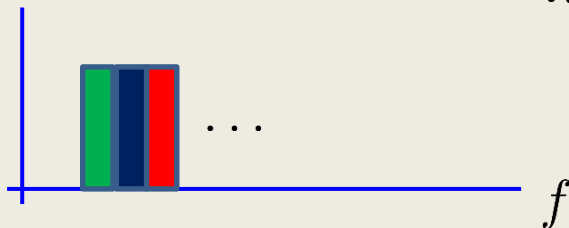
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Complex data (A+iB)

Carrier



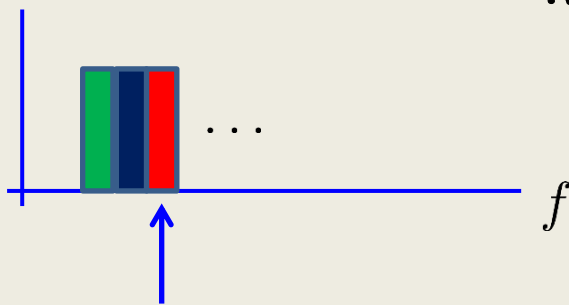
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One signal

A red oval highlights the term $\text{Re}\{a_k \exp(i2\pi f_k t)\}$ in the equation. A blue arrow points from the text "One signal" above to this term.

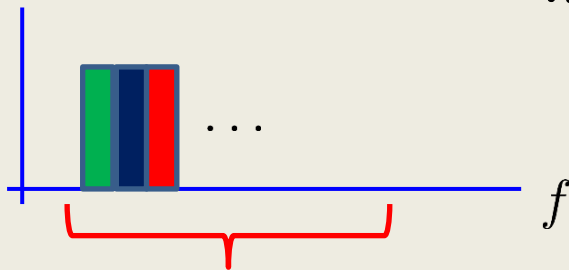
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All signals



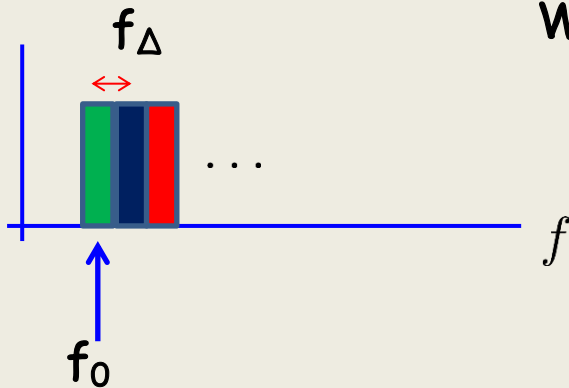
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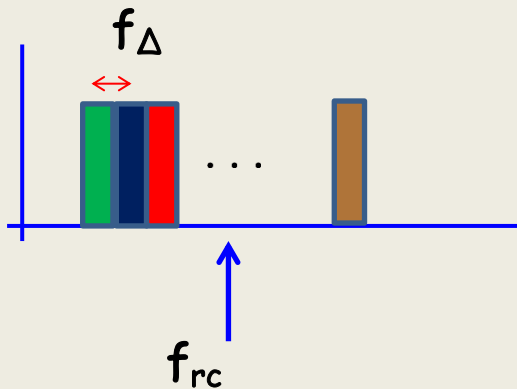
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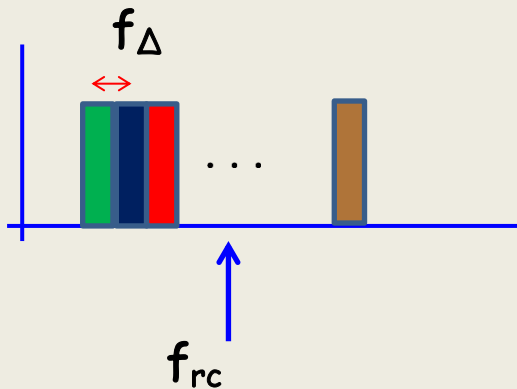
$$g_k = \dots -2, -1, 0, 1, 2, \dots$$

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$$g_k: -\frac{K-1}{2} = g_0, \dots, -1, 0, 1, \dots, \frac{K-1}{2} = g_{K-1} \quad \text{if } K \text{ is odd}$$

$$g_k: -\frac{K-2}{2} = g_0, \dots, -1, 0, 1, \dots, \frac{K}{2} = g_{K-1} \quad \text{if } K \text{ is even}$$

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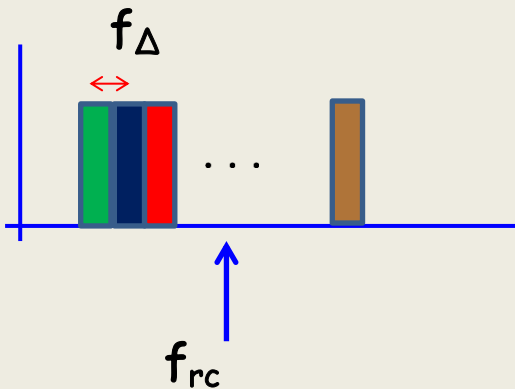
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$$\text{OFDM signal} = \text{Re}\{x(t) \exp(i2\pi f_{rc}t)\} \quad 0 \leq t \leq T_s$$

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$$x(t) = g_{\text{rec}}(t) \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}t)$$

Useful part at rx $T_h \leq t \leq T_s$

Definition $T_{\text{obs}} = T_s - T_h$

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So

$$\text{OFDM signal} = \text{Re}\{x(t) \exp(i2\pi f_{rc}t)\} \quad 0 \leq t \leq T_{\text{obs}} + T_h$$

We can also use (time-shift)

$$\text{OFDM signal} = \text{Re}\{x(t) \exp(i2\pi f_{rc}t)\} \quad -T_h \leq t \leq T_{\text{obs}}$$

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Objective of OFDM:

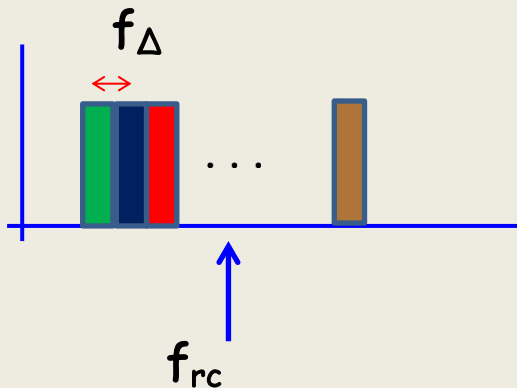
1. Create as many signal dimensions as there are symbols to be sent
2. Put one, and only one, symbol in each dimension (simple detection)

How many symbols? $2K$

How long time is used at the receiver? T_{obs}

How much bandwidth? $K f_{\Delta}$

Shannon? Max #dims = $2 \times \text{Time} \times \text{Bandwidth} = 2KT_{\text{obs}}f_{\Delta}$



$$\text{OFDM signal} = g_{\text{rec}}(t) \sum_{k=0}^{K-1} \text{Re}\{a_k \exp(i2\pi f_k t)\}$$

$$= g_{\text{rec}}(t) \sum_{k=0}^{K-1} \text{Re}\{a_k \exp(i2\pi g_k f_{\Delta} t) \exp(i2\pi f_{rc} t)\}$$

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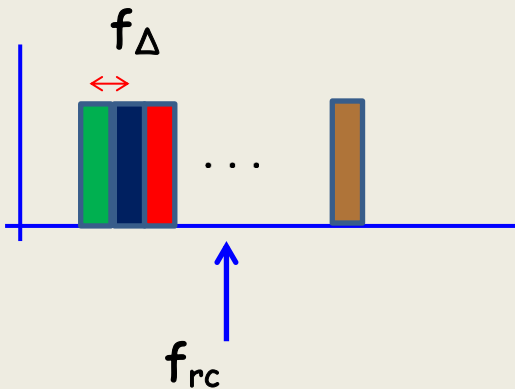
OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$ $0 \leq t \leq T_{\text{obs}}$

$$x(t) = g_{\text{rec}}(t) \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta} t)$$

Let's begin here

Important: $f_{\Delta} T_{\text{obs}} > 1$

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To synthesize the signal, we

1. Sample the OFDM signal
2. Check how we can efficiently construct those samples
3. Perform D/A conversion

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Sampling theorem: Sample twice as fast the highest frequency component

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OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$ $0 \leq t \leq T_{obs}$

$$x(t) = g_{rec}(t) \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta} t)$$

$$x_n = x(nT_{obs}/N) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k f_{\Delta} nT_{obs}}{N}\right)$$

$$f_{samp} = N/T_{obs} = N f_{\Delta}$$

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OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rct}t)\}$ $0 \leq t \leq T_{obs}$

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$$\begin{aligned} x_n = x(nT_{obs}/N) &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k f_{\Delta} nT_{obs}}{N}\right) \\ &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) \end{aligned}$$

The above gives a formula for how to compute the samples of the OFDM signal

Lecture 7: OFDM

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Let us now compute the Fourier transform of the samples
(as of now, for no particular reason)

$$X(\nu) = \sum_{n=0}^{N-1} x_n \exp(-i2\pi\nu n)$$

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Take N samples of this Fourier transform
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$$X_m = X(m/N) = \sum_{n=0}^{N-1} x_n \exp\left(-\frac{i2\pi mn}{N}\right)$$

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DFT

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

IDFT

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IDFT is VERY fast.
We can get x_n FAST if
we know X_m

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IDFT

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Logics

We have $\{a_n\}$

We need $\{x_n\}$

We know that
$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

And that this can be computed FAST

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Conclusion: We need to link $\{a_n\}$ and $\{X_m\}$

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$$x_n = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right)$$

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$$\begin{aligned}x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\ &= \sum_{k=0}^{-g_0-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) + \sum_{k=-g_0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right)\end{aligned}$$

$g_0 < 0$

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$g_0 < 0$

Variable substitutions

$$\left[m = g_0 + k + N \right] \qquad \left[m = g_0 + k \right]$$

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$$\begin{aligned}
 x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\
 &= \sum_{k=0}^{-g_0-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) + \sum_{k=-g_0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \quad g_0 < 0 \\
 &= \sum_{k=0}^{-g_0-1} a_k \exp\left(\frac{i2\pi(g_0 + k + N)n}{N}\right) + \sum_{k=-g_0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\
 &\quad \text{Variable substitutions} \\
 &\quad [m = g_0 + k + N] \quad [m = g_0 + k] \\
 &= \sum_{m=g_0+N}^{N-1} a_{m-(g_0+N)} \exp\left(\frac{i2\pi mn}{N}\right) + \sum_{m=0}^{g_{K-1}} a_{m-g_0} \exp\left(\frac{i2\pi mn}{N}\right)
 \end{aligned}$$

Lecture 7: OFDM

$$\begin{aligned}x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\ &= \sum_{m=g_0+N}^{N-1} a_{m-(g_0+N)} \exp\left(\frac{i2\pi m n}{N}\right) + \sum_{m=0}^{g_{K-1}} a_{m-g_0} \exp\left(\frac{i2\pi m n}{N}\right)\end{aligned}$$

Lecture 7: OFDM

$$\begin{aligned}x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\&= \sum_{m=g_0+N}^{N-1} a_{m-(g_0+N)} \exp\left(\frac{i2\pi m n}{N}\right) + \sum_{m=0}^{g_{K-1}} a_{m-g_0} \exp\left(\frac{i2\pi m n}{N}\right) \\&= \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi m n}{N}\right)\end{aligned}$$

This we know from the IDFT

Lecture 7: OFDM

$$\begin{aligned}x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\&= \sum_{m=g_0+N}^{N-1} a_{m-(g_0+N)} \exp\left(\frac{i2\pi m n}{N}\right) + \sum_{m=0}^{g_{K-1}} a_{m-g_0} \exp\left(\frac{i2\pi m n}{N}\right) \\&= \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi m n}{N}\right)\end{aligned}$$

$$X_m = N a_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

Thus,

Lecture 7: OFDM

$$\begin{aligned}x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi(g_0 + k)n}{N}\right) \\&= \sum_{m=g_0+N}^{N-1} a_{m-(g_0+N)} \exp\left(\frac{i2\pi m n}{N}\right) + \sum_{m=0}^{g_{K-1}} a_{m-g_0} \exp\left(\frac{i2\pi m n}{N}\right) \\&= \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi m n}{N}\right)\end{aligned}$$

Thus,

$$X_m = N a_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = N a_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Altogether,

1. Take a block of K data symbols $\{a_k\}$

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Altogether,

1. Take a block of K data symbols $\{a_k\}$
2. Select a sampling rate, by choosing $N \geq K$

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Altogether,

1. Take a block of K data symbols $\{a_k\}$
2. Select a sampling rate, by choosing $N \geq K$
3. Find N values $\{X_m\}$ according to the box below

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Altogether,

1. Take a block of K data symbols $\{a_k\}$
2. Select a sampling rate, by choosing $N \geq K$
3. Find N values $\{X_m\}$ according to the box below
4. Compute $\{x_n\}$ using an IDFT

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Altogether,

1. Take a block of K data symbols $\{a_k\}$
2. Select a sampling rate, by choosing $N \geq K$
3. Find N values $\{X_m\}$ according to the box below
4. Compute $\{x_n\}$ using an IDFT

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

Speed: N^2 multiplications

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Altogether,

FFT = "Fast Fourier transform"

1. Take a block of K data symbols $\{a_k\}$
2. Select a sampling rate, by choosing $N=2^L \geq K$
3. Find N values $\{X_m\}$ according to the box below
4. Compute $\{x_n\}$ using an **IFFT**

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

Speed: N^2 multiplications

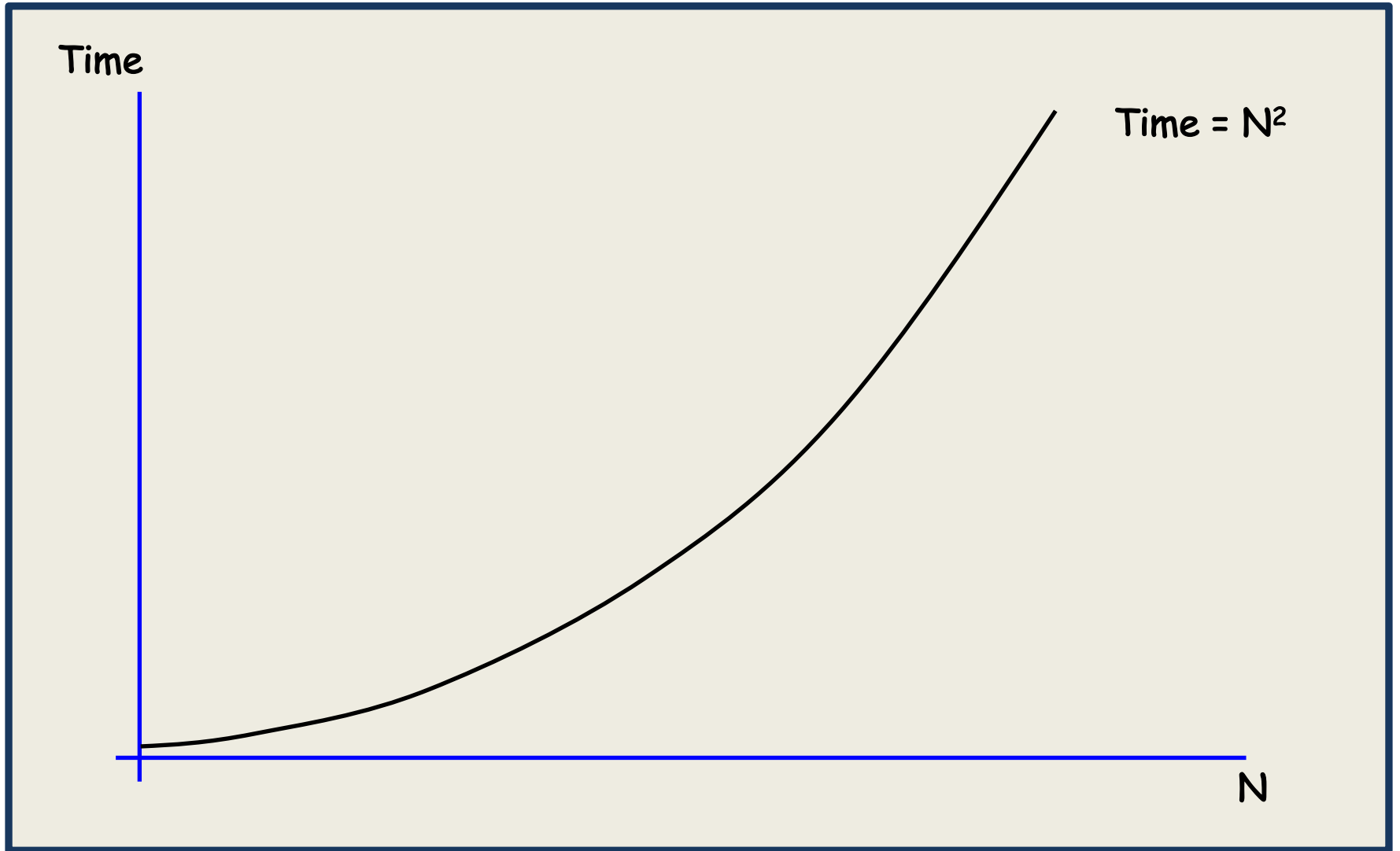
Speed: $N \log_2 N$

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

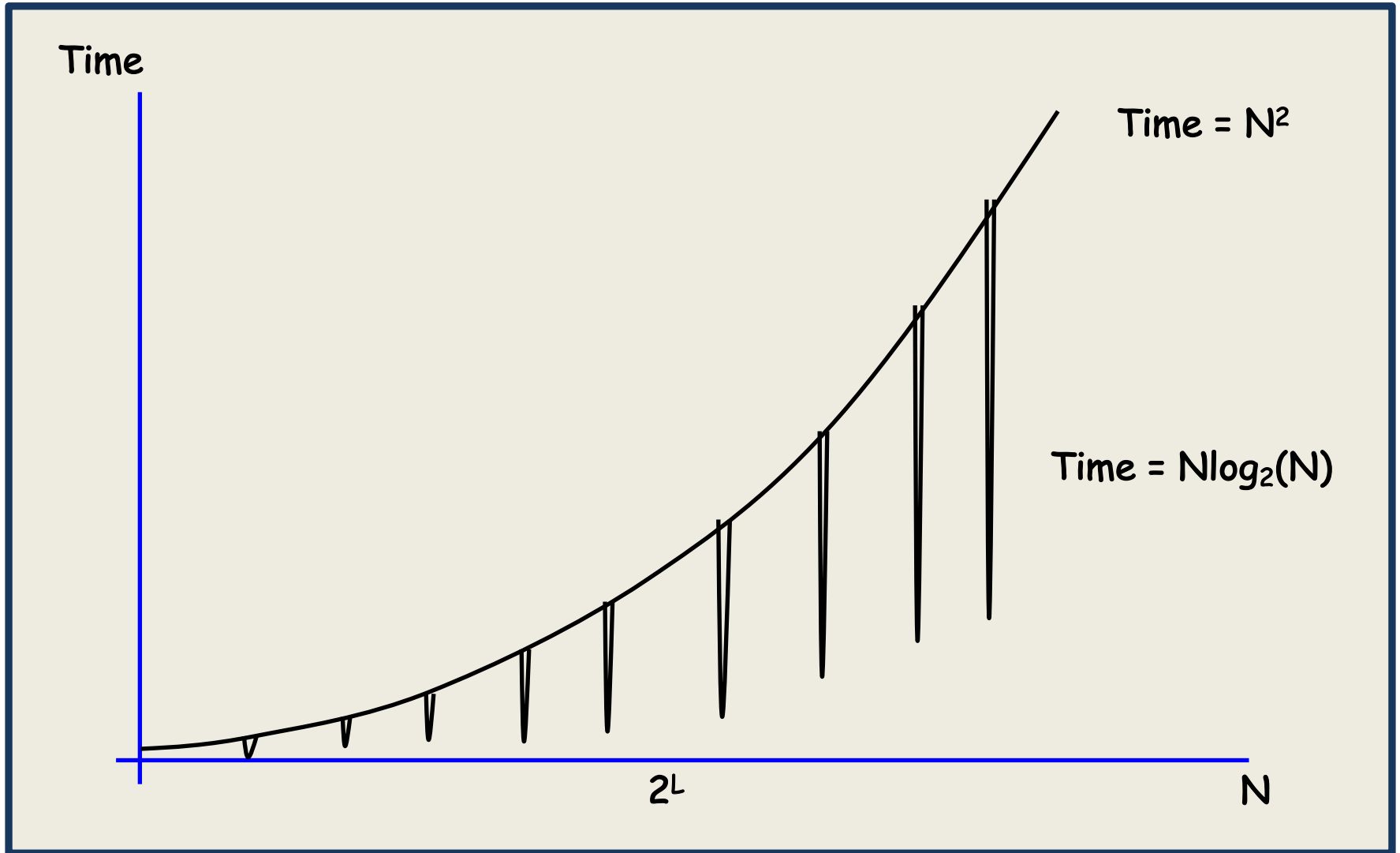
$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM



Lecture 7: OFDM



Lecture 7: OFDM

Left to do

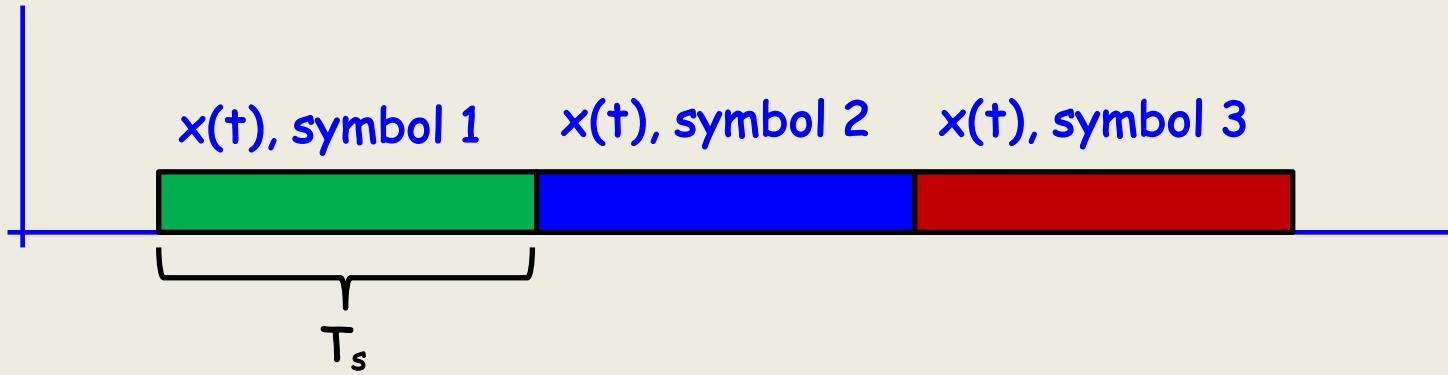
- 1.
2. D/A conversion
3. Modulation to band-pass
4. Receiver

Lecture 7: OFDM

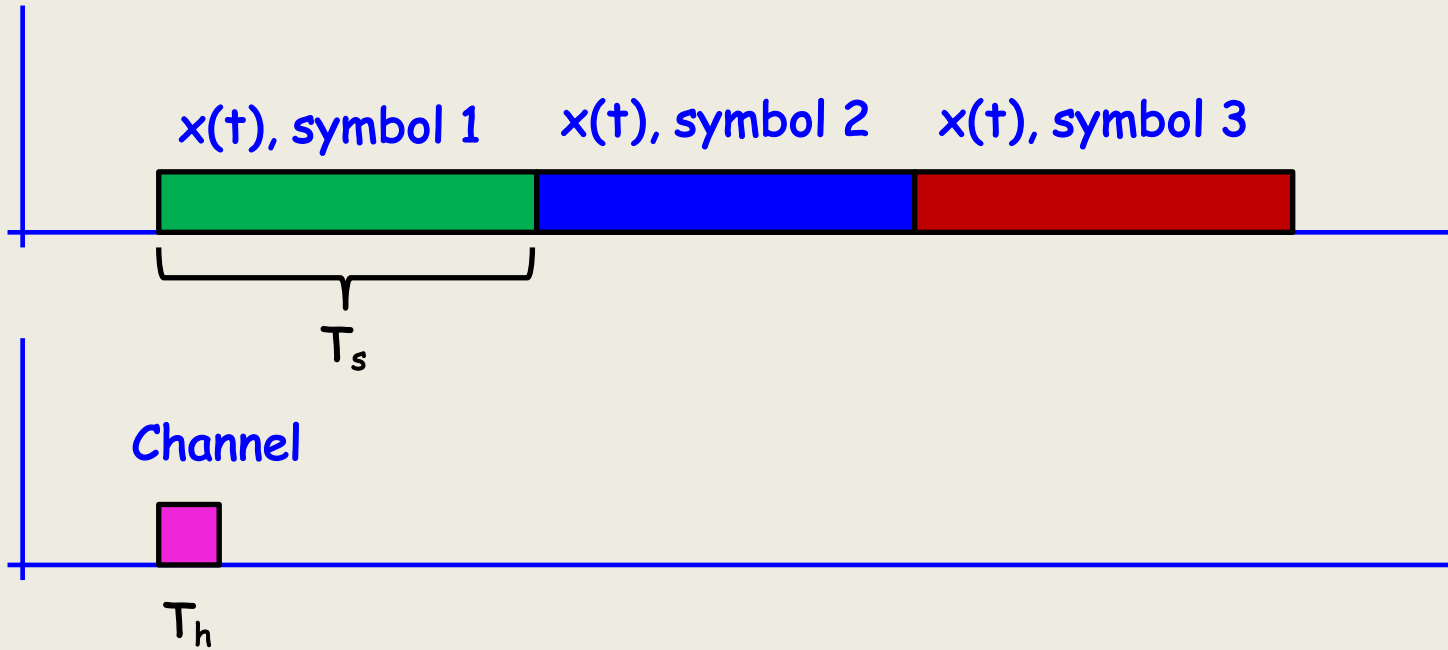
Left to do

1. Define signal in $-T_h < t < 0$ - The cyclic prefix
2. D/A conversion
3. Modulation to band-pass
4. Receiver

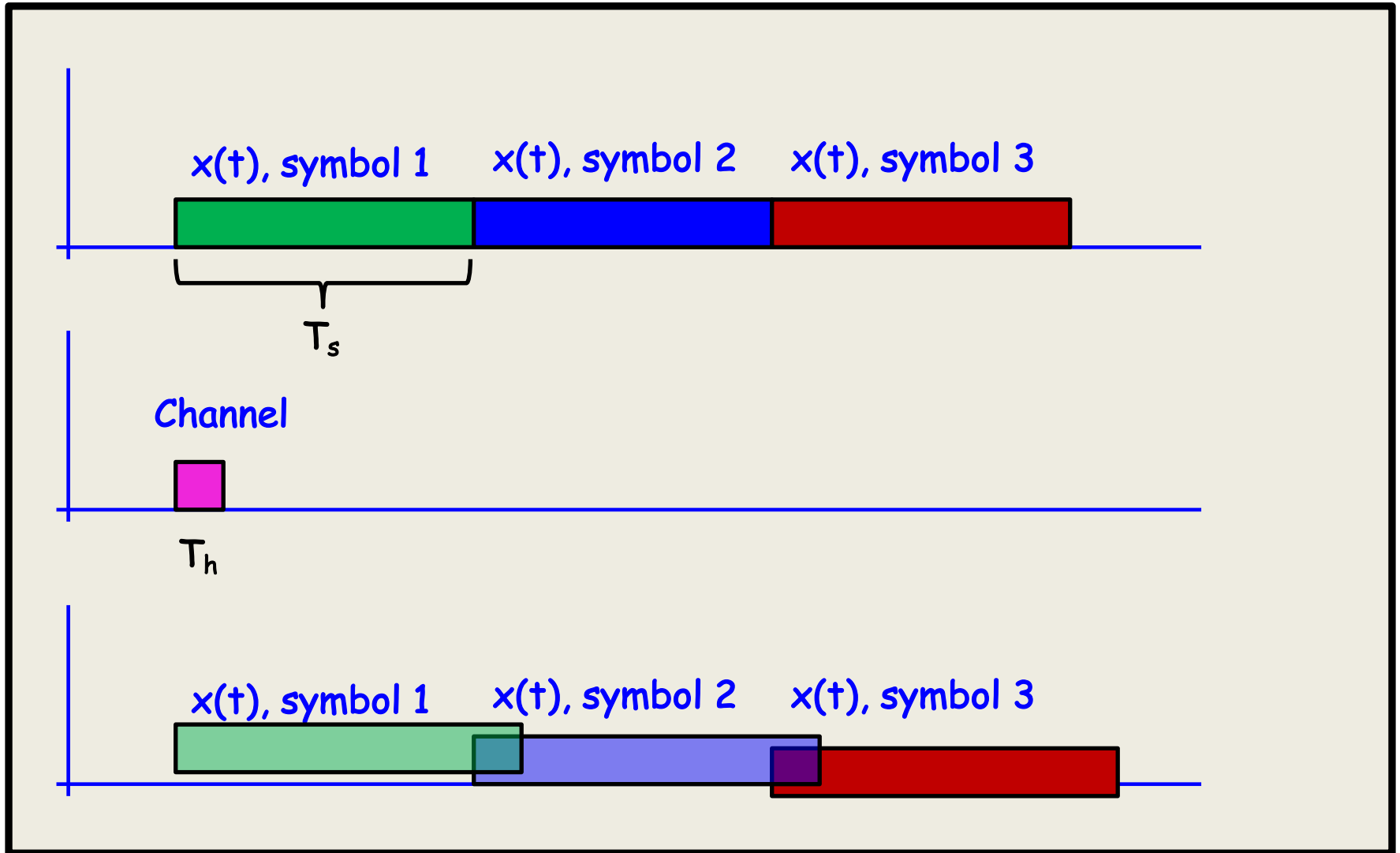
Lecture 7: OFDM



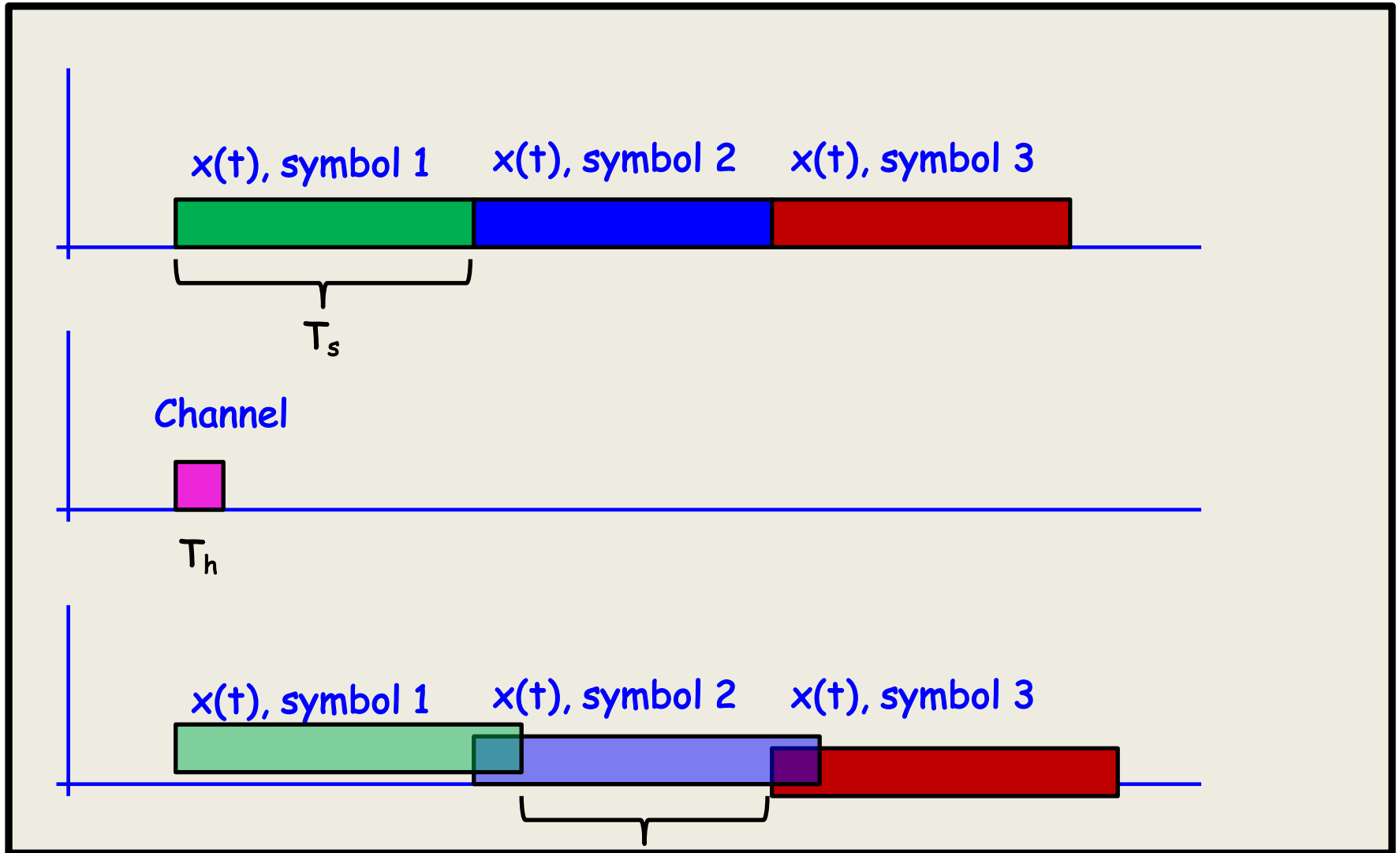
Lecture 7: OFDM



Lecture 7: OFDM

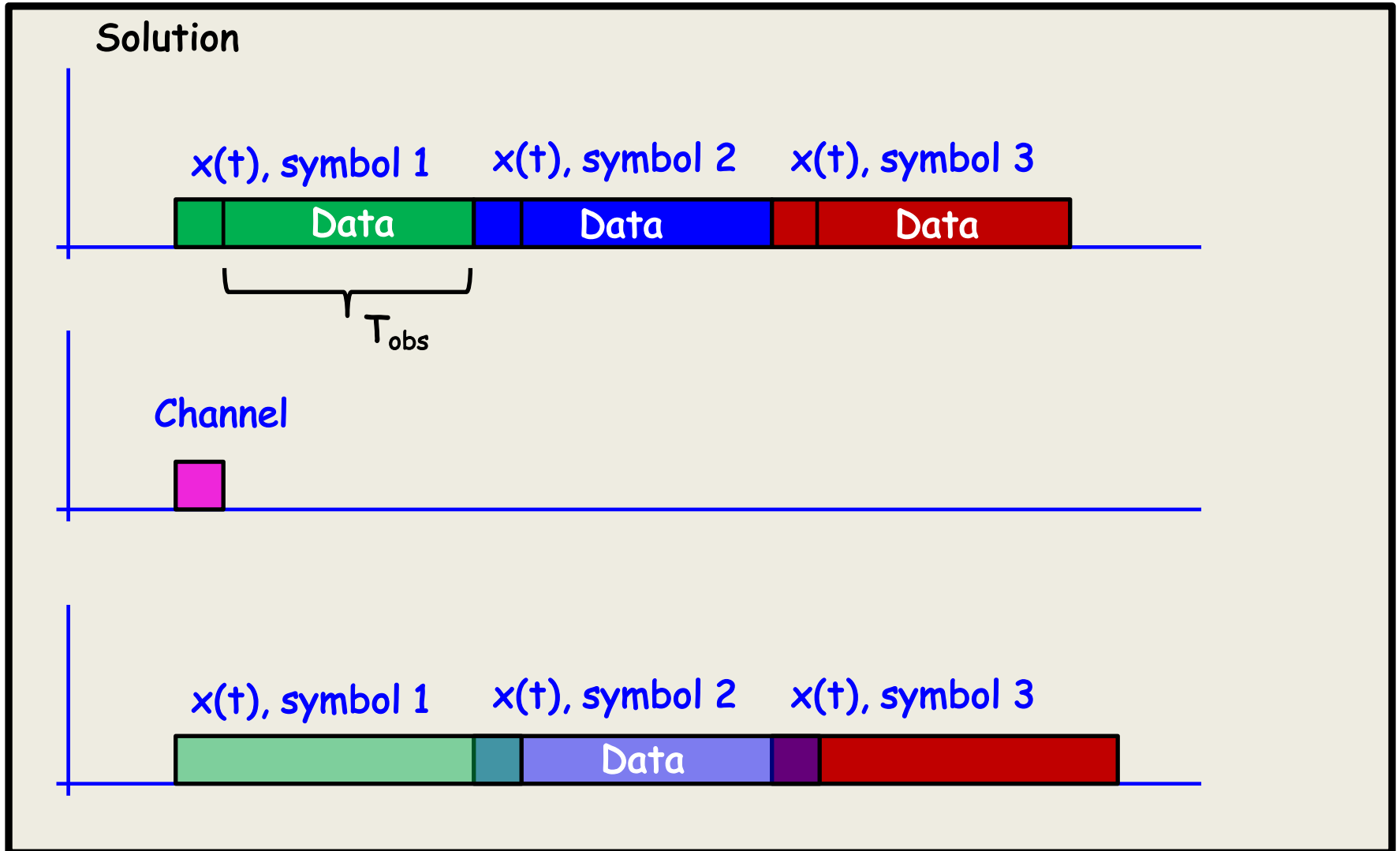


Lecture 7: OFDM



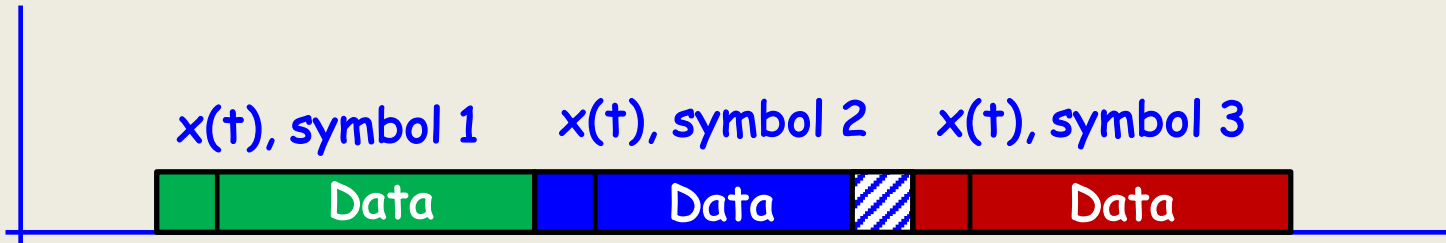
Useful part. T_{obs}

Lecture 7: OFDM



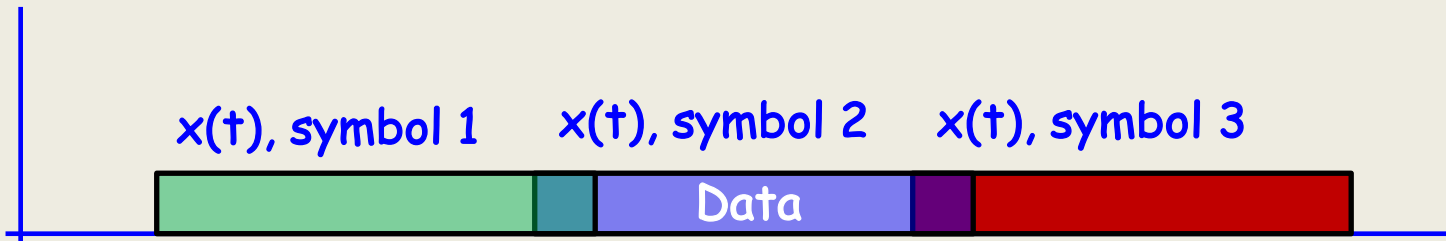
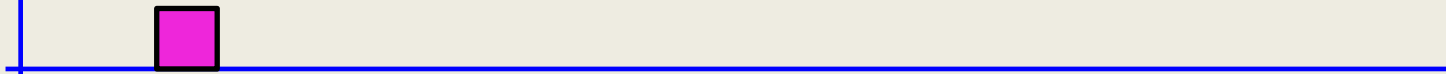
Lecture 7: OFDM

Solution



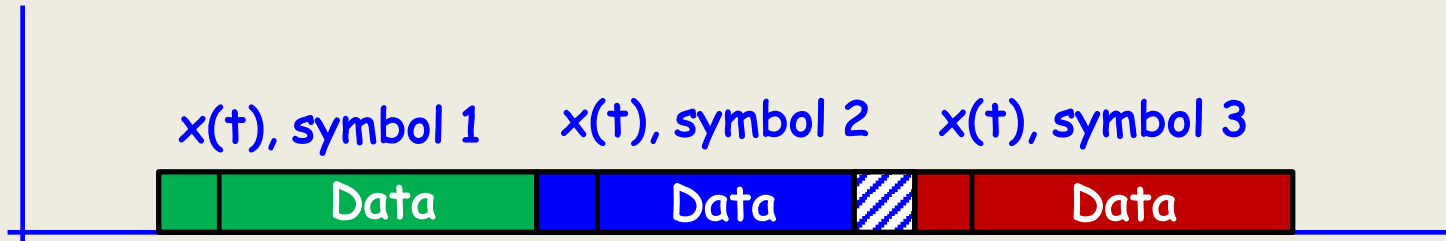
Whatever is sent here

Channel



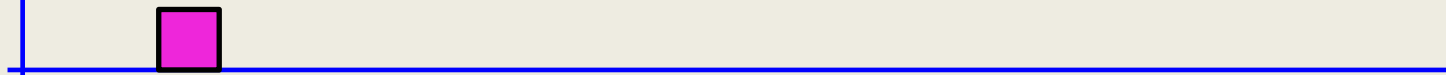
Lecture 7: OFDM

Solution

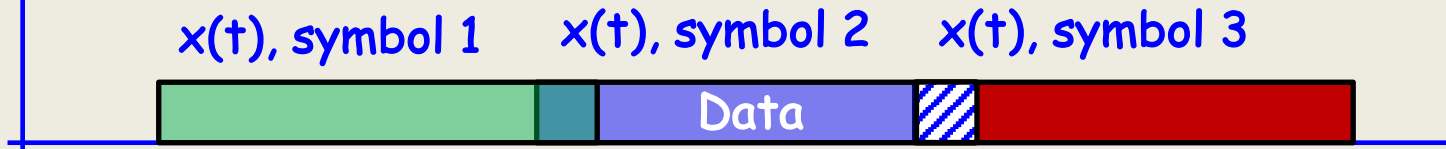


Whatever is sent here

Channel

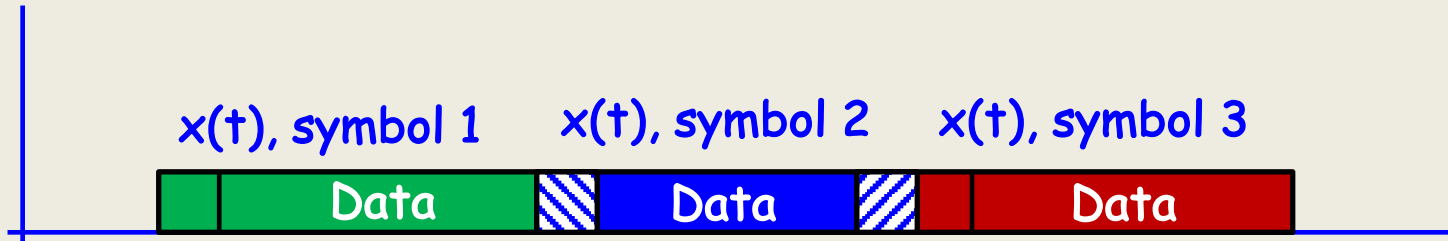


Is partly ending up here,
and therefore lost



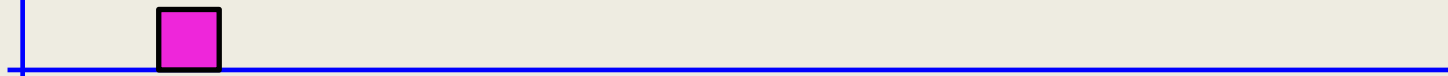
Lecture 7: OFDM

Solution

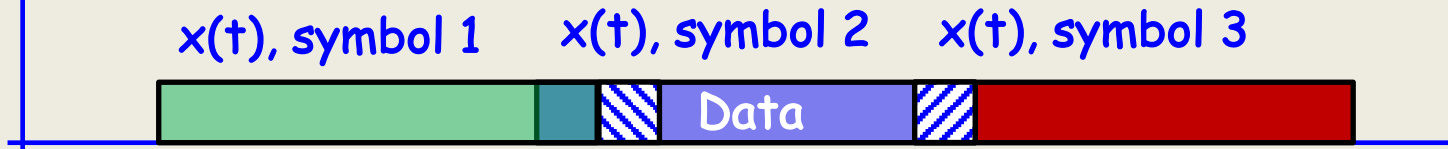


However, whatever is sent here

Channel

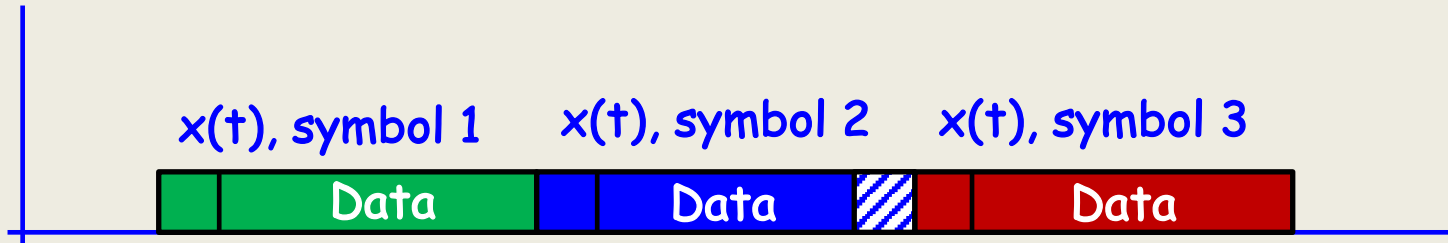


Is partly ending up here,
and therefore useful



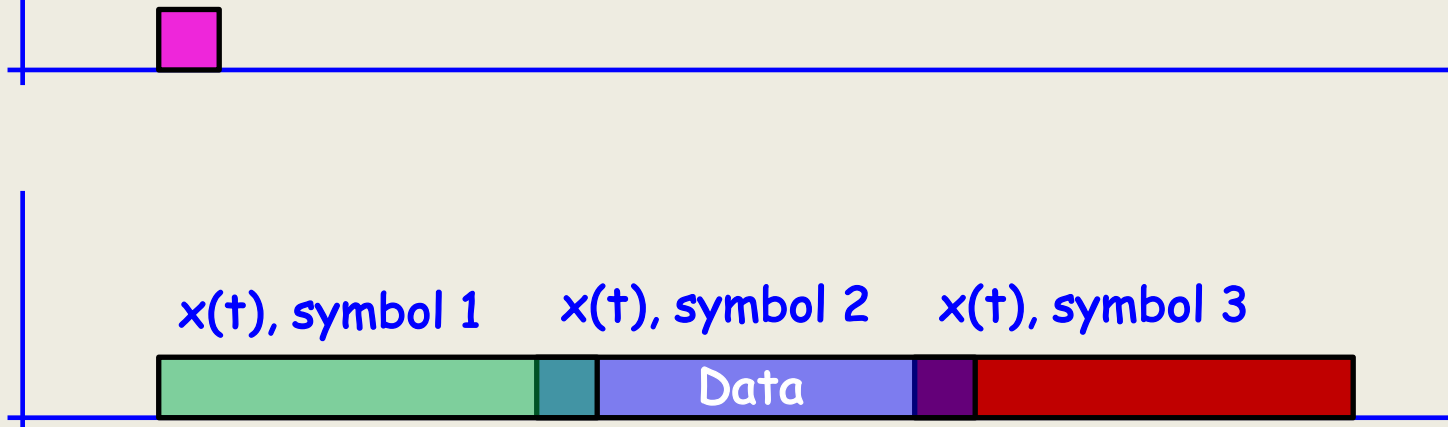
Lecture 7: OFDM

Solution



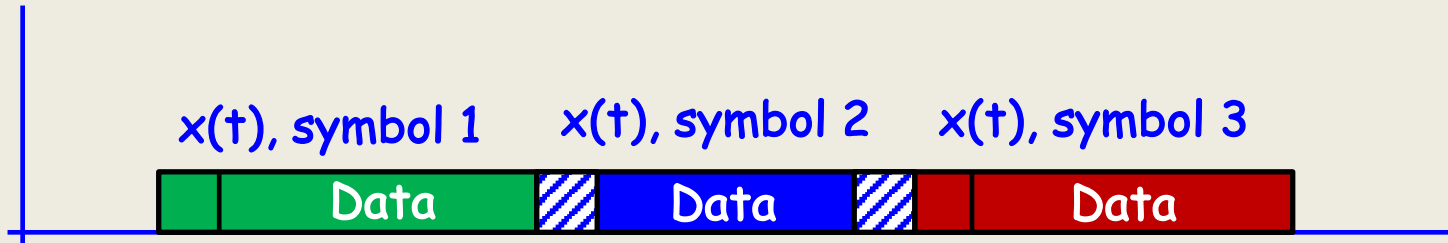
To regain, this

Channel



Lecture 7: OFDM

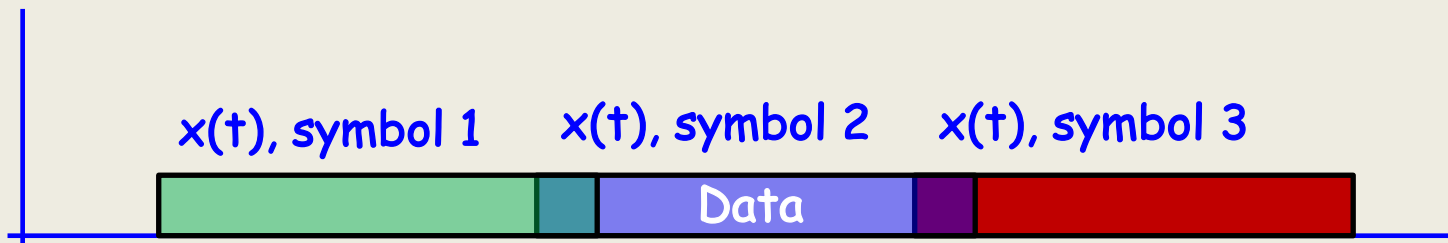
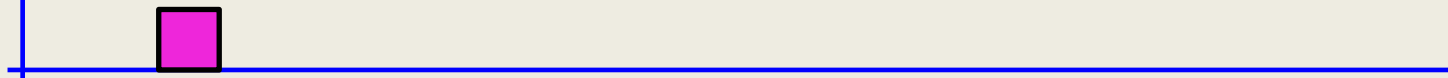
Solution



Put it here

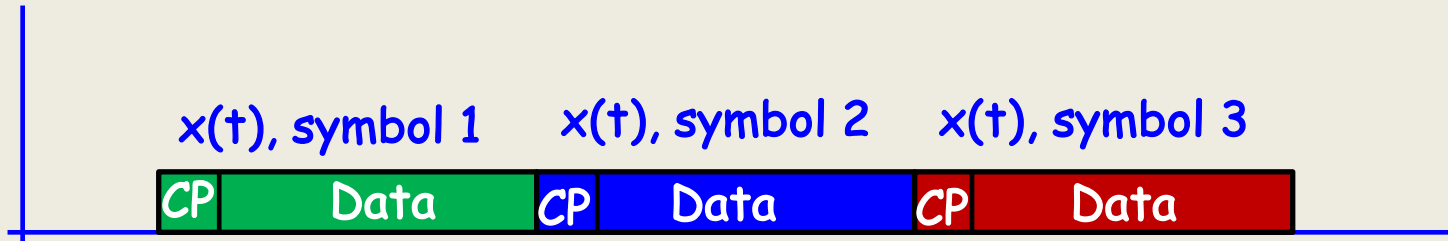
To regain, this

Channel



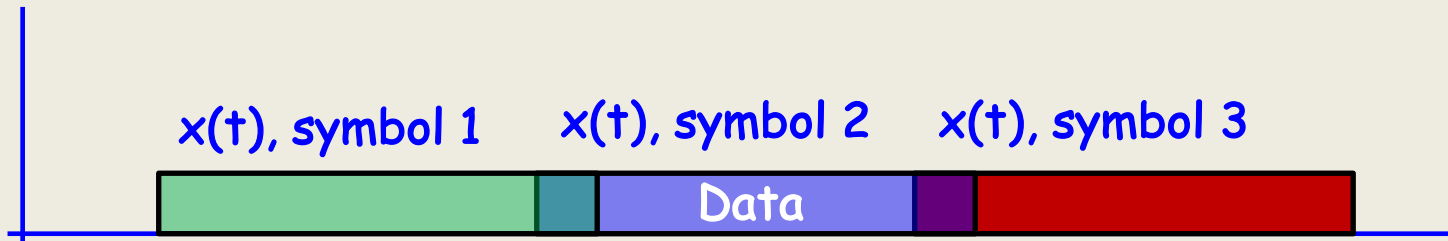
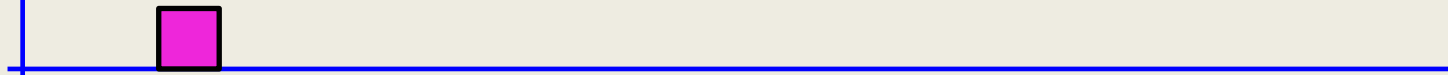
Lecture 7: OFDM

Solution

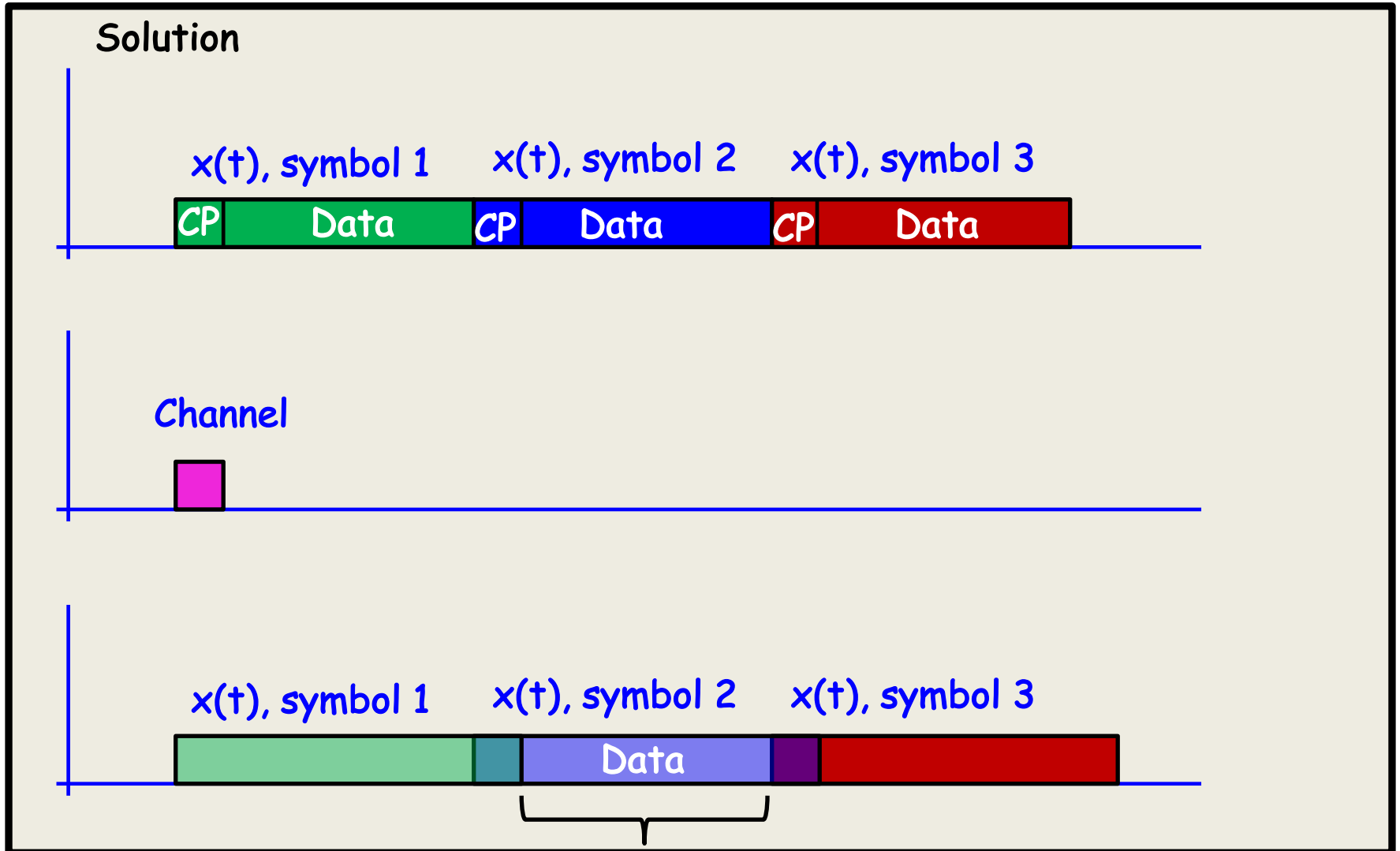


Called cyclic prefix (CP)

Channel

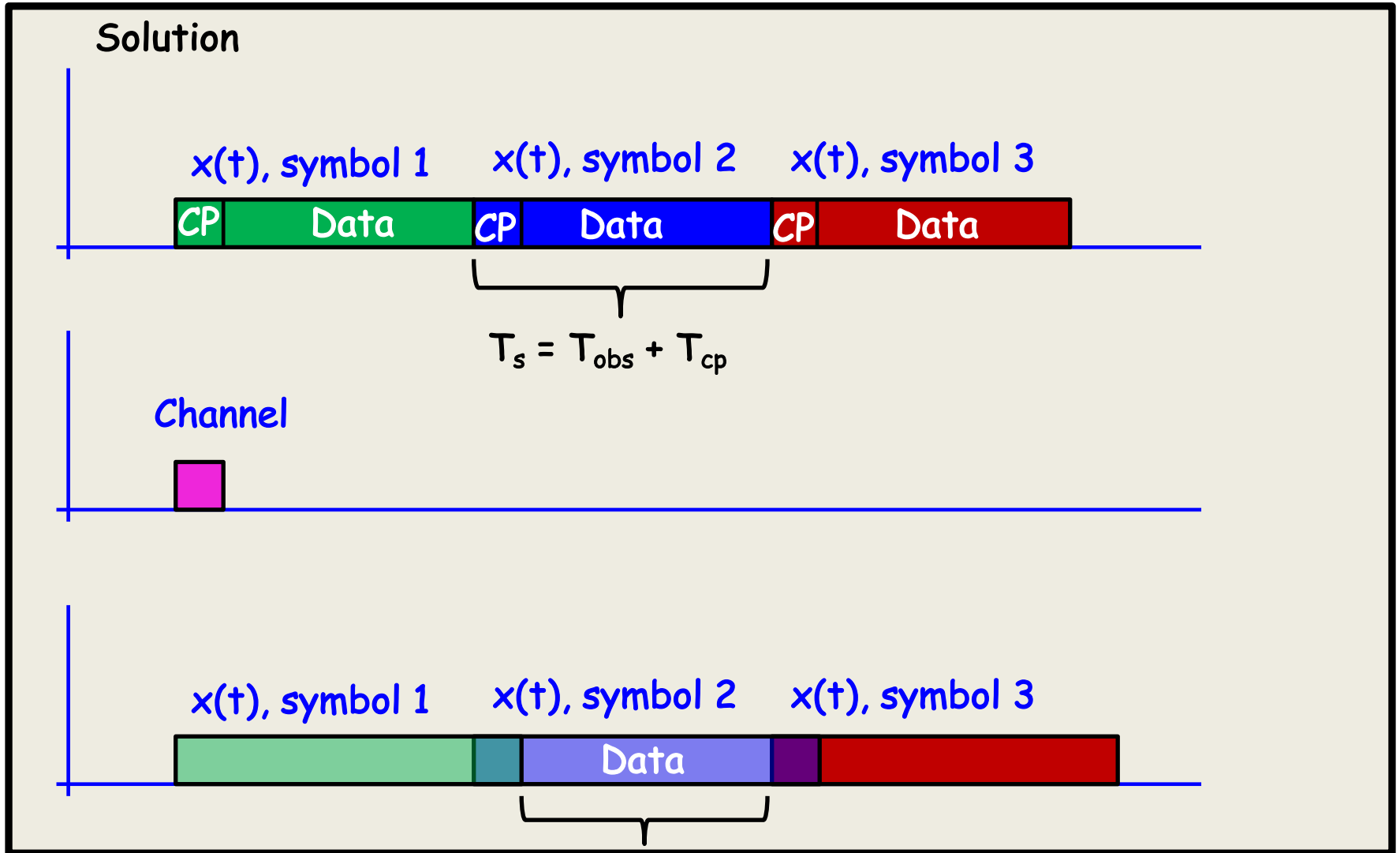


Lecture 7: OFDM



Useful part, should be length T_{obs}

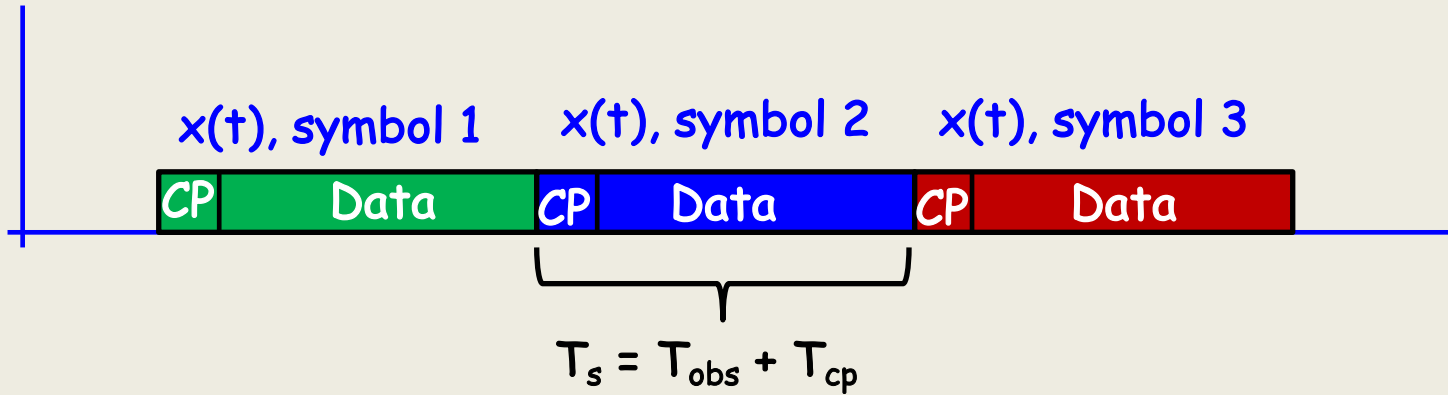
Lecture 7: OFDM



Useful part, should be length T_{obs}

Lecture 7: OFDM

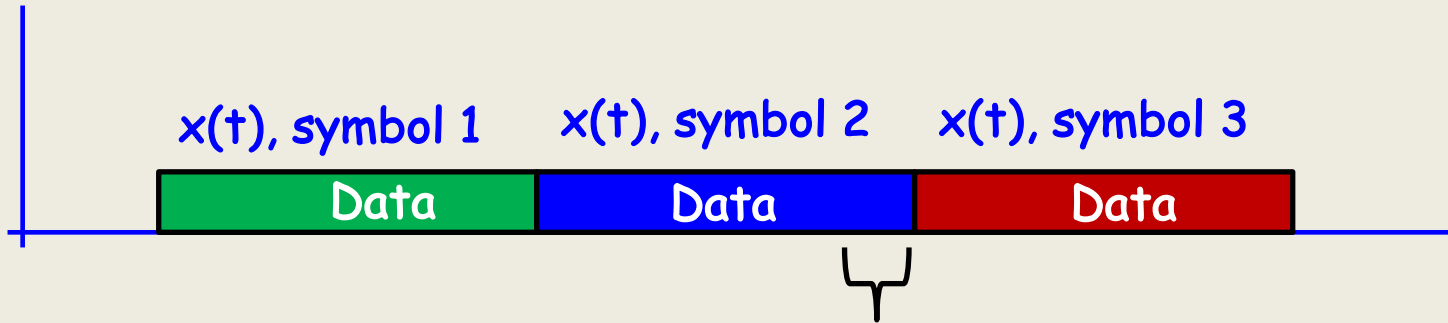
Solution



Summary:

Lecture 7: OFDM

Solution

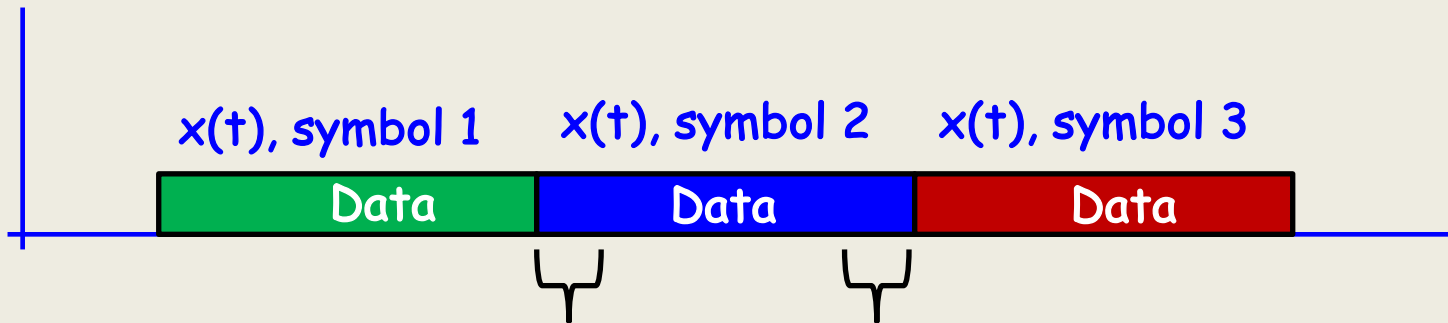


Summary:

1. Marked region is partly lost since it interferes with next block

Lecture 7: OFDM

Solution

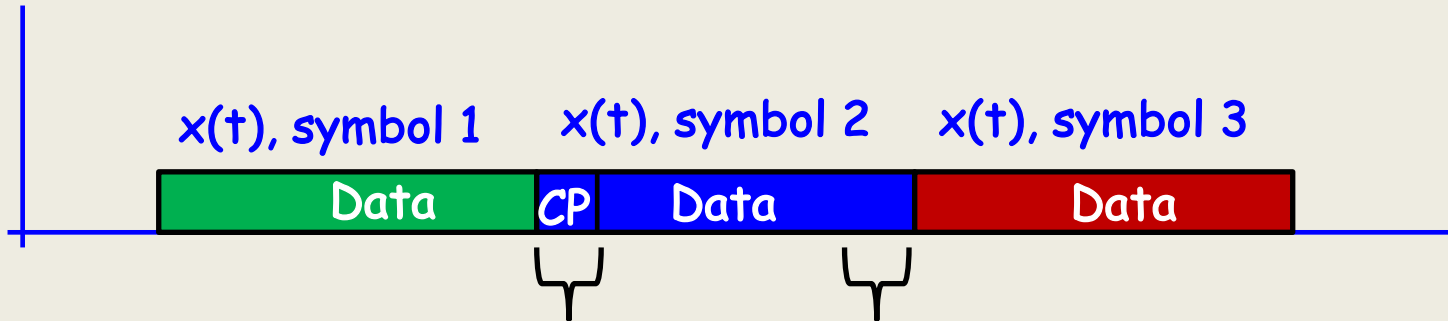


Summary:

1. Marked region is partly lost since it interferes with next block
2. First part is partly lost, since previous block interferes with it

Lecture 7: OFDM

Solution

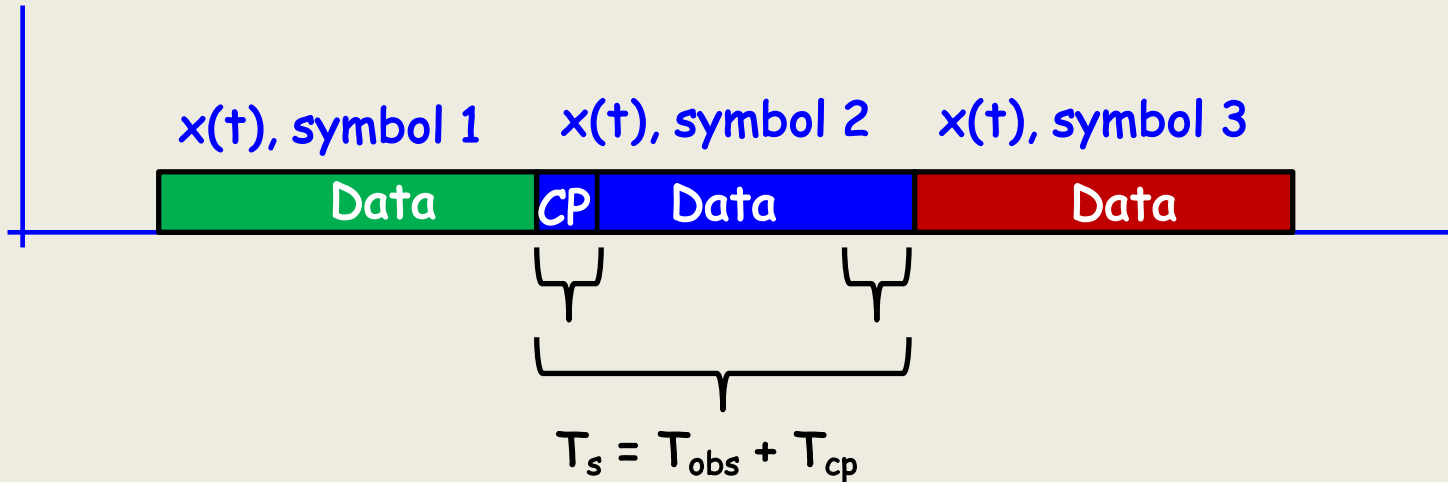


Summary:

1. Marked region is partly lost since it interferes with next block
2. First part is partly lost, since previous block interferes with it
3. Put the last part in the first part, call it CP

Lecture 7: OFDM

Solution

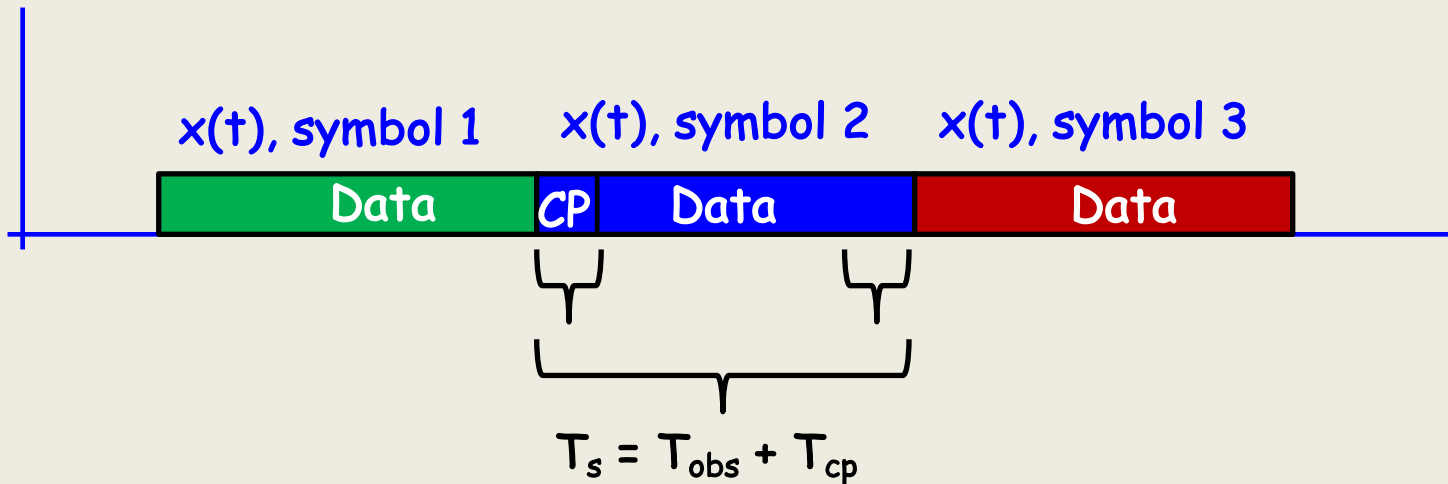


Summary:

1. Marked region is partly lost since it interferes with next block
2. First part is partly lost, since previous block interferes with it
3. Put the last part in the first part, call it CP
4. Length of data block should be $T_{obs}=1/f_{\Delta}$

Lecture 7: OFDM

Solution

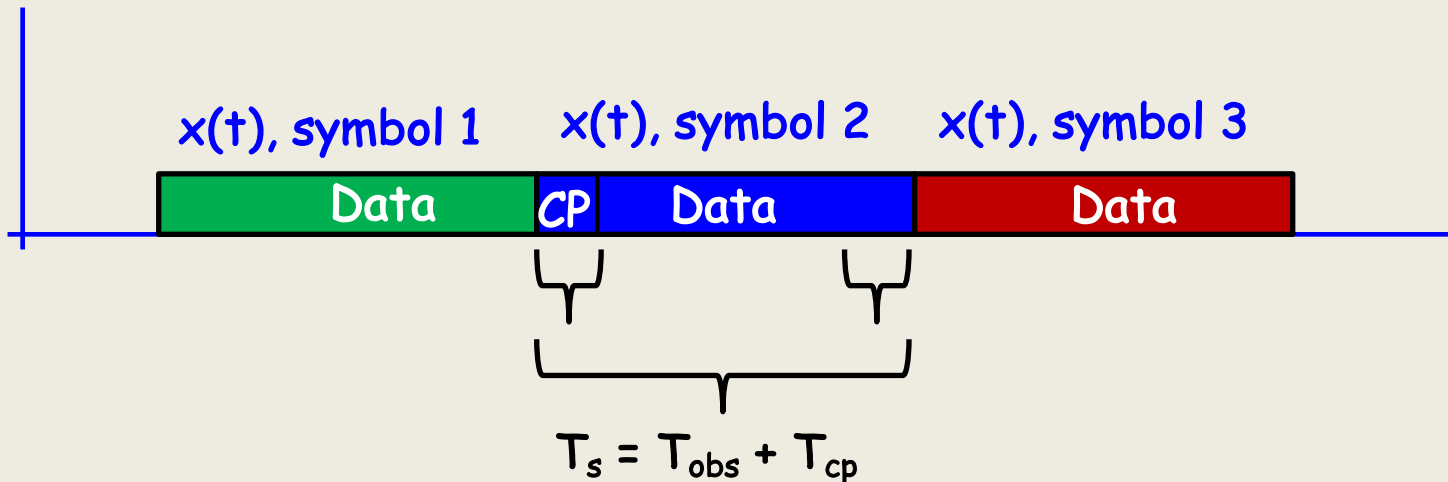


Summary:

1. Marked region is partly lost since it interferes with next block
2. First part is partly lost, since previous block interferes with it
3. Put the last part in the first part, call it CP
4. Length of data block should be $T_{obs}=1/f_{\Delta}$
5. No proof, yet, why this should be good, but it is plausible, since the lost part is present in the beginning

Lecture 7: OFDM

Solution

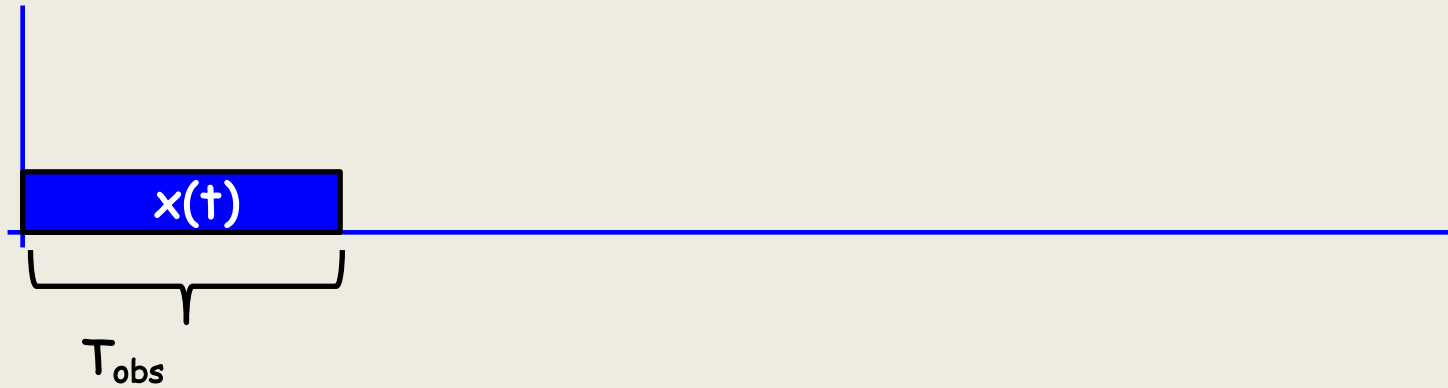


Summary:

1. Marked region is partly lost since it interferes with next block
2. First part is partly lost, since previous block interferes with it
3. Put the last part in the first part, call it CP
4. Length of data block should be $T_{obs}=1/f_{\Delta}$
5. No proof, yet, why this should be good, but it is plausible, since the lost part is present in the beginning
6. Spectral efficiency loss

Lecture 7: OFDM

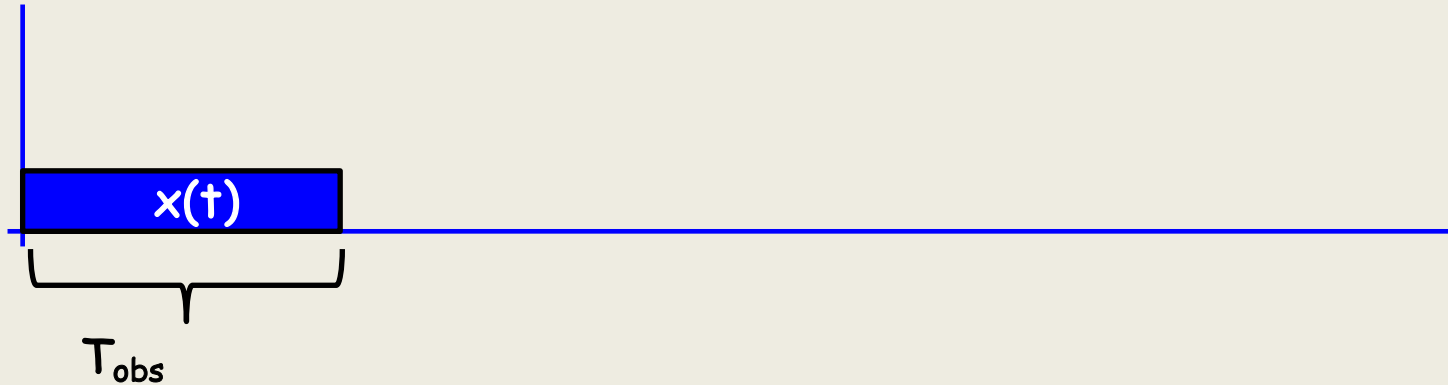
Formulas



$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta} t), \quad 0 \leq t \leq T_{obs}$$

Lecture 7: OFDM

Formulas

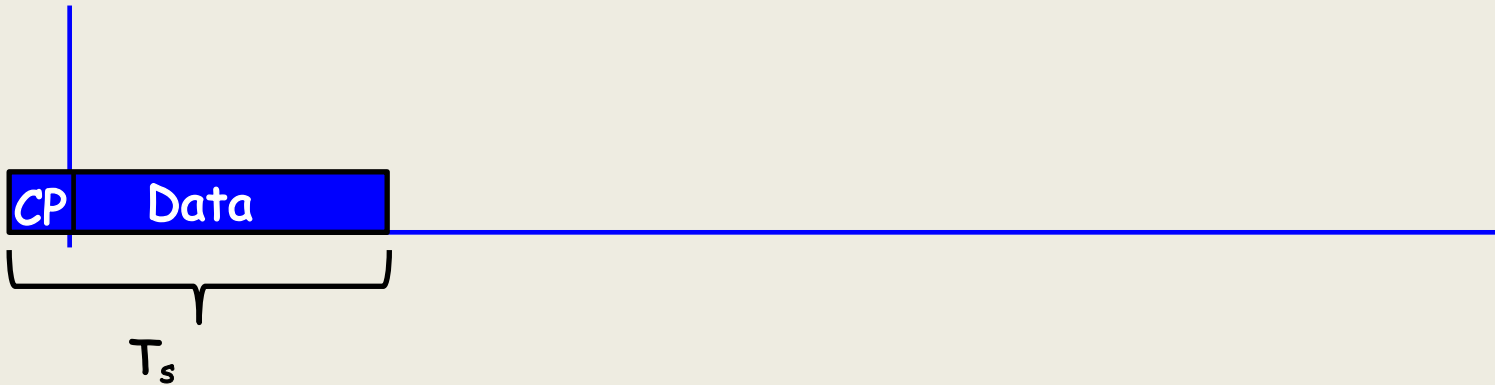


$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta} t)$$

↑
Periodic with period $T_{\text{obs}} = 1/f_{\Delta}$

Lecture 7: OFDM

Formulas



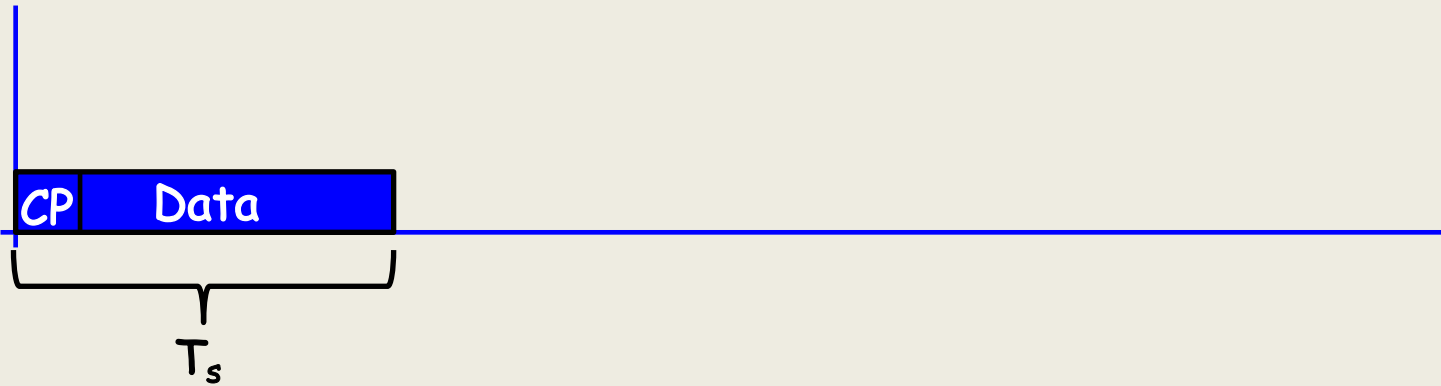
$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta} t), \quad -T_{cp} \leq t \leq T_{obs}$$



Has properties that we want: first part (left of 0) equals last part

Lecture 7: OFDM

Formulas



$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \leq t \leq T_s$$

↑
Make causal

Lecture 7: OFDM

Altogether,

1. Take a block of K data symbols $\{a_k\}$
2. Select a sampling rate, by choosing $N \geq K$
3. Find N values $\{X_m\}$ according to the box below
4. Compute $\{x_n\}$ using an IDFT
5. Add last L samples to the beginning

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

$$X_m = Na_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

Lecture 7: OFDM

Left to do

1. Define signal in $-T_h < t < 0$ - The cyclic prefix
2. D/A conversion
3. Modulation to band-pass
4. Receiver

Explanation of 1
comes in 4



Warning: numbering of steps is not the same as in the compendium

Lecture 7: OFDM

OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \leq t \leq T_s$$

Lecture 7: OFDM

$$\text{OFDM signal} = \text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$$

$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \leq t \leq T_s$$

$$u_n = \text{IFFT}(\mathbf{X}) \quad \text{Temporary variable}$$

$$X_m = N a_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = N a_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

$$\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots \ u_{N-1} \ u_0 \ \dots \ u_{N-1}]$$

$\underbrace{\hspace{10em}}_{\text{CP}}$

Lecture 7: OFDM

$$\text{OFDM signal} = \text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$$

$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \leq t \leq T_s$$

$$u_n = \text{IFFT}(\mathbf{X})$$

$$X_m = N a_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = N a_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

$$\mathbf{x} = [\underbrace{u_{N-L} \ u_{N-L+1} \ \dots \ u_{N-1}}_{\text{CP}} \ u_0 \ \dots \ u_{N-1}]$$

D/A

Lecture 7: OFDM

OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rct}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{\text{samp}})$$

$$u_n = \text{IFFT}(\mathbf{X})$$

$$X_m = N a_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = N a_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

$$\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots \ u_{N-1} \ u_0 \ \dots \ u_{N-1}]$$

$\underbrace{\hspace{10em}}_{\text{CP}}$

D/A

Lecture 7: OFDM

OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{\text{samp}})$$

$$g_{\text{ideal}}(t) = \frac{\sin(\pi f_{\text{samp}}t)}{\pi f_{\text{samp}}t}$$

$$u_n = \text{IFFT}(\mathbf{X})$$

$$X_m = N a_{m-g_0}, \quad 0 \leq m \leq g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \leq m \leq g_0 + N - 1$$

$$X_m = N a_{m-(g_0+N)}, \quad g_0 + N \leq m \leq N - 1$$

$$\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots \ u_{N-1} \ u_0 \ \dots \ u_{N-1}]$$

$\underbrace{\hspace{10em}}_{\text{CP}}$

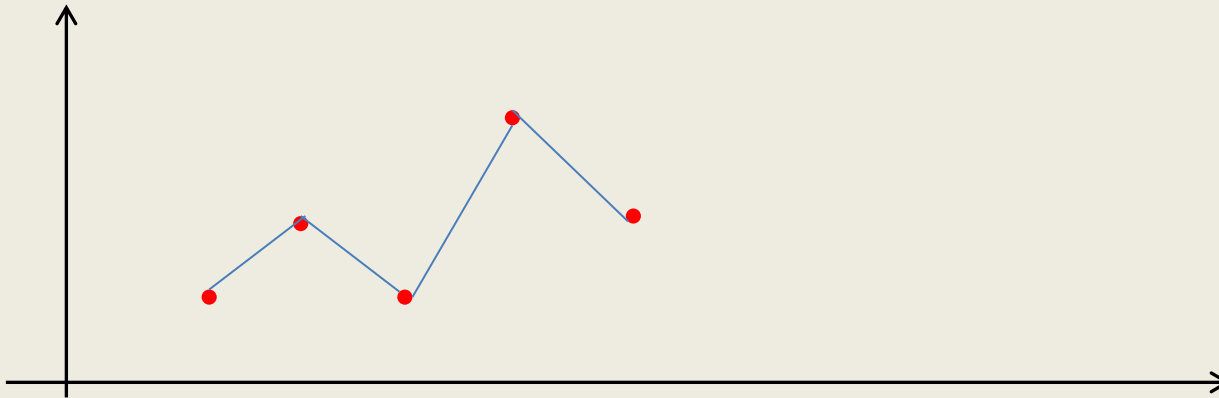
Lecture 7: OFDM

OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{\text{samp}})$$

$$g_{\text{ideal}}(t) = \frac{\sin(\pi f_{\text{samp}}t)}{\pi f_{\text{samp}}t}$$

Alternative (suboptimal)



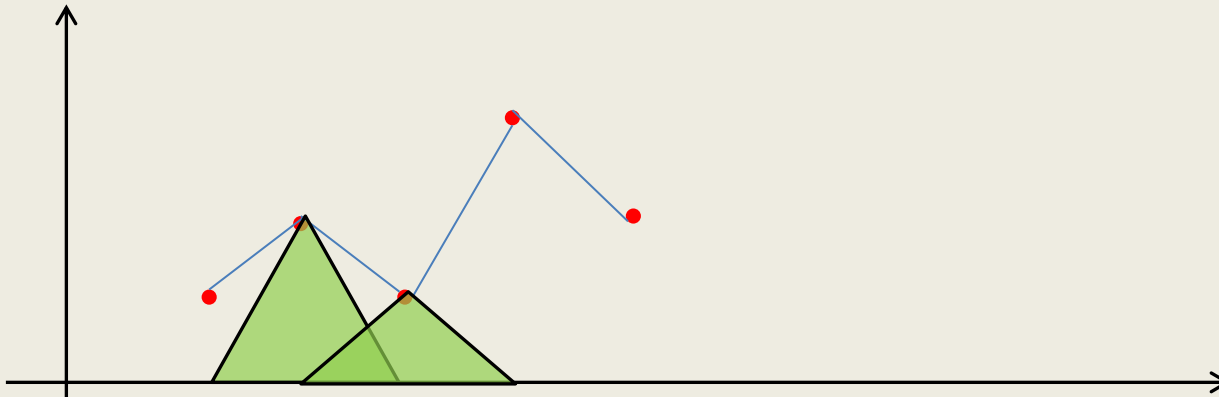
Lecture 7: OFDM

OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{\text{samp}})$$

$$g_{\text{ideal}}(t) = \frac{\sin(\pi f_{\text{samp}}t)}{\pi f_{\text{samp}}t}$$

Alternative (suboptimal)



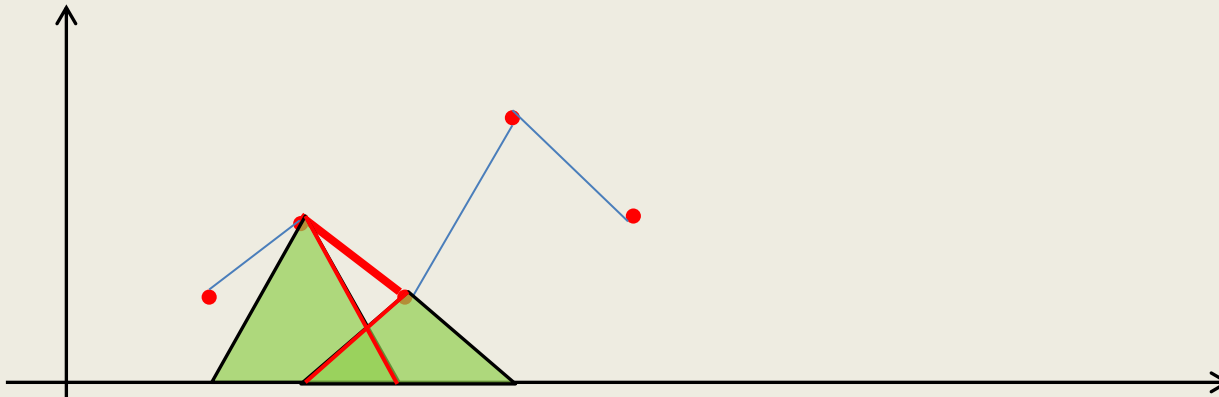
Lecture 7: OFDM

OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{\text{samp}})$$

$$g_{\text{ideal}}(t) = \frac{\sin(\pi f_{\text{samp}}t)}{\pi f_{\text{samp}}t}$$

Alternative (suboptimal)



Lecture 7: OFDM

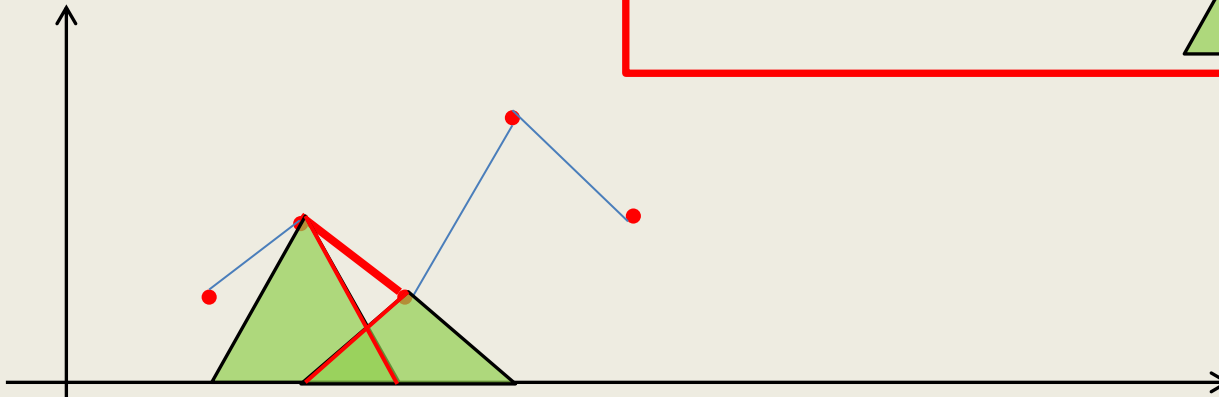
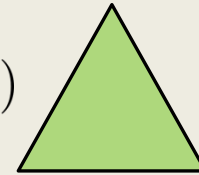
OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rct}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{\text{samp}})$$

$$g_{\text{ideal}}(t) = \frac{\sin(\pi f_{\text{samp}}t)}{\pi f_{\text{samp}}t}$$

Alternative (suboptimal)

$$g_{\text{simple}}(t) = \text{tri}(t)$$



Lecture 7: OFDM

Left to do

1. Define signal in $-T_h < t < 0$ - The cyclic prefix
2. D/A conversion
3. Modulation to band-pass
4. Receiver

Explanation of 1
comes in 4



Warning: numbering of steps is not the same as in the compendium

Lecture 7: OFDM

Left to do

1. Define signal in $-T_h < t < 0$ - The cyclic prefix

2. D/A conversion

3. **Modulation to band-pass**

Very simple
OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

4. Receiver

Warning: numbering of steps is not the same as in the compendium

Lecture 7: OFDM

Left to do

1. Define signal in $-T_h < t < 0$ - The cyclic prefix

2. D/A conversion

3. Modulation to band-pass OFDM signal = $\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$

4. Receiver Next lecture

Warning: numbering of steps is not the same as in the compendium