

1

general WF rule : More power to better channels

$\Rightarrow R_2(f)$ not possible

High SNR : All channels get same power
(since $\frac{N_0}{|H(f)|^2} \approx 0$)

$\Rightarrow R_3(f) \leftrightarrow N_{o,1} = \text{very low}$

Low SNR : Best channels get all power
(since $\frac{N_0}{|H(f)|^2} \gg 0$)

$\Rightarrow R_4(f) \leftrightarrow N_{o,3} = \text{very high}$

$\Rightarrow R_1(f) \leftrightarrow N_{o,2} = \text{medium}$

2

i) Question : $2W \log\left(\frac{P}{2N_0}\right) \stackrel{?}{=} W \log\left(\frac{2P}{WN_0}\right)$

Numerical check \rightarrow No

ii/ Yes N signals can always be represented in at most N dimensions.

iii) Yes. Let $y_k(t) = \alpha_k \phi_k(t) \cdot s_{e,k} + n_k(t)$
 and $\ell=0, 1 \quad k=1 \dots N$

In signal space:

$$y_k = \alpha_k s_{e,k} + n_k$$

equal power allocation:

$$s_{0,k} = \frac{\sqrt{E}}{N} \quad \forall k$$

$$s_{1,k} = -\frac{\sqrt{E}}{N} \quad \forall k$$

$$\begin{aligned} D_{0,1}^2 &= \sum_{k=1}^N \left| \alpha_k [s_{0,k} - s_{1,k}] \right|^2 \\ &= \sum_{k=1}^N \alpha_k^2 \cdot 4 \frac{E}{N} = 4 E \frac{1}{N} \sum_{k=1}^N \alpha_k^2 \end{aligned}$$

best channel only

assume $|\alpha_1| > |\alpha_k| \quad k \geq 2$

$$s_{0,1} = \sqrt{E} \quad s_{1,1} = -\sqrt{E}$$

$$s_{j,k} = 0 \quad k \geq 2$$

$$D_{0,1}^2 = 4 |\alpha_1|^2 E \quad |\alpha_1|^2 > \frac{1}{N} \sum_{k=1}^N \alpha_k^2 \quad \downarrow$$

(since α_i^2 is largest

and $\frac{1}{N} \sum_{k=1}^N \alpha_k^2$ is
an average
value)

iv/ No M-PSK was hard
and required Numerical integration

v/ No

$$R_b = \frac{k \cdot \log_2(m)}{T_{obs} + T_{cp}}$$

$$BW = k f_D = \frac{k}{T_{obs}} \rightarrow$$

$$T_{obs} = \frac{k}{BW} \rightarrow$$

$$R_b = \frac{k \log_2(m)}{\frac{k}{BW} + T_{cp}}$$

$$P = \frac{R_b}{BW} = \frac{k \log_2(m)}{\left(\frac{k}{BW} + T_{cp}\right) BW} =$$

$$= \frac{\log_2(M)}{1 + \frac{T_{cp-BW}}{K}}$$

P not linearly prop. to K
i.e. $P \neq K \cdot \text{something}$

vii) No. channel must be constant during one OFDM symbol.

CP protects against multipath spread.

3

$$R_b = \frac{R \log_2(M)}{T_S} = 10^7 R \cdot \log_2(M)$$

a) $R=1, R_b = 6 \cdot 10^7 \rightarrow M=64$

b) $P_s \approx Q\left(\sqrt{d_{min}^2 \frac{E_b}{N_0}}\right) \quad P_b \approx \frac{P_s}{6}$ *very loose*

64-QAM $\rightarrow d_{min}^2 = \frac{2}{7}$

Make approximation at $P_e \approx 10^{-3}$

$$Q\left(\sqrt{\frac{2}{3} \frac{E_b}{N_0}}\right) = 10^{-3} \quad \frac{E_b}{N_0} = \dots \text{ [check table]}$$

c/ $R_b = 3 \cdot 10^7 \quad M=64 \rightarrow R=\frac{1}{2}$
 $R_b = 1 \cdot 10^7 \quad M=64 \rightarrow R=\frac{1}{6}$

d/ optimal code \rightarrow capacity applies

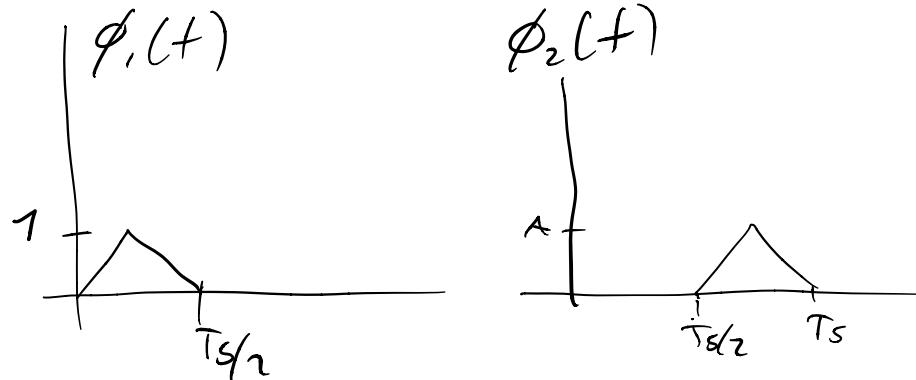
$$\frac{E_b}{N_0} > \frac{2^P - 1}{P}$$

$$\text{Assume BW} = \frac{1}{T_s} = 10^2 \text{ Hz}$$

$$R_b = 3 \cdot 10^7 \quad P = \frac{R_b}{BW} = 3$$

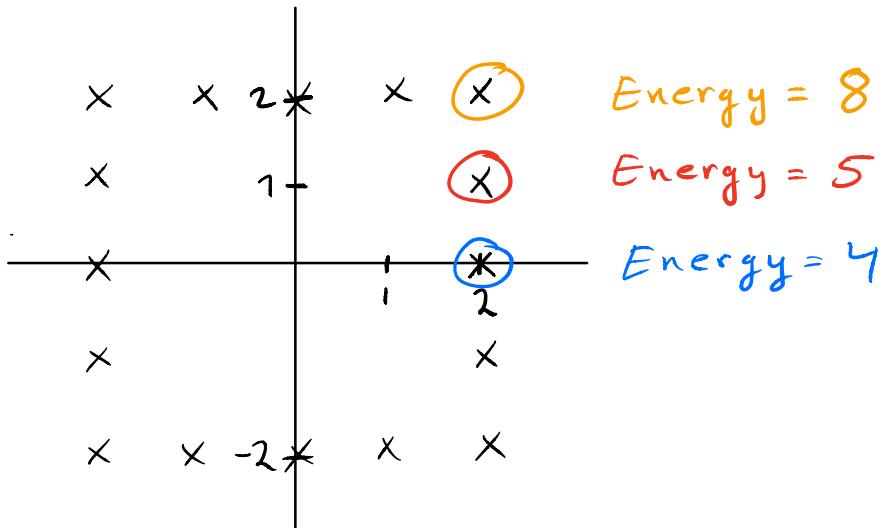
$$\boxed{\frac{E_b}{N_0} > \frac{2^3}{3}}$$

4



$$\int \phi_i^2(t) dt = \dots = \frac{A T_s^2}{2} = 1$$

a)



$$b/ E_S = \frac{1}{16} [4 \cdot 8 + 8 \cdot 5 + 4 \cdot 4]$$

$$= \frac{1}{16} [32 + 40 + 16] = \frac{1}{16} [88]$$

$$= \frac{22}{4}$$

$$E_b = \frac{E_S}{4} = \frac{22}{16}$$

c/ $D_{\min}^2 = 1$ (from figure)

$$d_{\min}^2 = \frac{D_{\min}^2}{2E_b} = \frac{1}{44} = \frac{16}{11} = \frac{4}{11}$$

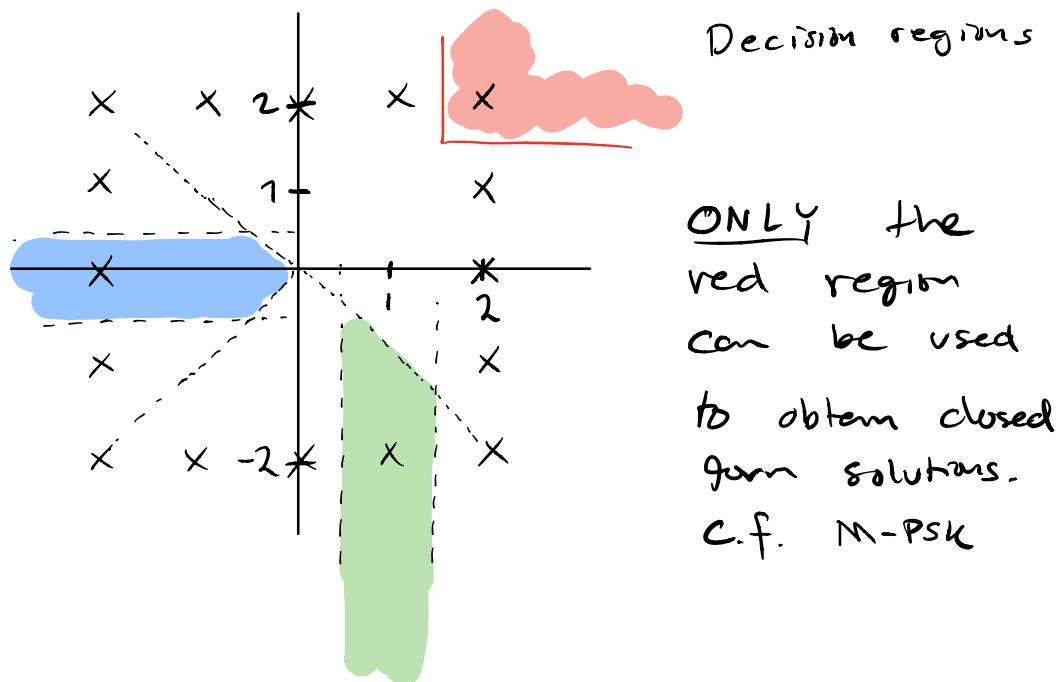
$$P_e \approx Q\left(\sqrt{\frac{4}{11} \frac{E_b}{N_0}}\right)$$

d/ 16-PSK (see table) $d_{\min}^2 = 0.6090$

$$10 \log_{10}\left(\frac{0.6090}{0.3636}\right) = 2.24$$

16 PSK	2.24 dB better
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e/



5

$$E_b = \frac{P \cdot T_{obs}}{\log_2(M) \cdot K}$$

← Tot energy in
 1 ofDM symbol
 ← # bits in one
 symbol
 $M=4$

$$= \frac{T_{obs}}{4000}$$

$$P_c \approx \frac{1}{2 d^2 m \frac{E_b}{N_0}} = \frac{1000}{T_{obs}} \cdot N_0$$

\uparrow
 $=2$

$$\frac{1000 \cdot N_0}{T_{obs}} = \frac{10^{-9}}{T_{obs}} < 10^{-5} \Rightarrow$$

$$T_{obs} > 10^{-4} \Rightarrow$$

$$\frac{1}{f_\Delta} > 10^{-4} \Rightarrow$$

$f_\Delta < 10^4$

Also $T_{obs} < t_{con} \Rightarrow$

$$\frac{1}{f_\Delta} < t_{con} \Rightarrow f_\Delta > \frac{1}{t_{con}} = 100$$

$$\Rightarrow 100 < f_\Delta < 10000 \quad (\text{Hz})$$