

$$\textcircled{1} \quad a/ \quad \int_0^T |s(t) - \hat{s}(t)|^2 dt = \int_0^T \left| s(t) - \sum_{k=1}^K s_k f_k(t) \right|^2 dt$$

$$= \underbrace{\int s^2(t) dt}_E + \int \left| \sum s_k f_k(t) \right|^2 dt - 2 \int s(t) \sum s_k f_k(t) dt$$

$$= E + \sum_{k=1}^K s_k^2 - 2 \sum s_k \underbrace{\int s(t) f_k(t) dt}_{\tilde{s}_k}$$

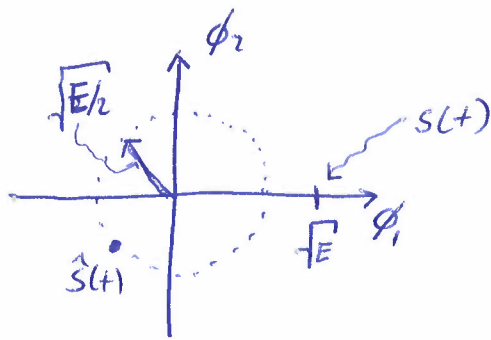
$$= \underbrace{E + \sum_{k=1}^K s_k^2 - 2 \sum_{k=1}^K s_k \tilde{s}_k}$$

$$\frac{\partial}{\partial s_k} [\cdot] = 2s_k - 2\tilde{s}_k = 0$$

$$s_k = \tilde{s}_k = \int_0^T s(t) f_k(t) dt$$

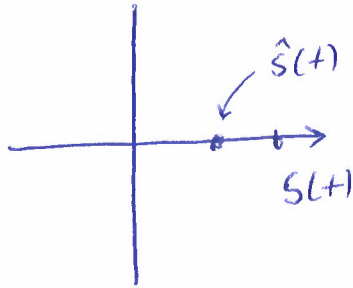
b/ • Let $s(t) = \sqrt{E} \phi_1(t)$ for some $\phi_1(t)$
with $\int \phi_1^2(t) dt = 1$

• Let $\hat{s}(t) = s_1 \phi_1(t) + s_2 \phi_2(t)$, $s_1^2 + s_2^2 = E/2$



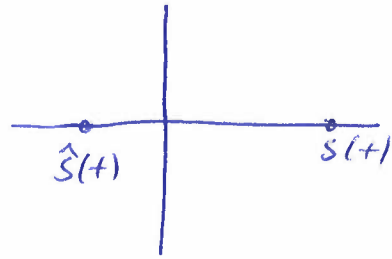
$\hat{S}(t)$ is on circle

$|S(t) - \hat{S}(t)|^2$ minimized



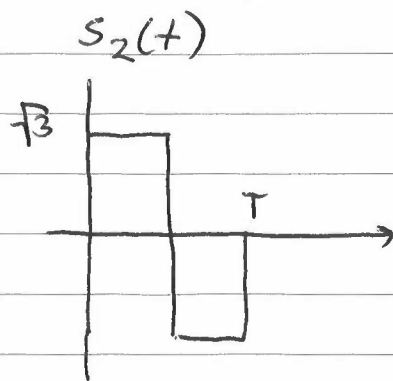
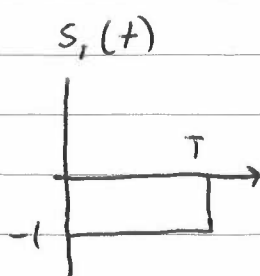
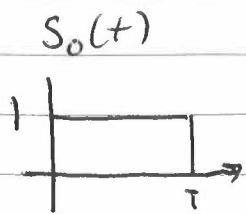
$$|S(t) - \hat{S}(t)|^2 = \dots = (\sqrt{E} - \sqrt{E/2})^2 \approx 0.1E$$

$|S(t) - \hat{S}(t)|^2$ maximized



$$\dots (\sqrt{E} + \sqrt{E/2})^2 \approx 3E$$

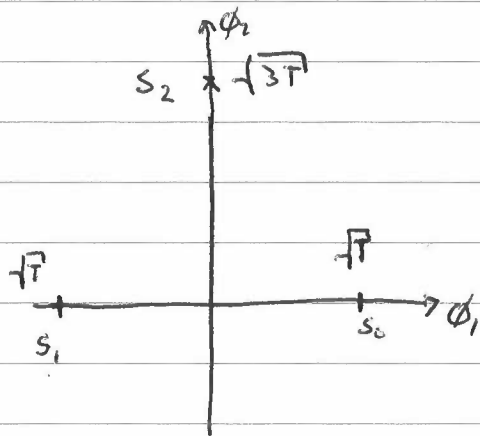
$$c/ \int_0^T |S(t) - \sum s_k f_k(t)|^2 dt = \underbrace{\int s_k^2(t) dt + \sum_{k=1}^K s_k^2 - 2 \int s(t) \sum s_k f_k(t)}_0$$



$$E_0 = T = E_1 \quad E_2 = 3T$$

since $\int s_0(t) s_2(t) dt = 0 \Rightarrow$

$$\phi_1(t) = \frac{s_0(t)}{\sqrt{T}} \quad \phi_2(t) = \frac{s_2(t)}{\sqrt{3T}}$$



$$\text{Let } D_{i,j}^2 = \int |s_i(t) - s_j(t)|^2 dt$$

$$\tilde{D}_{i,j}^2 = \int |F_i(t) - f_j(t)|^2 dt$$

$$E_i = \int s_i^2(t) dt \quad \tilde{E}_i = \int F_i^2(t) dt$$

It follows that $\tilde{D}_{i,j}^2 = D_{i,j}^2$ wherefore

error performance is not affected.

$$\begin{aligned} \tilde{E}_i &= \int |s_i(t) - c(t)|^2 dt = \int s_i^2(t) dt + \int c^2(t) dt - 2 \int s_i(t)c(t) dt \\ &= E_i + \int c^2(t) dt - 2 \int s_i(t)c(t) dt. \end{aligned}$$

transmit Energy is affected.

Reasonable goal: find $c(t)$ that results in least average transmit energy, i.e.,

$$\min_{c(t)} \sum_{i=1}^3 P_i \tilde{E}_i \quad \text{where } P_i \text{ is the probability of message } i$$

$$\text{Let } c(t) = c_1 \phi_1(t) + c_2 \phi_2(t) + c_3 \phi_3(t)$$

$\uparrow$
 $c(t)$ possibly not
 in same space

$$\tilde{E}_1 = T + c_1^2 + c_2^2 + c_3^2 - 2\sqrt{T}c_1$$

$$\tilde{E}_2 = T + c_1^2 + c_2^2 + c_3^2 + 2\sqrt{T}c_1$$

$$\tilde{E}_3 = 3T + c_1^2 + c_2^2 + c_3^2 - 2\sqrt{3}\sqrt{T}c_2$$

$$\text{Av. energy} = \frac{1}{3}(\tilde{E}_1 + \tilde{E}_2 + \tilde{E}_3) = \frac{5T}{3} + c_1^2 + c_2^2 + c_3^2 - 2\sqrt{\frac{T}{3}}c_2$$

clearly, minimal for $c_1 = c_3 = 0$

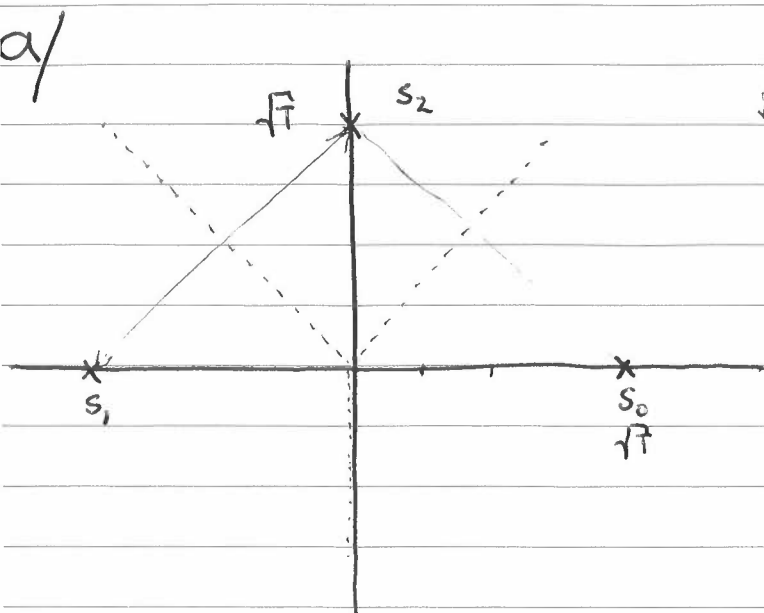
$$\frac{\partial}{\partial c_2} = 2c_2 - 2\sqrt{\frac{T}{3}} = 0 \rightarrow c_2 = \sqrt{\frac{T}{3}}$$

$$\text{plug back} \rightarrow \text{average power} = \frac{5T}{3} + \frac{T}{3} - 2\frac{T}{3} = \frac{4T}{3}$$

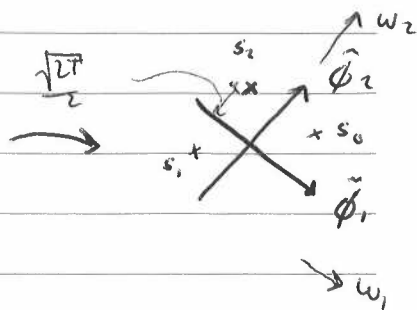
$$\text{average power for } c(t)=0 = \frac{5T}{3}$$

$$\text{gain (saving)} \quad \frac{4}{5}$$

Yes since we used $P_i = \frac{1}{3} \forall i$



$$D_{2,1}^2 = D_{2,0}^2 = 2T$$



w_1, w_2 noise
in new basis

$$P(\text{error}|s_2) = 1 - P(\text{correct}|s_2) = 1 - P\left(w_2 > -\sqrt{\frac{T}{2}} \text{ and } w_1 < \sqrt{\frac{T}{2}}\right)$$

$$= 1 - P\left(w_2 > -\sqrt{\frac{T}{2}}\right) P\left(w_1 < \sqrt{\frac{T}{2}}\right)$$

$$= 1 - (1 - P(w_2 > \sqrt{\frac{T}{2}})) (1 - P(w_1 > \sqrt{\frac{T}{2}}))$$

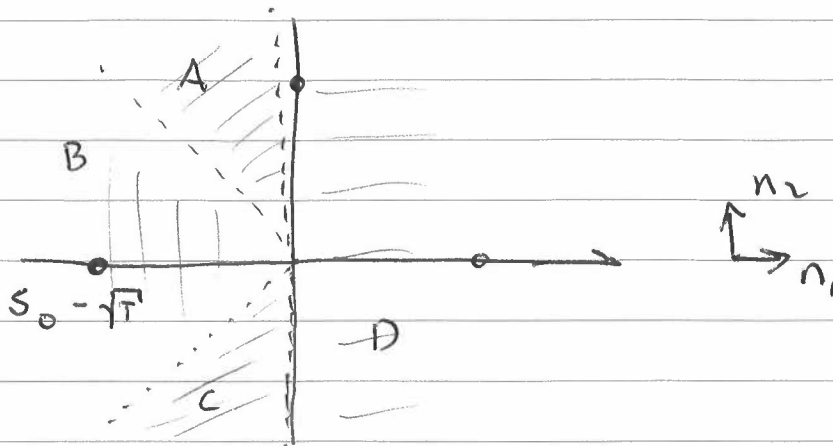
$$P(w_2 > \sqrt{\frac{T}{2}}) = P(w_1 > \sqrt{\frac{T}{2}}) = P(x > \sqrt{\frac{T}{2}} / \sqrt{\frac{T}{N_0}})$$

$$x \sim N(0, 1)$$

$$= Q\left(\sqrt{\frac{T}{N_0}}\right)$$

$$\Rightarrow P(\text{error}|s_2) = 1 - (1 - Q(\cdot))(1 - Q(\cdot))$$

$$= 1 - [1 - 2Q(\cdot) + Q^2(\cdot)] = \boxed{2Q\left(\sqrt{\frac{T}{N_0}}\right) - Q^2\left(\sqrt{\frac{T}{N_0}}\right)}$$



Consider the four regions A-D

Decision region for s_0 is BUC

↑ union

↓ received signal

$$\cancel{P(r \in B)} \quad P(r \in B) = P(\text{correct} | s_2)$$

↑ symmetry

$$\cancel{P(r \in A \cup C)} \quad P(r \in A \cup C) = P(r \in A \cup B \cup C) - P(r \in B)$$

$$= P(n_1 < \sqrt{T}) - P(\text{correct} | s_2)$$

$$P(r \in C) = \frac{1}{2} P(r \in A \cup C)$$

↑ symmetry

$$P(r \in B \cup C) = P(r \in B) + P(r \in C)$$

$$= \underbrace{\frac{1}{2} \Pr(\text{correct} | s_2) + \frac{1}{2} \Pr(n_1 < \sqrt{T})}_{P(\text{correct} | s_0)}$$

7)

$$C = R_s \log_2 \left(1 + \frac{P}{N_0 R_s} \right)$$

$$R_s = R_b / 3 \quad \begin{array}{l} \text{(3 bit per symbol} \\ \text{due to the 3/4 code)} \end{array}$$

$$= \frac{10^4}{3}$$

→

$$C = \frac{10^4}{3} \log_2 \left(1 + \frac{10^4}{10^4/3} \right) = \frac{10^4}{3} \log_2(4) = \frac{2}{3} 10^4 < R_b$$

breaks capacity

can be done via Water filling, but
also using " $W \log_2(1 + \frac{P}{N_0 W})$ "

Let P be power in $[-1000, 1000]$

$\rightarrow$ $4-P$ power in $\pm [1000, 2000]$

Rx power

$$\frac{P |H(f)|^2}{(4-P) |H(f)|^2} = P \cdot \frac{1000 N_0}{2} \quad (4-P) \cdot \frac{500 N_0}{2}$$

$$C = 1000 \log_2 \left(1 + \frac{P \cdot 1000 \frac{N_0}{2}}{1000 N_0} \right) + 1000 \log_2 \left(1 + \frac{(4-P) \frac{500 N_0}{2}}{1000 N_0} \right)$$

$$= 1000 \left[\log_2 \left(1 + \frac{P}{2} \right) + \log_2 \left(1 + \frac{4-P}{4} \right) \right]$$

$$= 1000 \log_2 \left(\underbrace{\left(1 + \frac{P}{2} \right) \left(2 - \frac{P}{4} \right)}_{f(P)} \right)$$

$$\frac{\partial f(P)}{\partial P} = 0 \rightarrow P=3$$

$$C|_{P=3} = 1000 \left(\log_2 \left(\frac{5}{2} \right) + \log_2 \left(\frac{5}{4} \right) \right) < 2000$$

breaks

$$T_{obs} = \frac{1}{f_{\Delta}} \quad T_s = \frac{1}{f_{\Delta}} + 5 \mu s$$

$$K = \frac{9 \cdot 10^6}{f_{\Delta}} \quad \swarrow 16\text{-QAM}$$

$$R_b = \frac{K \cdot 4}{T_s} = \frac{4 \cdot 9 \cdot 10^6}{f_{\Delta} \left[\frac{1}{f_{\Delta}} + 5 \cdot 10^{-6} \right]}$$

$$= \frac{36 \cdot 10^6}{1 + 5 f_{\Delta} \cdot 10^{-6}}$$

$$(i) \approx 3.47 \cdot 10^7 \text{ bits/s}$$

$$(ii) \approx 3.35 \cdot 10^7 \text{ bits/s}$$

b/

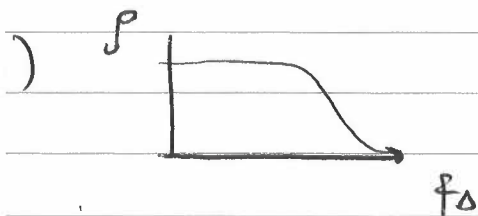
$$R_b = \frac{K \log_2(M)}{T_{obs} + T_{cp}}$$

$$W = f_{\Delta} K$$

$$\rho = \frac{R_b}{W} = \frac{K \log_2(M)}{K f_{\Delta} \left(\frac{1}{f_{\Delta}} + T_{cp} \right)} = \frac{\log_2(M)}{1 + f_{\Delta} T_{cp}}$$

$$i) \quad T_{cp} \nearrow \rightarrow \rho \searrow$$

$$(iii) \quad \text{No dep.}$$



$$(iv) \quad \text{No dep.}$$

C/ Yes. The channel must be constant over one OFDM symbol (T_s)

$$\text{At high SNR, } \text{BER} \sim \frac{1}{2d_{\min}^2 \frac{E_b}{N_0}} = \frac{1}{4 \frac{E_b}{N_0}}$$

But this holds only for R_s where we have slow fading and frequency-non-selectivity.

$$= 100 \ll R_s \ll f_{\text{coh}} = 100000$$

$$E_b R_b \rightarrow \frac{E_b}{N_0} = \frac{P}{R_b N_0} = \frac{10^7}{R_b} \gg 1 \text{ for all } R_b$$

$$\Rightarrow \frac{1}{4} R_b \frac{1}{10^7}$$

