Recall: If $x(t) = A\cos(\omega_c t) - B\sin(\omega_c t), \ 0 \le t \le T_s$

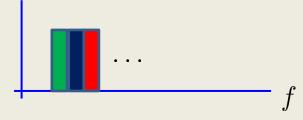
Then
$$z(t) = A_z \cos(\omega_c t) - B_z \sin(\omega_c t), \ T_h \le t \le T_s$$

$$A_z + iB_z = (A + iB)H(\omega_c)$$

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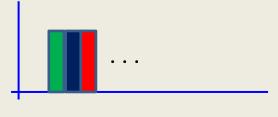


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We can transmit multiple signals at different sub-carriers: This is OFDM



We have

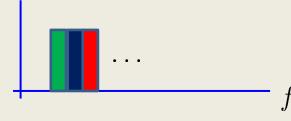
OFDM signal =
$$g_{rec}(t) \sum_{k=0}^{N-1} \text{Re}\{a_k \exp(i2\pi f_k t)\}$$

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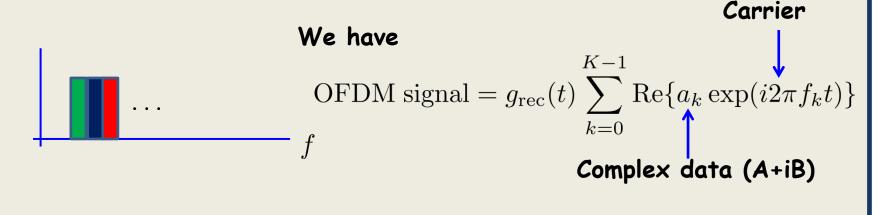
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Complex data (A+iB)

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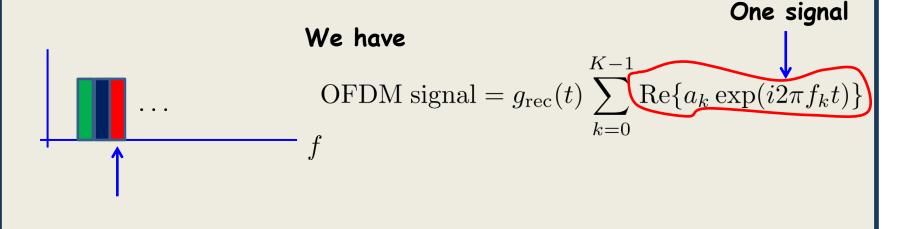
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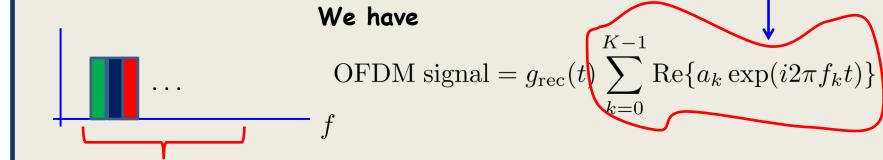


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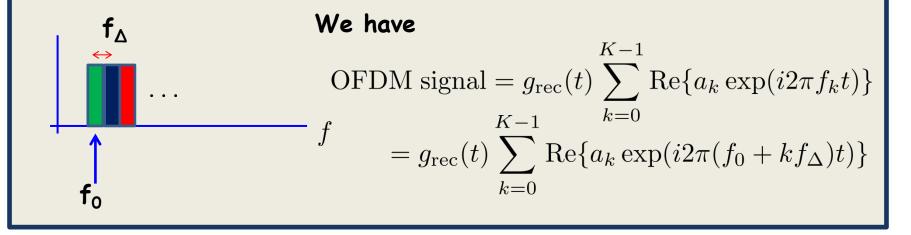
All signals



Recall: If $x(t) = A\cos(\omega_c t) - B\sin(\omega_c t), \ 0 \le t \le T_s$

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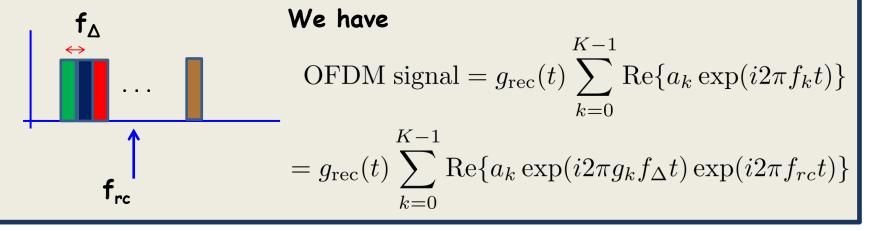
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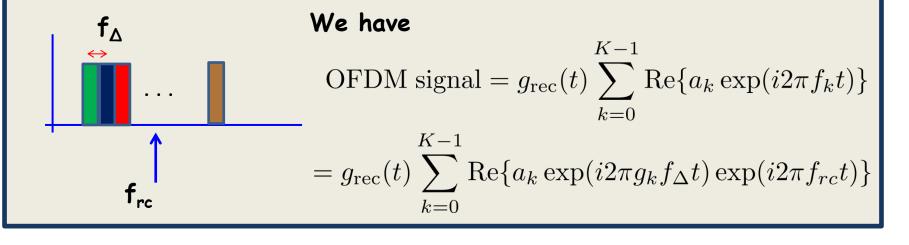
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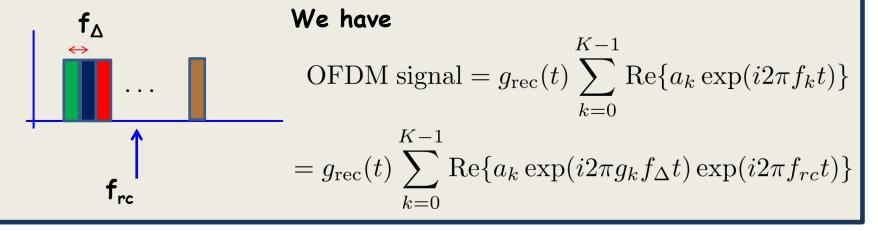
$$g_k = ...-2, -1, 0, 1, 2, ...$$

$$g_k$$
: $-\frac{K-1}{2} = g_0, \dots, -1, 0, 1, \dots, \frac{K-1}{2} = g_{K-1}$ if K is odd

$$g_k$$
: $-\frac{K-2}{2} = g_0, ..., -1, 0, 1, ..., \frac{K}{2} = g_{K-1}$ if K is even



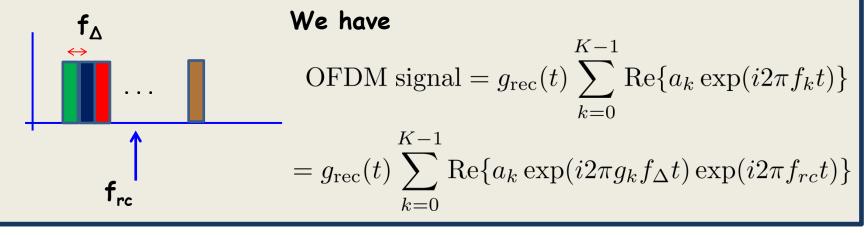
OFDM signal = Re
$$\{x(t) \exp(i2\pi f_{rc}t)\}$$
 $0 \le t \le T_{\text{obs}}$
$$x(t) = g_{\text{rec}}(t) \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}t)$$



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Important: $f_{\Delta}T_{\rm obs} > 1$



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To synthesize the signal, we

- 1. Sample the OFDM signal
- 2. Check how we can efficiently construct those samples
- 3. Perform D/A conversion

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How often do we need to sample?

Sampling theorem: Sample twice as fast the highest frequency component

Total bandwidth $W_{OFDM} \approx K f_{\Delta}$

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At baseband
$$W_{OFDM} pprox K f_{\Delta}/2$$

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$$W_{OFDM} pprox K f_{\Delta}/2 \qquad f_{samp} > K f_{\Delta}$$

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$$f_{samp} = N/T_{obs}$$

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N samples per symbol
$$f_{samp} = N/T_{obs} = Nf_{\Delta} > Kf_{\Delta}$$

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$$\{x(t) \exp(i2\pi f_{rc}t)\}$$
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$$x(t) = g_{\text{rec}}(t) \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}t)$$

$$x_n = x(nT_{obs}/N) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k f_{\Delta}nT_{obs}}{N}\right)$$

$$f_{samp} = N/T_{obs} = Nf_{\Delta}$$

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The above gives a formula for how to compute the samples of the OFDM signal

OFDM signal = $\operatorname{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$ $0 \le t \le T_{\text{obs}}$

$$x_n = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right)$$

Let us now compute the Fourier transform of the samples (as of now, for no particular reason)

$$X(\nu) = \sum_{n=0}^{N-1} x_n \exp(-i2\pi\nu n)$$

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Take N samples of this Fourier transform (as of now, for no particular reason)

$$X_m = X(m/N) = \sum_{n=0}^{N-1} x_n \exp\left(-\frac{i2\pi mn}{N}\right)$$

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DFT
IDFT

OFDM signal = $\operatorname{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$ $0 \le t \le T_{\text{obs}}$

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$$X(\nu) = \sum_{n=0}^{N-1} x_n \exp(-i2\pi\nu n)$$

IDFT is <u>VERY</u> fast. We can get x_n FAST if we know X_m

Take N samples of this Fourier transform (as of now, for no particular reason)

$$X_m = \sum_{n=0}^{N-1} x_n \exp\left(-\frac{i2\pi mn}{N}\right) \qquad x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$
DFT IDFT

Logics

We have $\{a_n\}$

We need $\{x_n\}$

We know that
$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

And that this can be computed FAST

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Conclusion: We need to link $\{a_n\}$ and $\{X_m\}$

$$x_n = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right)$$

$$x_n = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right)$$

$$= \sum_{k=0}^{-g_0 - 1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right) + \sum_{k=-g_0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right)$$

$$x_{n} = \sum_{k=0}^{K-1} a_{k} \exp\left(\frac{i2\pi g_{k}n}{N}\right) = \sum_{k=0}^{K-1} a_{k} \exp\left(\frac{i2\pi (g_{0} + k)n}{N}\right)$$

$$g_{0} < 0$$

$$= \sum_{k=0}^{-g_{0}-1} a_{k} \exp\left(\frac{i2\pi (g_{0} + k)n}{N}\right) + \sum_{k=-g_{0}}^{K-1} a_{k} \exp\left(\frac{i2\pi (g_{0} + k)n}{N}\right)$$

$$= \sum_{k=0}^{-g_{0}-1} a_{k} \exp\left(\frac{i2\pi (g_{0} + k + N)n}{N}\right) + \sum_{k=-g_{0}}^{K-1} a_{k} \exp\left(\frac{i2\pi (g_{0} + k)n}{N}\right)$$

$$\begin{split} x_n &= \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right) \\ &= \sum_{k=0}^{-g_0-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right) + \sum_{k=-g_0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right) \\ &= \sum_{k=0}^{-g_0-1} a_k \exp\left(\frac{i2\pi (g_0 + k + N)n}{N}\right) + \sum_{k=-g_0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right) \\ &= \sum_{k=0}^{Variable \ \text{substitutions}} \begin{bmatrix} w = g_0 + k \end{bmatrix} & [m = g_0 + k] \end{split}$$

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$$x_n = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi g_k n}{N}\right) = \sum_{k=0}^{K-1} a_k \exp\left(\frac{i2\pi (g_0 + k)n}{N}\right)$$

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$$= \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

This we know from the IDFT

$$x_{n} = \sum_{k=0}^{K-1} a_{k} \exp\left(\frac{i2\pi g_{k}n}{N}\right) = \sum_{k=0}^{K-1} a_{k} \exp\left(\frac{i2\pi (g_{0} + k)n}{N}\right)$$

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Thus,

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

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Thus,

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

$$X_m = Na_{m-(q_0+N)}, \quad g_0 + N \le m \le N-1$$

Altogether,

1. Take a block of K data symbols {a_k}

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

 $X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$
 $X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N - 1$

- 1. Take a block of K data symbols $\{a_k\}$
- 2. Select a sampling rate, by choosing $N \ge K$

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- 1. Take a block of K data symbols {a_k}
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- 3. Find N values $\{X_m\}$ according to the box below

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$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N-1$$

- 1. Take a block of K data symbols {a_k}
- 2. Select a sampling rate, by choosing N ≥ K
- 3. Find N values $\{X_m\}$ according to the box below
- 4. Compute $\{x_n\}$ using an IDFT

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

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$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

Speed: N² multiplications

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$
 $X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$
 $X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N - 1$

Altogether,

FFT = "Fast Fourier transform"

- 1. Take a block of K data symbols $\{a_k\}$
- 2. Select a sampling rate, by choosing $N=2^{L} \ge K$
- 3. Find N values $\{X_m\}$ according to the box below
- 4. Compute $\{x_n\}$ using an IFFT

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

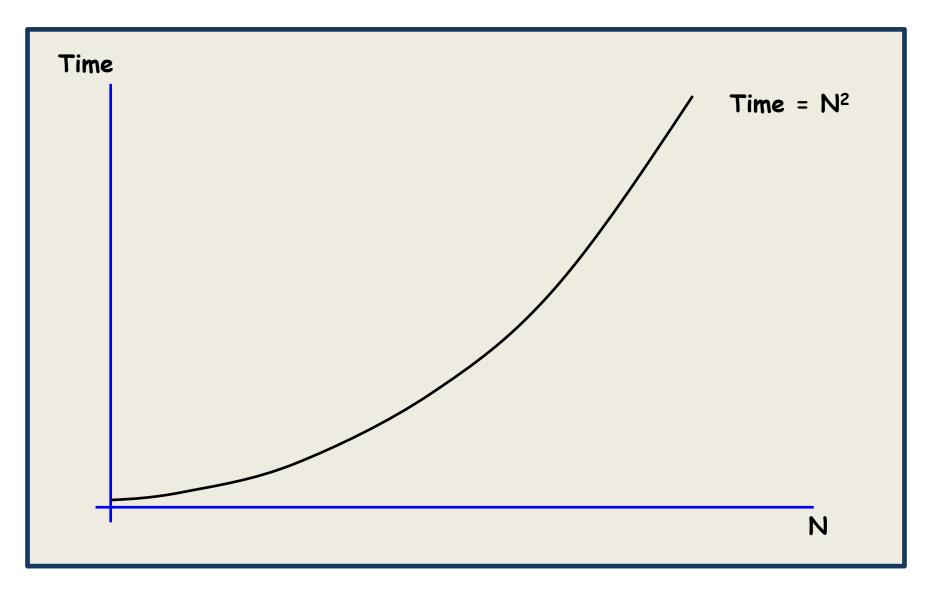
Speed: N² multiplications

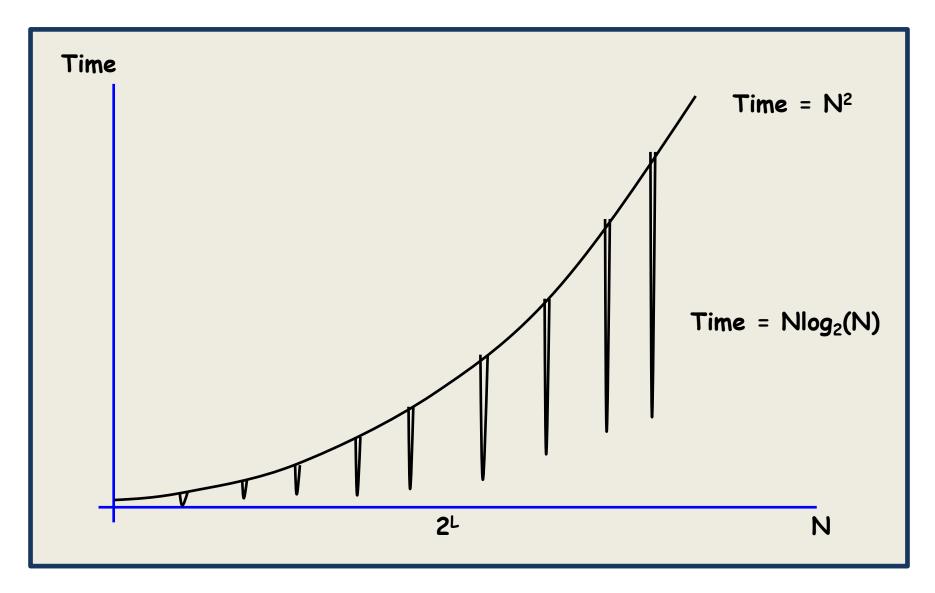
Speed: Nlog2N

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N-1$$





Left to do

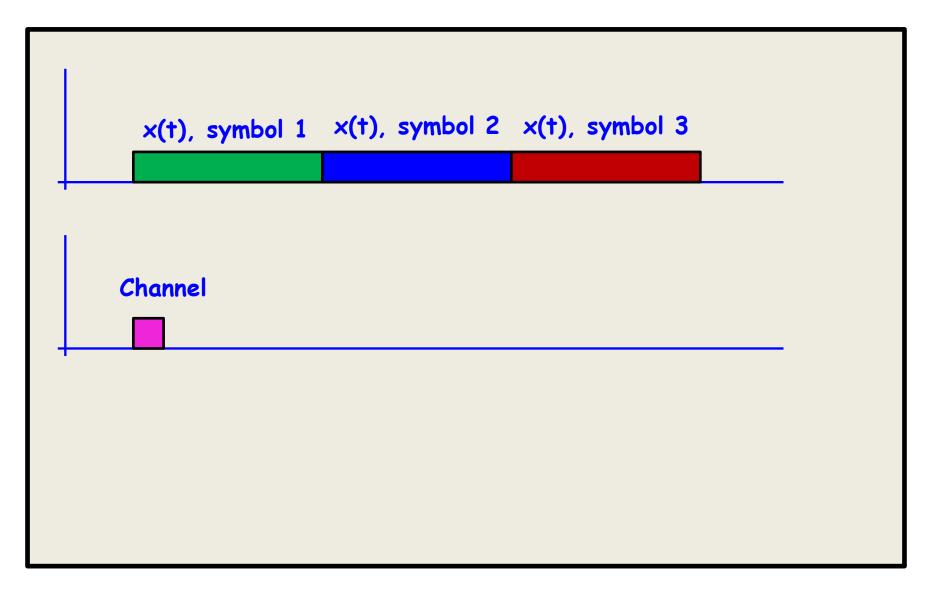
1

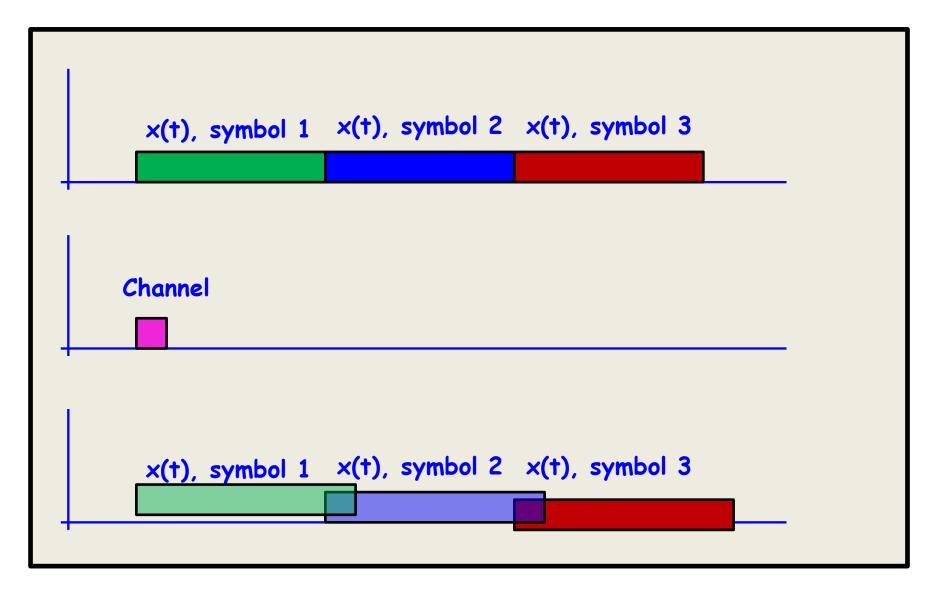
- 2. D/A conversion
- 3. Modulation to band-pass
- 4. Recevier

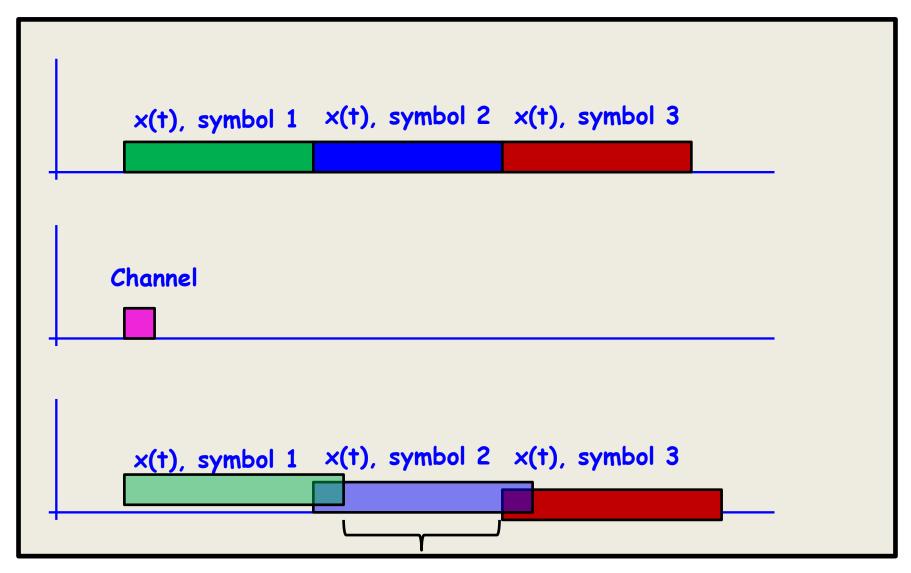
Left to do

- 1. Add a cyclic prefix to deal with channel effects
- 2. D/A conversion
- 3. Modulation to band-pass
- 4. Recevier

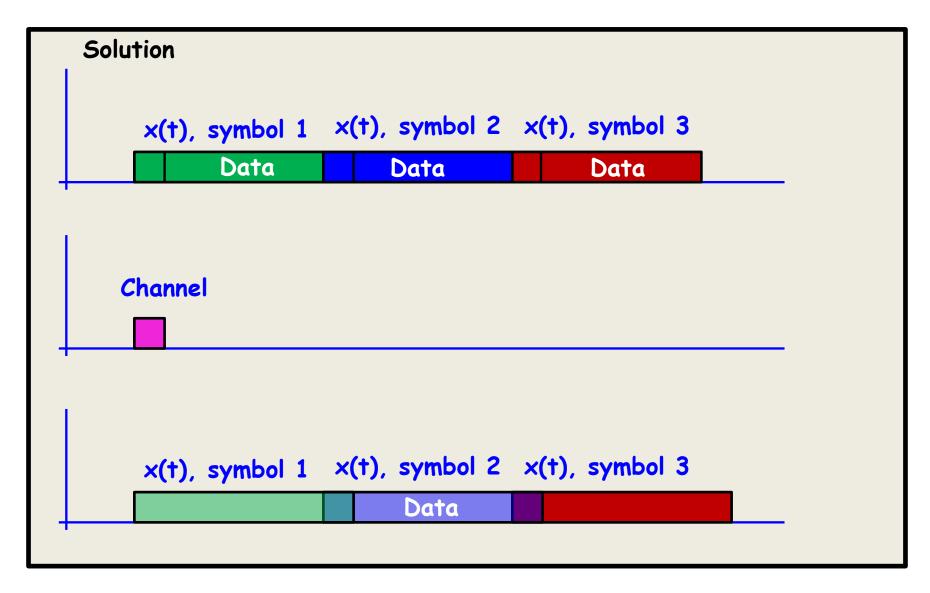
```
x(t), symbol 1 x(t), symbol 2 x(t), symbol 3
```

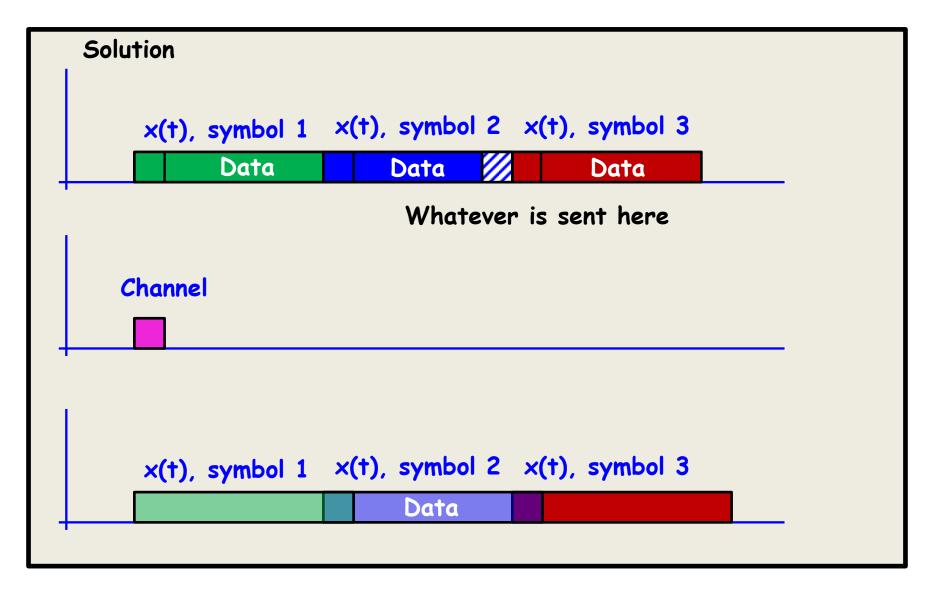


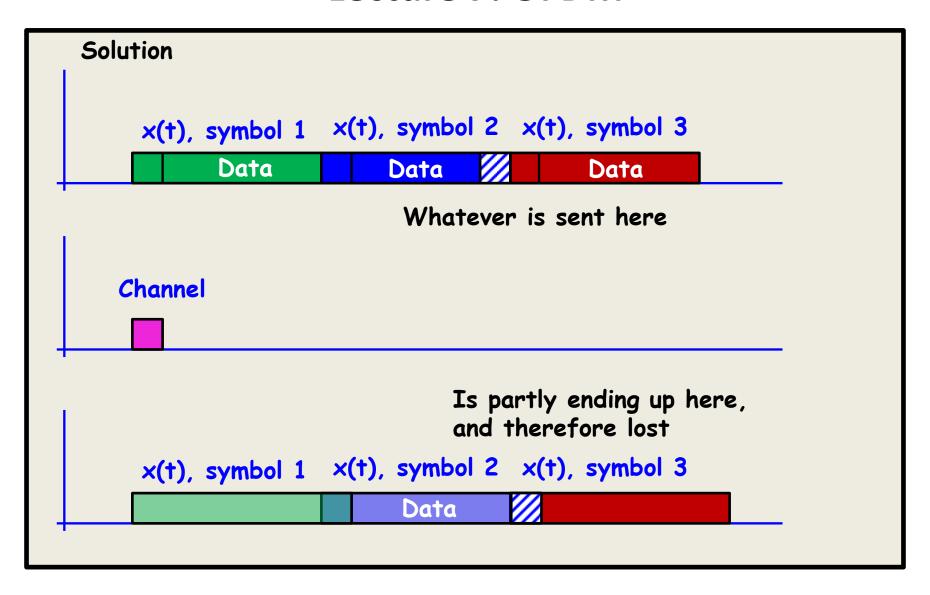


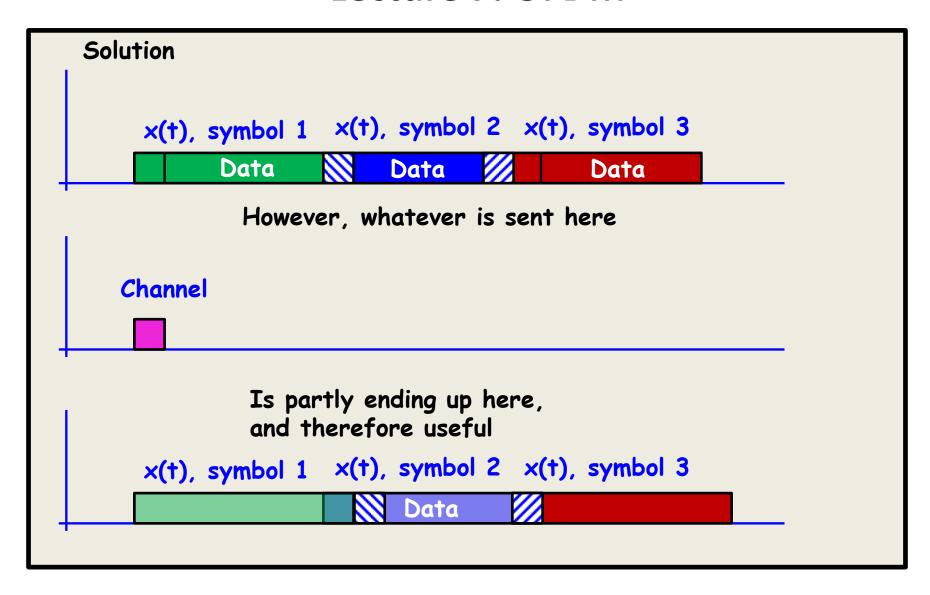


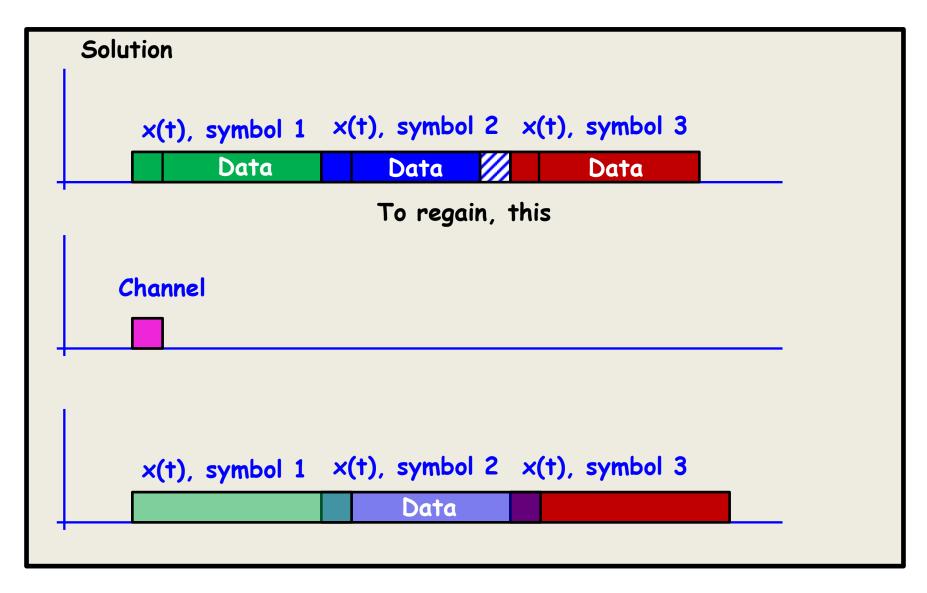
Useful part

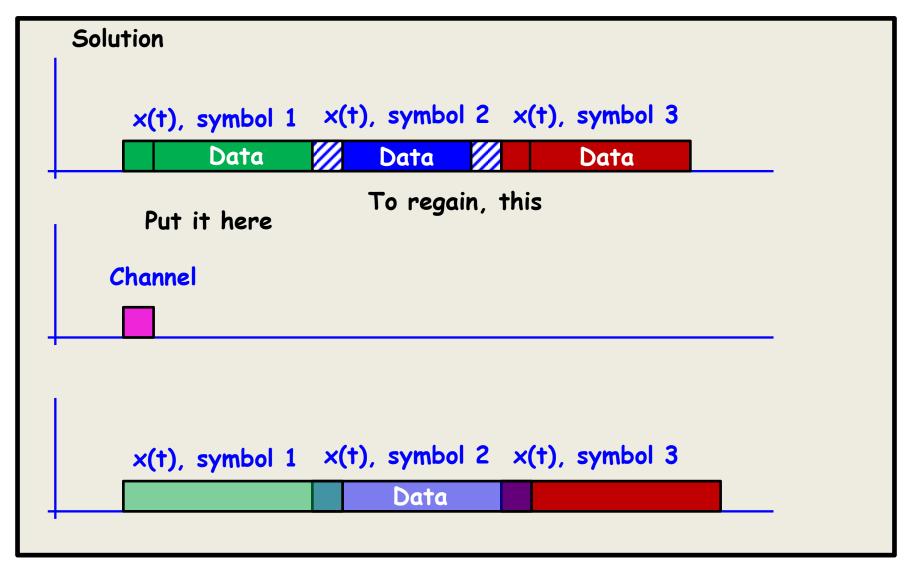




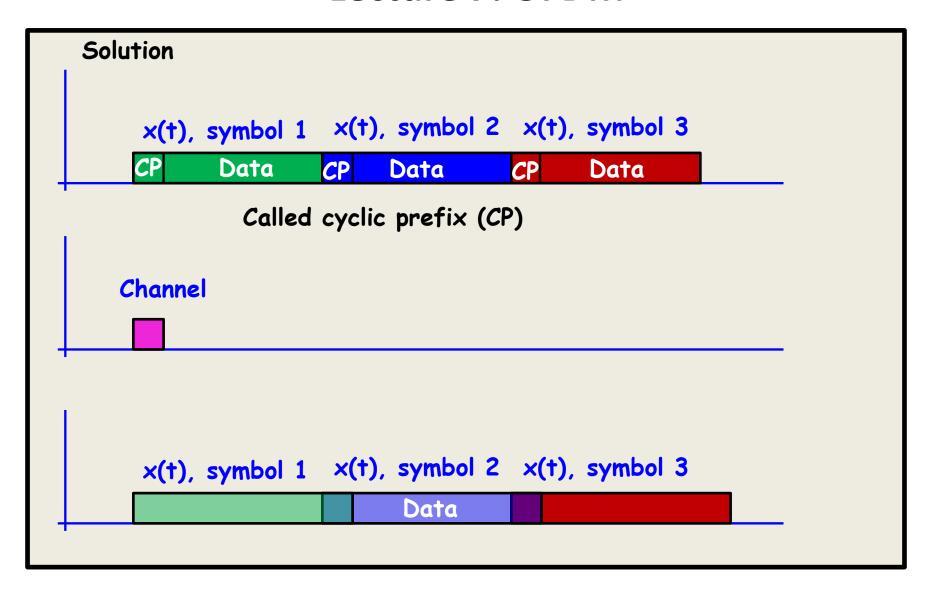


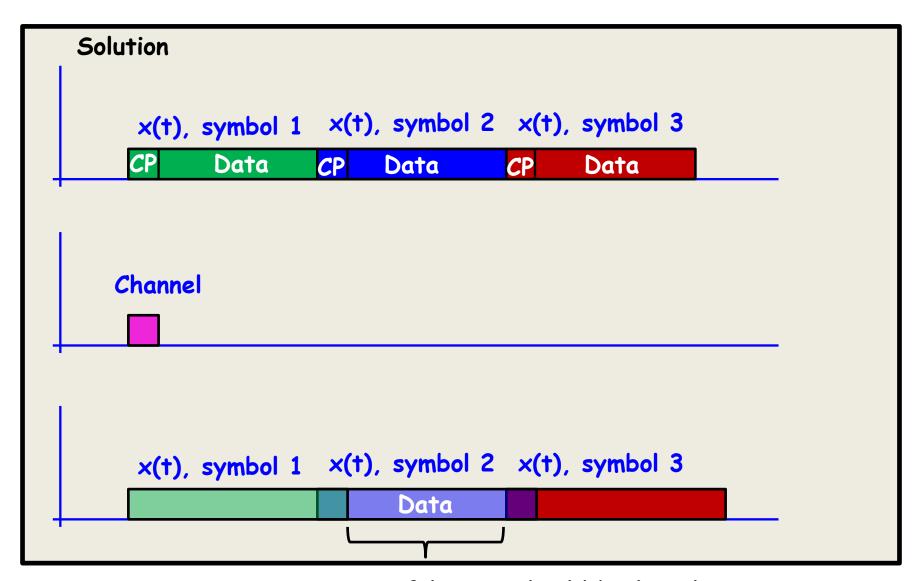




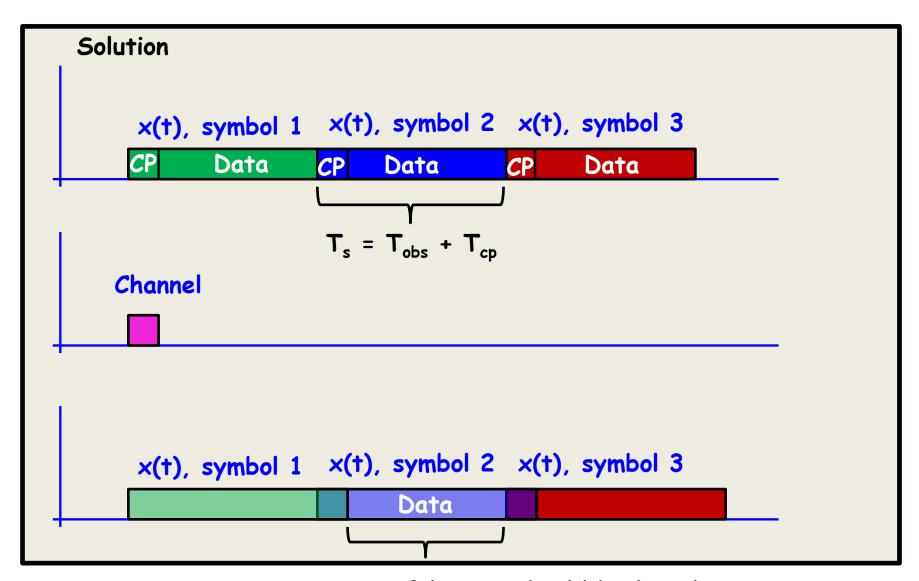


Useful part

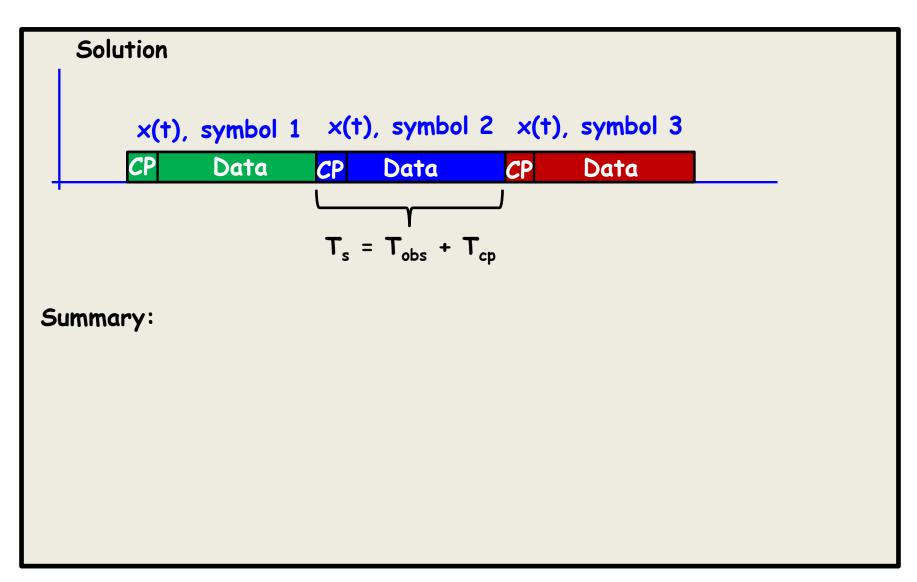


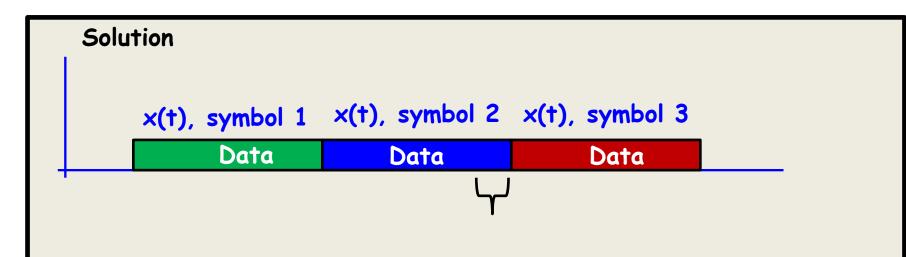


Useful part, should be length Tobs



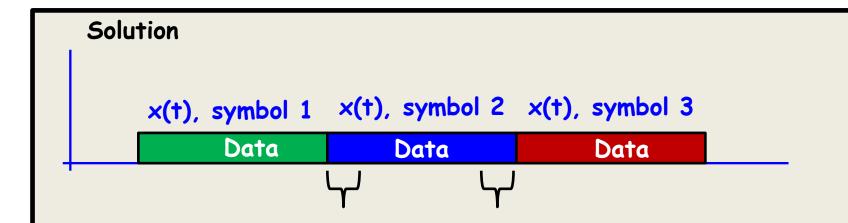
Useful part, should be length Tobs



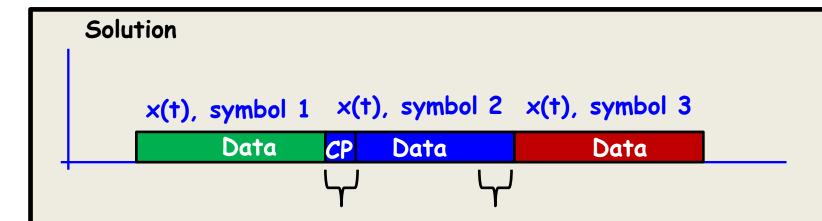


Summary:

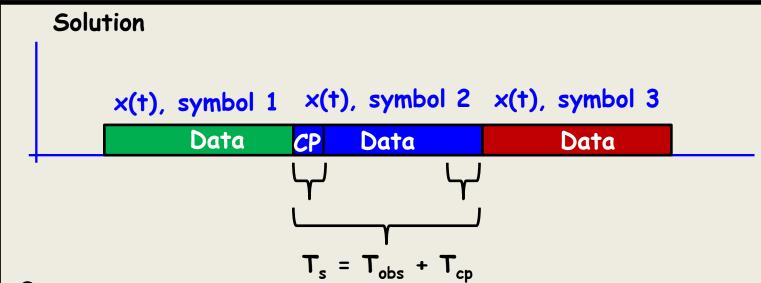
1. Marked region is partly lost since it interferes with next block



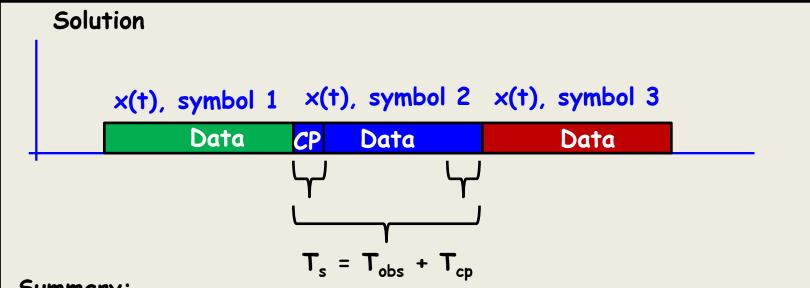
- 1. Marked region is partly lost since it interferes with next block
- 2. First part is partly lost, since previous block interferes with it



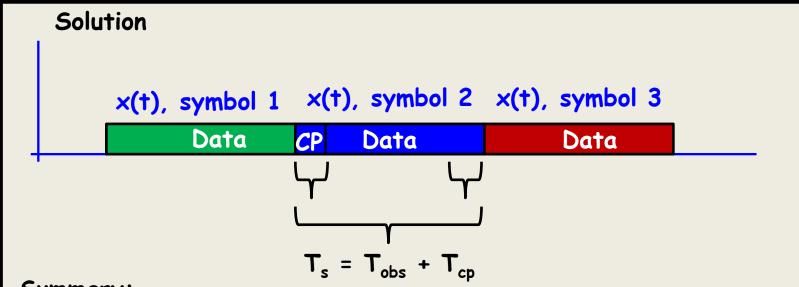
- 1. Marked region is partly lost since it interferes with next block
- 2. First part is partly lost, since previous block interferes with it
- 3. Put the last part in the first part, call it CP



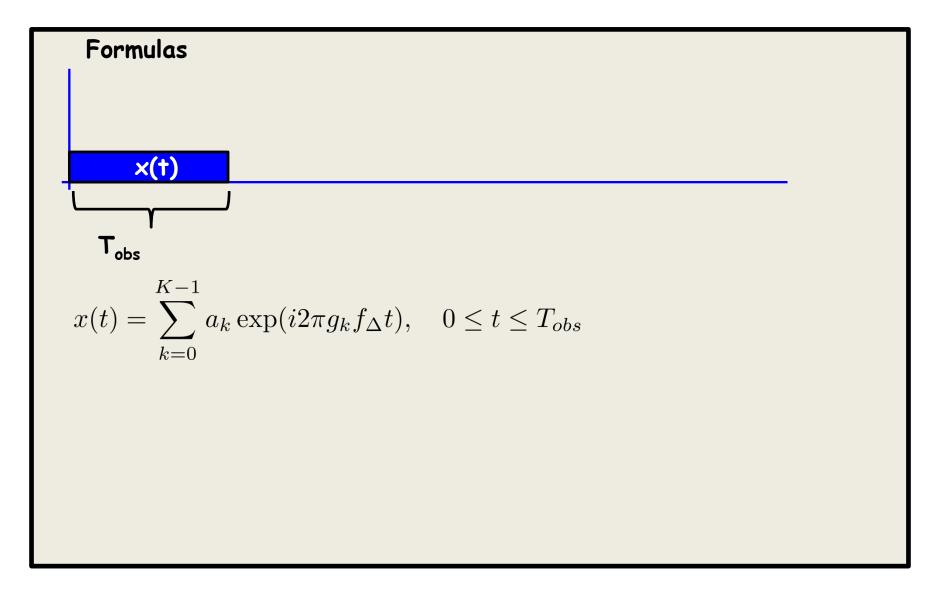
- 1. Marked region is partly lost since it interferes with next block
- 2. First part is partly lost, since previous block interferes with it
- 3. Put the last part in the first part, call it CP
- 4. Length of data block should be $T_{obs}=1/f_{\Delta}$

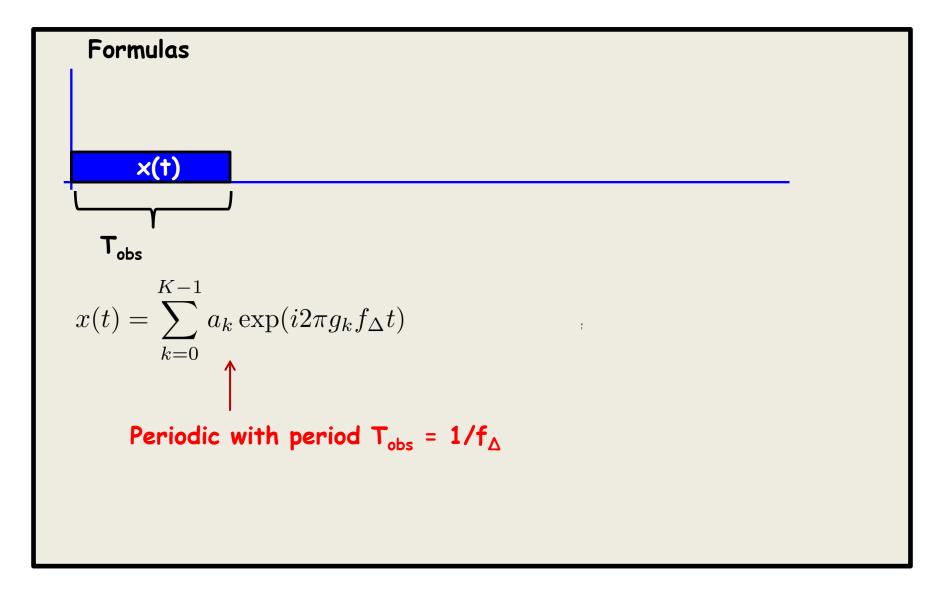


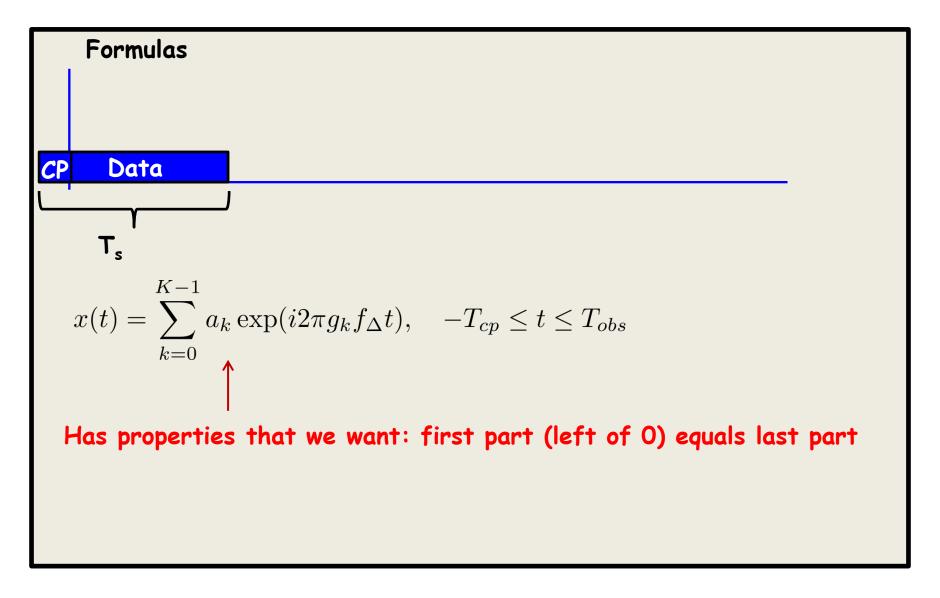
- 1. Marked region is partly lost since it interferes with next block
- 2. First part is partly lost, since previous block interferes with it
- 3. Put the last part in the first part, call it CP
- 4. Length of data block should be $T_{obs}=1/f_{\Delta}$
- 5. No proof, yet, why this should be good, but it is plausible, since the lost part is present in the beginning

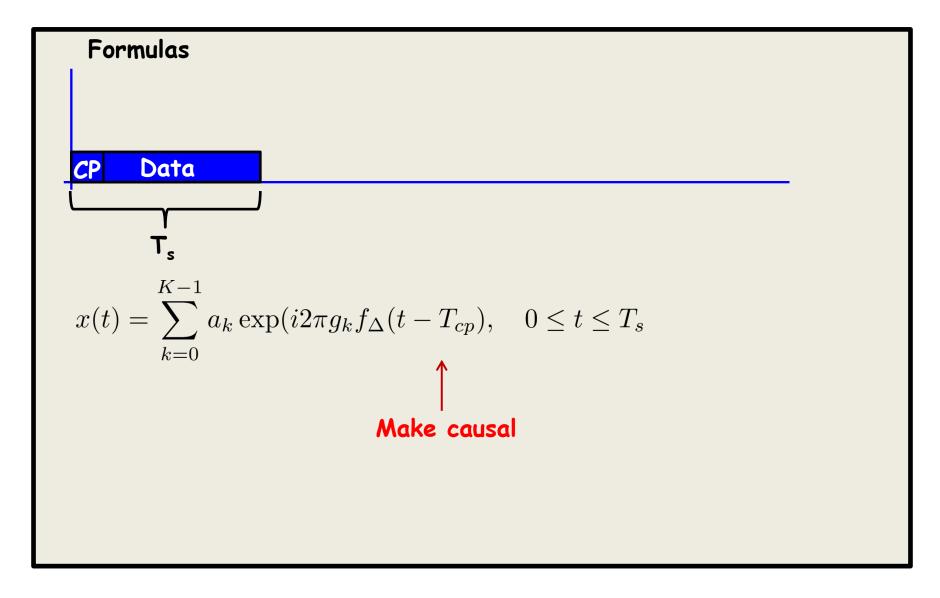


- 1. Marked region is partly lost since it interferes with next block
- 2. First part is partly lost, since previous block interferes with it
- 3. Put the last part in the first part, call it CP
- 4. Length of data block should be $T_{obs}=1/f_{\Delta}$
- 5. No proof, yet, why this should be good, but it is plausible, since the lost part is present in the beginning
- 6. Spectral efficiency loss







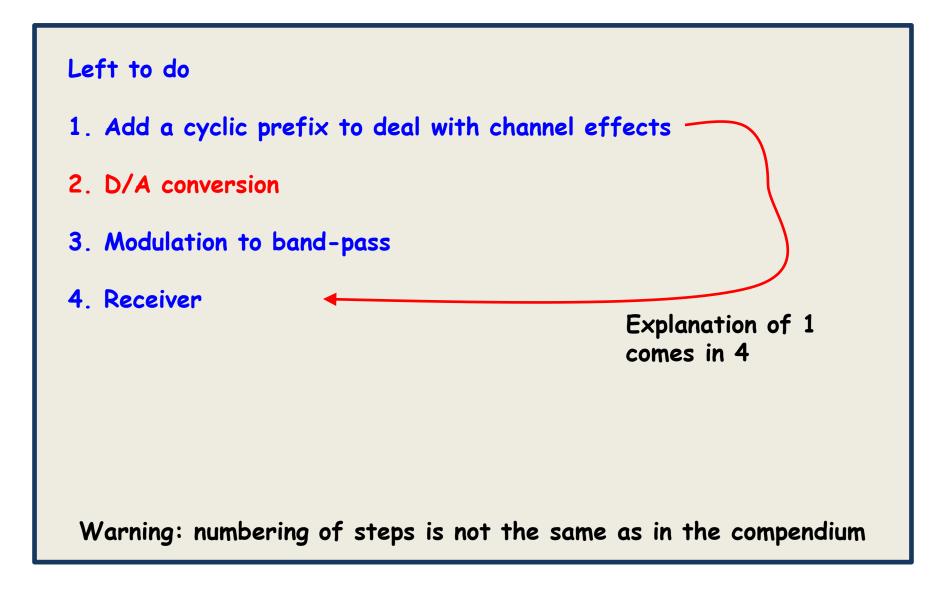


- 1. Take a block of K data symbols $\{a_k\}$
- 2. Select a sampling rate, by choosing $N \ge K$
- 3. Find N values $\{X_m\}$ according to the box below
- 4. Compute $\{x_n\}$ using an IDFT
- 5. Add last L samples to the beginning

$$x_n = \frac{1}{N} \sum_{m=0}^{N-1} X_m \exp\left(\frac{i2\pi mn}{N}\right)$$

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

 $X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$
 $X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N - 1$



OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \le t \le T_s$$

OFDM signal =
$$\text{Re}\{x(t) \exp(i2\pi f_{rc}t)\}$$

 $x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \le t \le T_s$
 $u_n = \text{IFFT}(\mathbf{X})$
 $X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$
 $X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$
 $X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N - 1$
 $\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots u_{N-1}]$

OFDM signal = Re
$$\{x(t) \exp(i2\pi f_{rc}t)\}$$

$$x(t) = \sum_{k=0}^{K-1} a_k \exp(i2\pi g_k f_{\Delta}(t - T_{cp})), \quad 0 \le t \le T_s$$

$$u_n = \text{IFFT}(\mathbf{X})$$

$$X_m = N a_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$$

$$X_m = N a_{m-(g_0+N)}, \quad g_0 + N \le m \le N - 1$$

$$\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots u_{N-1} \ u_0 \ \dots u_{N-1}]$$
CP

OFDM signal = Re
$$\{x(t) \exp(i2\pi f_{rc}t)\}$$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{samp})$$

$$u_n = \text{IFFT}(\mathbf{X})$$

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

$$X_m = 0, \quad g_{K-1} + 1 \le m \le g_0 + N - 1$$

$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N - 1$$

$$\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots u_{N-1} \ u_0 \ \dots u_{N-1}]$$

$$\mathbf{CP}$$

OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{samp})$$

$$g_{ideal}(t) = \frac{\sin(\pi f_{samp}t)}{\pi f_{samp}t}$$

$$u_n = IFFT(\mathbf{X})$$

$$X_m = Na_{m-g_0}, \quad 0 \le m \le g_{K-1}$$

$$X_m = 0$$
, $g_{K-1} + 1 \le m \le g_0 + N - 1$

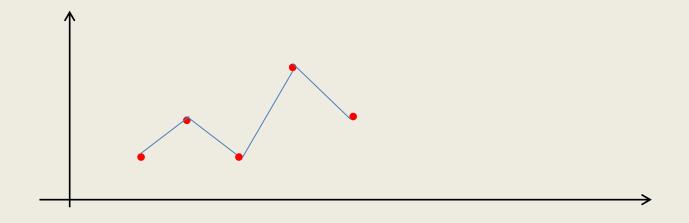
$$X_m = Na_{m-(g_0+N)}, \quad g_0 + N \le m \le N-1$$

$$\mathbf{x} = [u_{N-L} \ u_{N-L+1} \ \dots u_{N-1}] u_0 \ \dots u_{N-1}]$$

OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{samp})$$

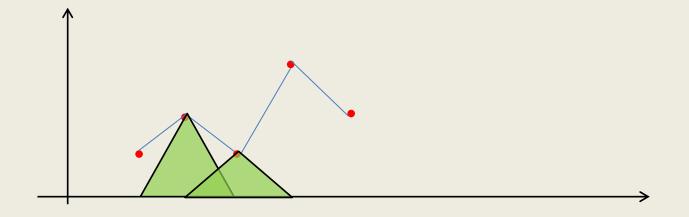
$$g_{ideal}(t) = \frac{\sin(\pi f_{samp}t)}{\pi f_{samp}t}$$



OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{samp})$$

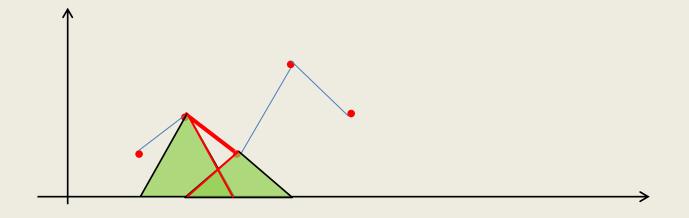
$$g_{ideal}(t) = \frac{\sin(\pi f_{samp}t)}{\pi f_{samp}t}$$



OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{samp})$$

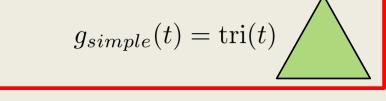
$$g_{ideal}(t) = \frac{\sin(\pi f_{samp}t)}{\pi f_{samp}t}$$

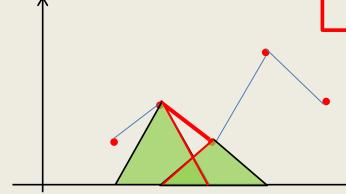


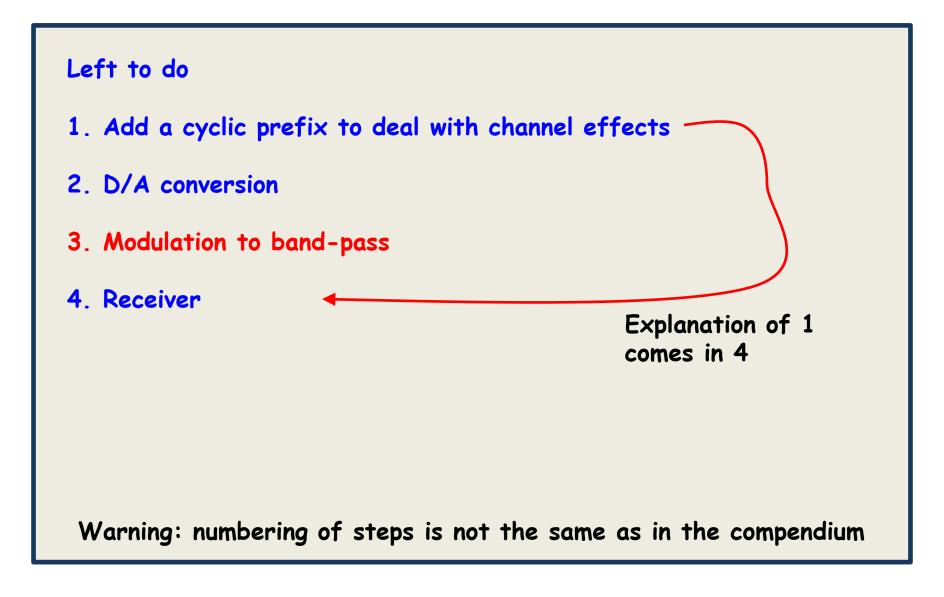
OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$

$$x(t) = \sum_{\ell} x_{\ell} g(t - \ell/f_{samp})$$

$$g_{ideal}(t) = \frac{\sin(\pi f_{samp}t)}{\pi f_{samp}t}$$







Left to do

- 1. Add a cyclic prefix to deal with channel effects
- 2. D/A conversion
- 3. Modulation to band-pass OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$
- 4 Receiver

Very simple

Warning: numbering of steps is not the same as in the compendium

Left to do

- 1. Add a cyclic prefix to deal with channel effects
- 2. D/A conversion
- 3. Modulation to band-pass OFDM signal = $Re\{x(t) \exp(i2\pi f_{rc}t)\}$
- 4. Receiver Next lecture

Warning: numbering of steps is not the same as in the compendium