

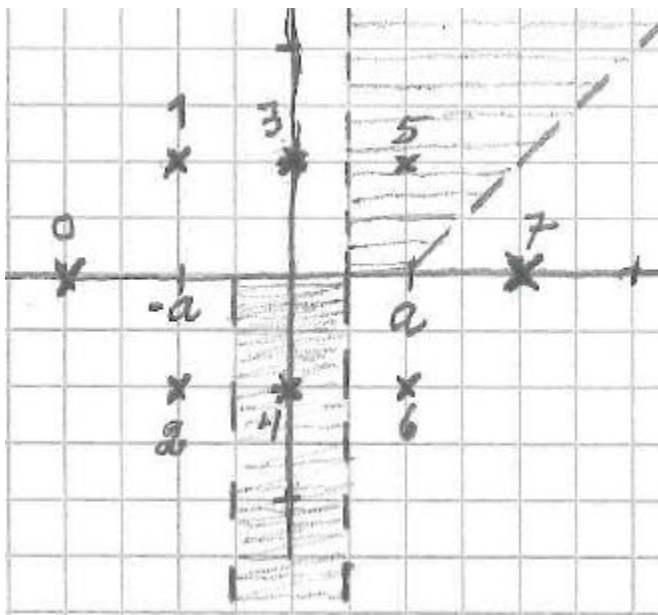
**Answers & Hints to the exam in the course Digital Communications, Advanced Course (ETTN01), December 14, 2010, 08-13.**

Problem 1

- a) True, since  $R_b/R_s = 6$  in both cases.
- b) True, since  $T_s = 8T_b/3$ . Alternatively: Each coded bit carries  $2/3$  of an information bit.
- c) False, since  $C_{max} = 1.4427$  Mbps which is smaller than 1.5 Mbps.
- d) True.
- e) False, diversity helps indeed in this case.

Problem 2

ai)



aii) The average received energy per bit is found to be  $0.75a^2$ , and  $D_{min}^2 = a^2$ . So,  $d_{min}^2 = 2/3$ .

bi) Let  $d = -a$  correspond to the signal point  $z_0 = (0.4a, 0, 0.6a, -0.4a)$ .

Similarly, let  $d = a$  correspond to the signal point  $z_1 = (-0.4a, 0, -0.6a, 0.4a) = -z_0$ .

Then we find that  $D_{r,z_0}^2 = 0.79a^2$  and  $D_{r,z_1}^2 = 0.87a^2$ . Hence, the decision is  $\hat{d} = -a$ , i.e. that  $-a$  was sent.

bii) We find that  $D_{z_0,z_1}^2 = 2.72a^2$ . So,  $P_b = Q(\sqrt{1.36a^2/N_0})$ .

### Problem 3

a)  $C/W = 4$  implies that  $P_z/N_0 = 15W = 75 \cdot 10^6$ . If the bandwidth is increased to  $W_1 = 4W = 20$  MHz, then the capacity  $C_1$  will be,

$C_1 = W_1 \log_2(1 + \frac{P_z}{N_0 W_1}) = 4W \log_2(19/4) = 4W 2.248 = 45$  Mbps, and  $\rho_1 = C_1/W_1 = 2.248$  bps/Hz.

b)  $C_2 = W \log_2(1 + \frac{P_z}{N_0 W}) = 4W 2.248 = W \log_2(509.17)$ . Hence,  $P_z/N_0 = 508.17W = 2.54 \cdot 10^9$ . ( $\rho_2 = C_2/W = 8.99$  bps/Hz).

c) In a) the bandwidth used is quite large which is a disadvantage. On the other hand the capacity is kept the same as in b).

In b) the received signal power is significantly stronger. We can use a higher bandwidth efficiency, and a smaller bandwidth than in a). The case in b) could represent a certain communication range, and the case in a) could represent a longer communication range.

### Problem 4

ai) The path in the trellis is: state 13, state 6, state 3, state 9.

The corresponding output sequence of signal alternatives is:  $s_3(t), s_3(t), s_2(t)$ .

aii) The bandwidth efficiency  $R_b/W$  will be doubled which is good. However, the implementation will be slightly more complicated since we have to get a rate 1/2 encoder and a 16-ary constellation to work together.

b) See the compendium.

## Problem 5

$$P_b \approx \frac{M^{N_t}}{y^{N_r}} \leq 10^{-6}$$

$$\text{Hence, } \frac{M^{N_t/N_r}}{y} \leq 10^{-6/N_r}.$$

Using that  $R_b = N_t \log_2(M)/T_s$  in the given expression for  $y$ , and re-arranging, we obtain that,

$$\frac{L^3}{a} \leq \frac{10^{-6/N_r} 1.5 P_t T_s / N_0}{(M-1) N_t M^{N_t/N_r}}$$

So,  $M = 4$  gives  $L_{max}$ , implying that,

$$\frac{L_{max}^3}{a} = \frac{10^{-6/N_r} 0.5 P_t T_s / N_0}{N_t 4^{N_t/N_r}}$$

$$\text{and } R_b = N_t 2 / T_s.$$

5i) With  $N_t = N_r = 1$  (SISO) we obtain that,

$$\frac{L_{max}^3}{a} = \frac{10^{-6} P_t T_s / N_0}{8}$$

$$\text{So, } L_{max} = 5 \cdot 10^{-3} (a P_t T_s / N_0)^{1/3}, \text{ and } R_{b,L_{max}} = 2 / T_s.$$

5ii) With  $N_t = 2$  and  $N_r = 4$  we obtain that,

$$\frac{L_{max}^3}{a} = \frac{10^{-6/4} P_t T_s / N_0}{8}$$

So,  $L_{max} = 0.158 (a P_t T_s / N_0)^{1/3}$ , and  $R_{b,L_{max}} = 4 / T_s$ . This means twice as high bitrate and almost 32 times longer range compared with SISO.

5iii) With  $N_t = 2$ ,  $N_r = 4$  and  $L = L_{max}/16$  we obtain:

$$\frac{L^3}{a} = \frac{L_{max}^3}{a} \cdot \frac{1}{16^3}$$

$$\frac{L^3}{a} = \frac{10^{-6/4} P_t T_s / N_0}{8} \cdot \frac{1}{16^3} \leq \frac{10^{-6/4} 1.5 P_t T_s / N_0}{(M-1) 2M^{1/2}}$$

From the inequality we find that,

$$(M-1) \sqrt{M} \leq 8 \cdot 16^3 \cdot 3/4 = 24576$$

This gives us that  $M = 256$  and that  $R_b = 16 / T_s$  (8 times higher than SISO).

The diversity gain obtained by increasing  $N_r$  here gives us a significantly longer communication distance. Increasing also  $N_t$  increases the bit rate.