

Bipolar Junction Transistors (BJT)

Ideal Transistor

Bipolar Transistor - Terminals

NPN Bipolar Transistor Physics

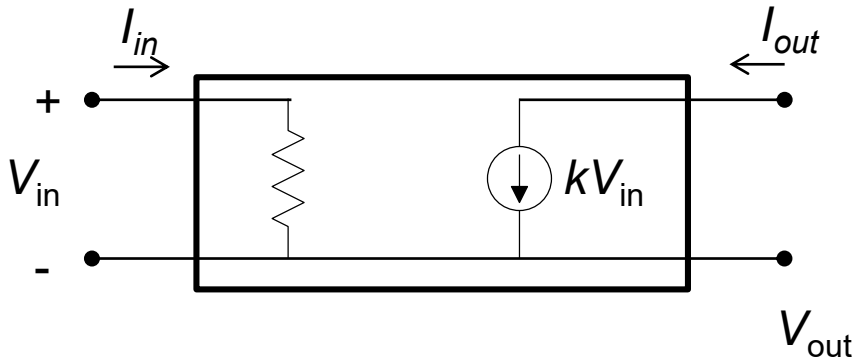
Large Signal Model

Early Effect

Small Signal Model

Reading: (Sedra, Smith, 7th edition)
4.1 - 4.2
6.2.2 (small signal model)
9.2.2 small signal model with
capacitances

Ideal Transistor - characteristics

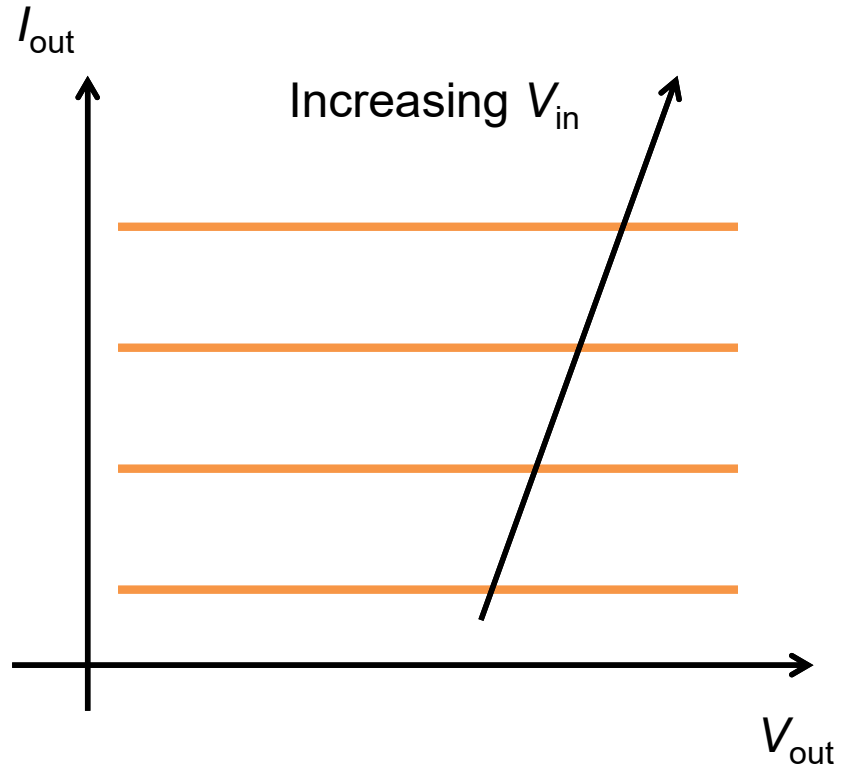


Two main transistor types:

Bipolar Transistors (Power Amplifiers, HighSpeed AD/DA converters)

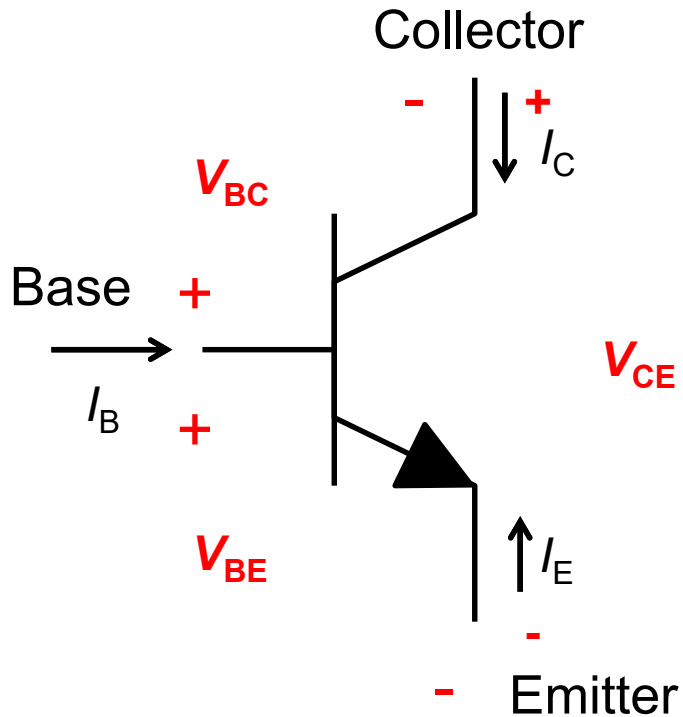
Field Effect Transistors (99.999% - mainly integrated circuits / digital)

current source controlled by a voltage (V_{in})



I_{out} independent of V_{out} !
 I_{in} independent of V_{out} !

Bipolar Transistor - npn



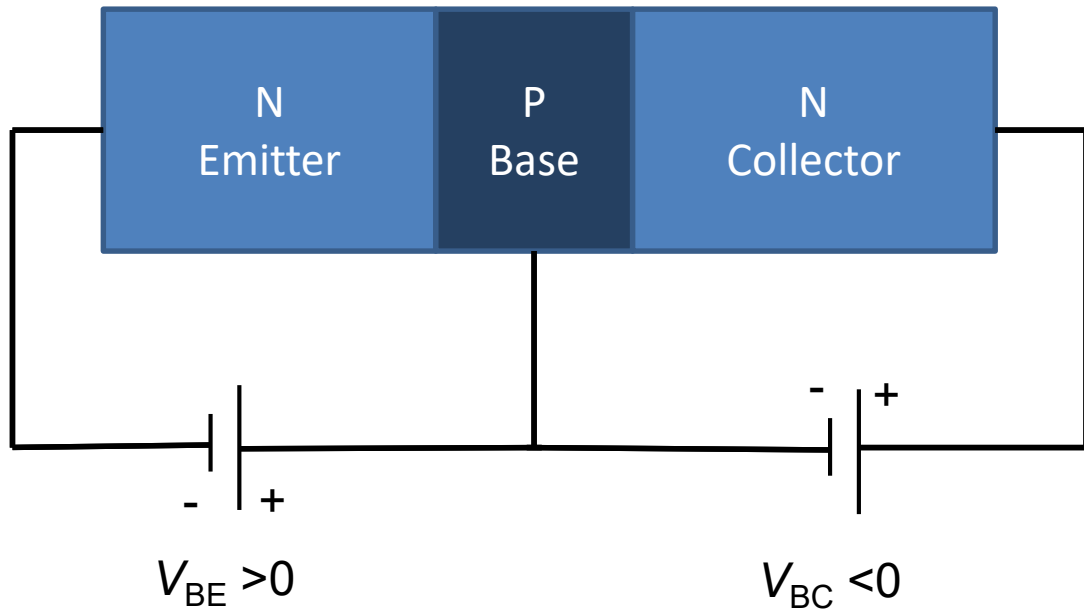
Sign convention: $V_{xy} = V_x - V_y$

I_B – base current

I_C – collector current

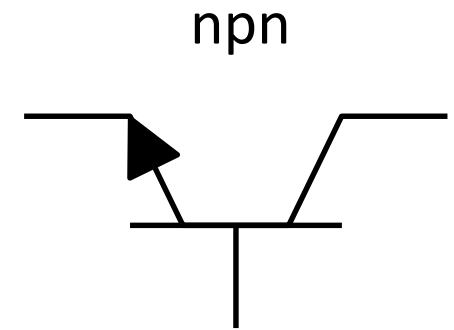
I_E – emitter Current = $-(I_C + I_B)$

Bipolar Transistor - npn in active mode

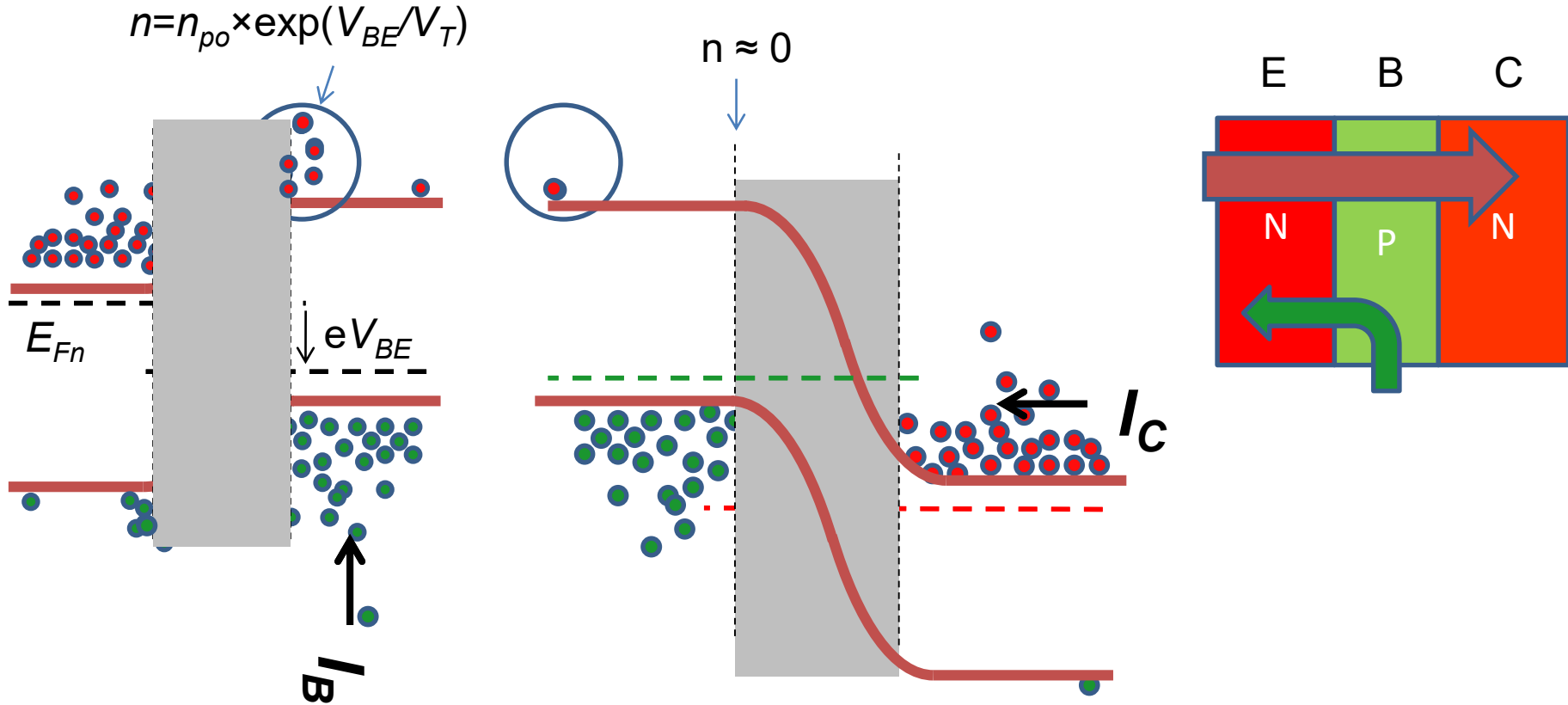


Forward biased PN junction

Reverse biased PN junction



Bipolar transistor: band structure



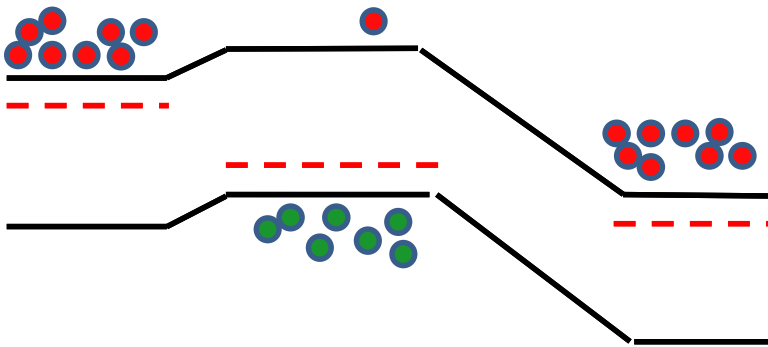
Forward biased emitter-base junction injects electrons

Reverse biased base-collector junction sweeps away electrons (independent of V_{BC})

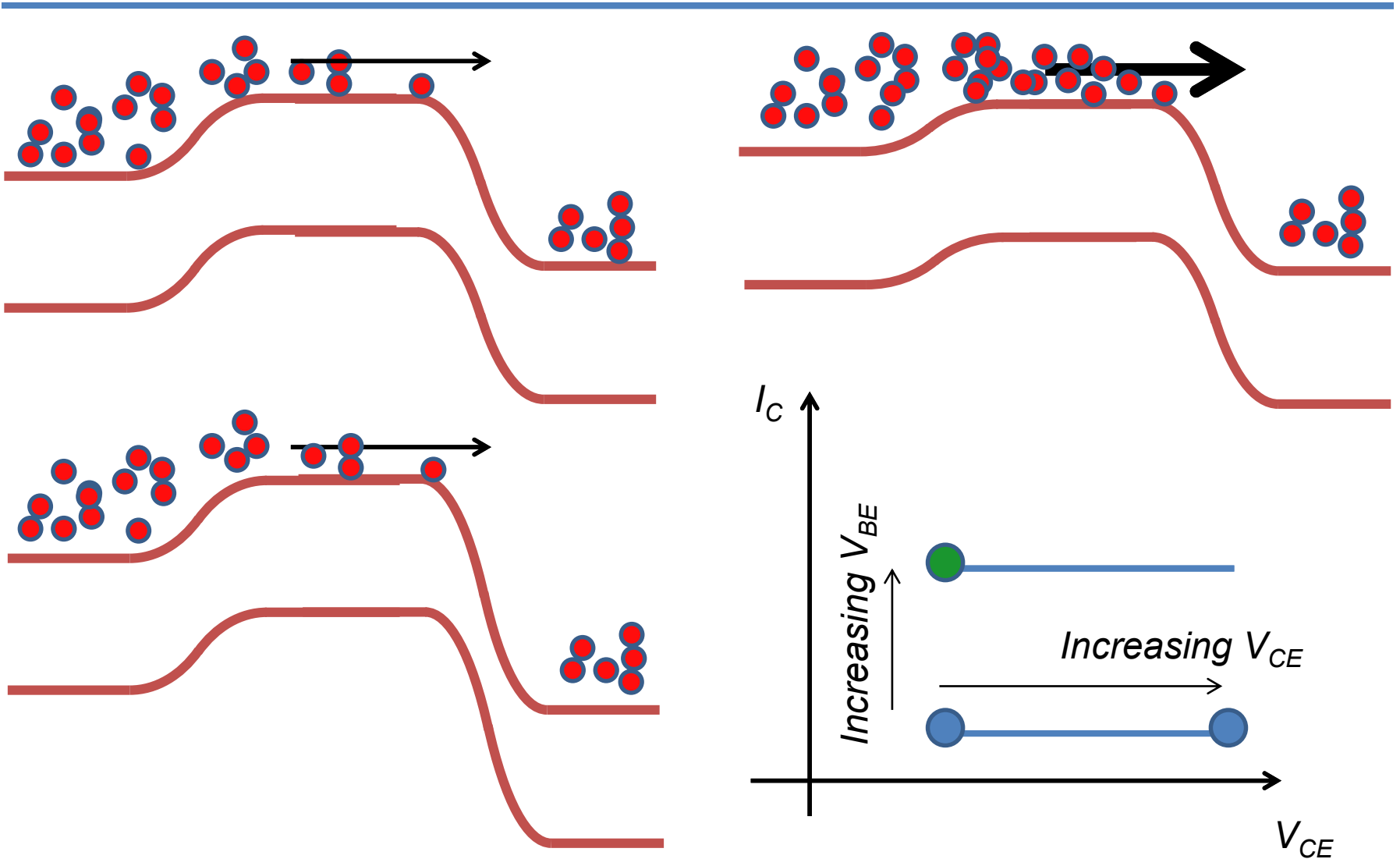
Operating modes

Active

EBJ: forward / CBJ: reverse

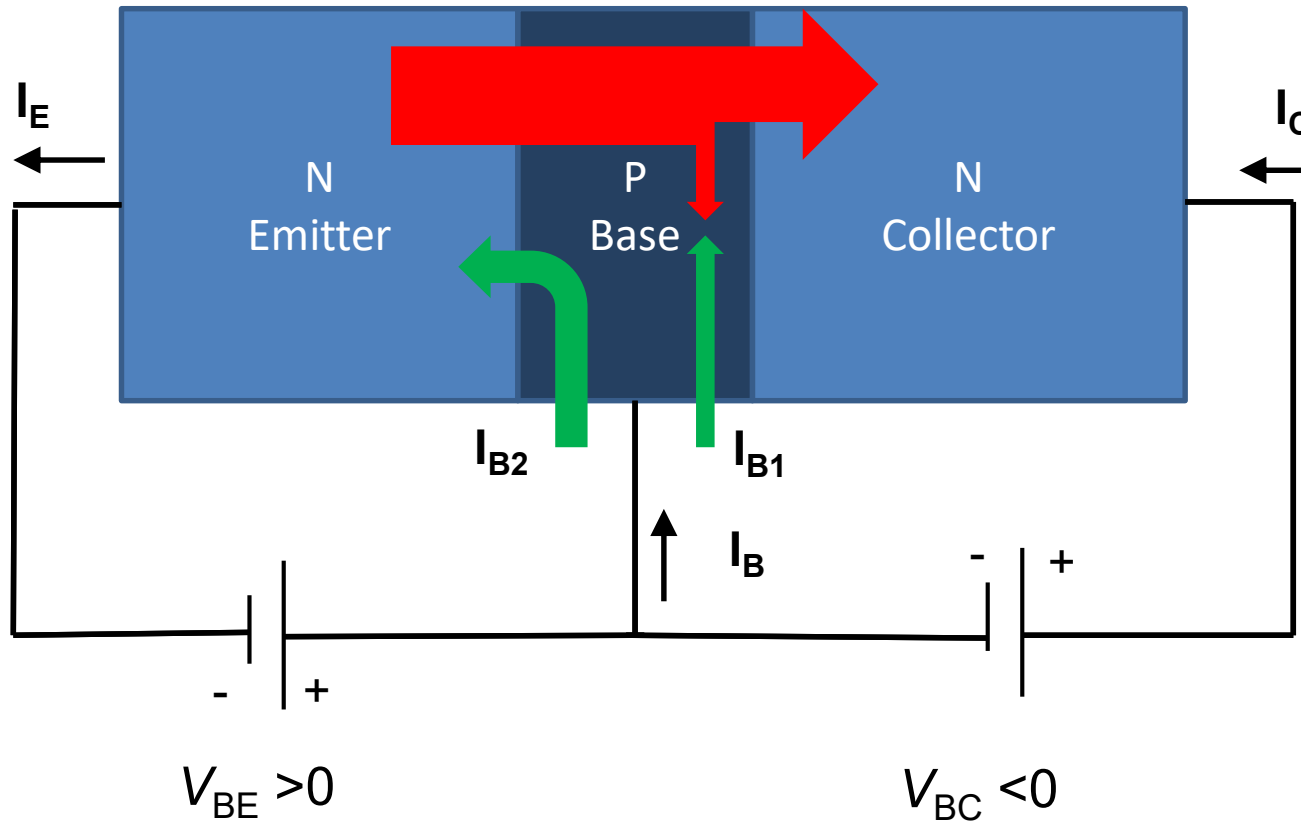


Electron diffusion currents – active mode



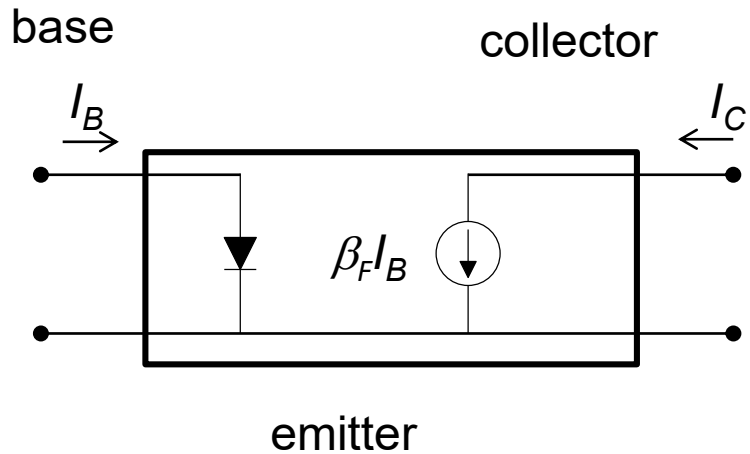
Currents in active mode

$$I_E + I_{B1} + I_{B2} + I_C = 0$$



On whiteboard: calculate I_C , I_{B1} , I_{B2} and the gain β

Large Signal Model – active mode



$$I_C = I_S e^{\frac{V_{BE}}{V_T}}$$

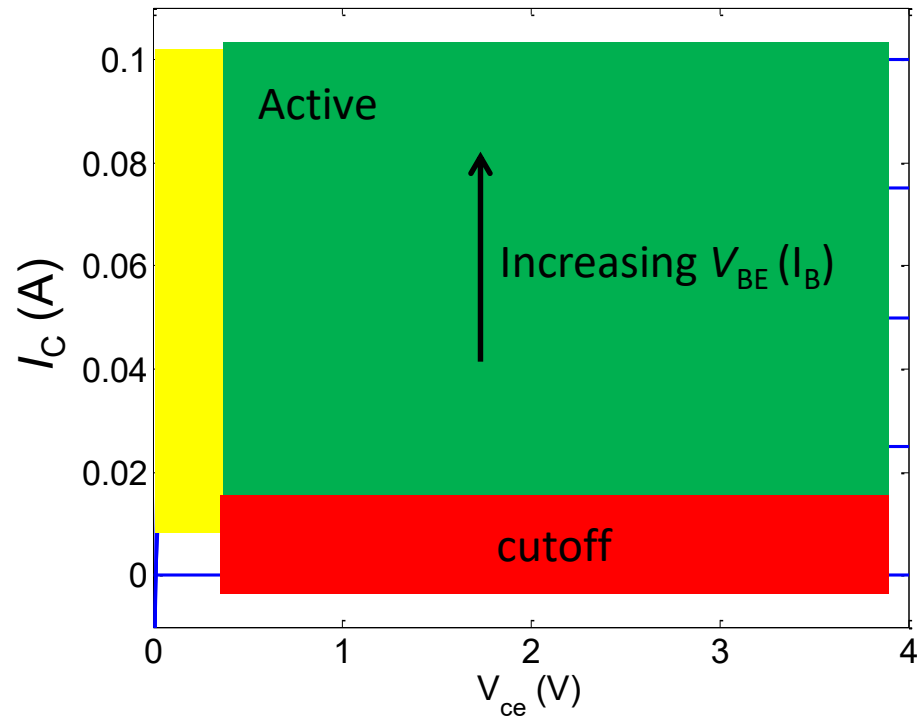
$$I_B = \frac{I_C}{\beta_-}$$

$$V_{BE} > 0.6V$$

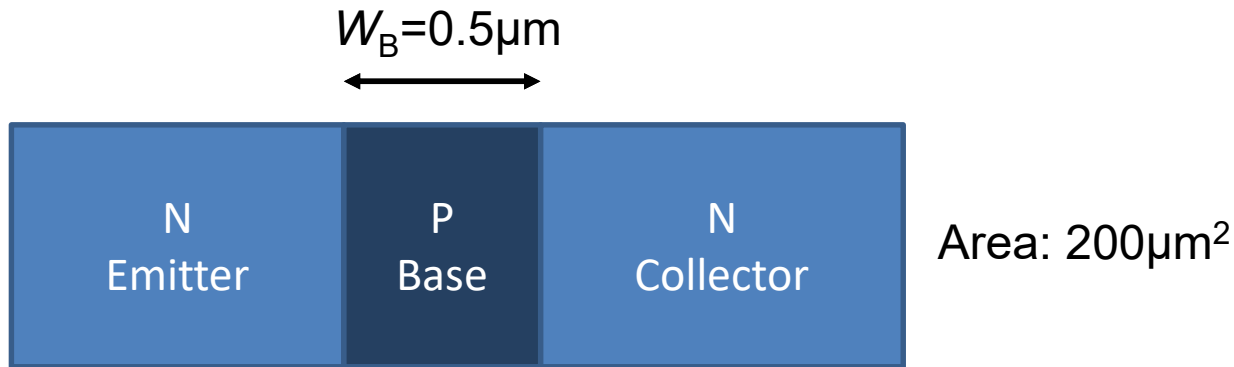
$$V_{BC} < 0.4V$$

$$V_{CE} < V_{BR}$$

β : current gain



Example – typical Si NPN Transistor



$N_D = 1 \times 10^{19} \text{ cm}^{-3}$ (emitter doping)
 $D_p = 1 \text{ cm}^2/\text{s}$ (hole diffusion constant)
 $L_p = 0.5 \mu\text{m}$ (diffusion length)

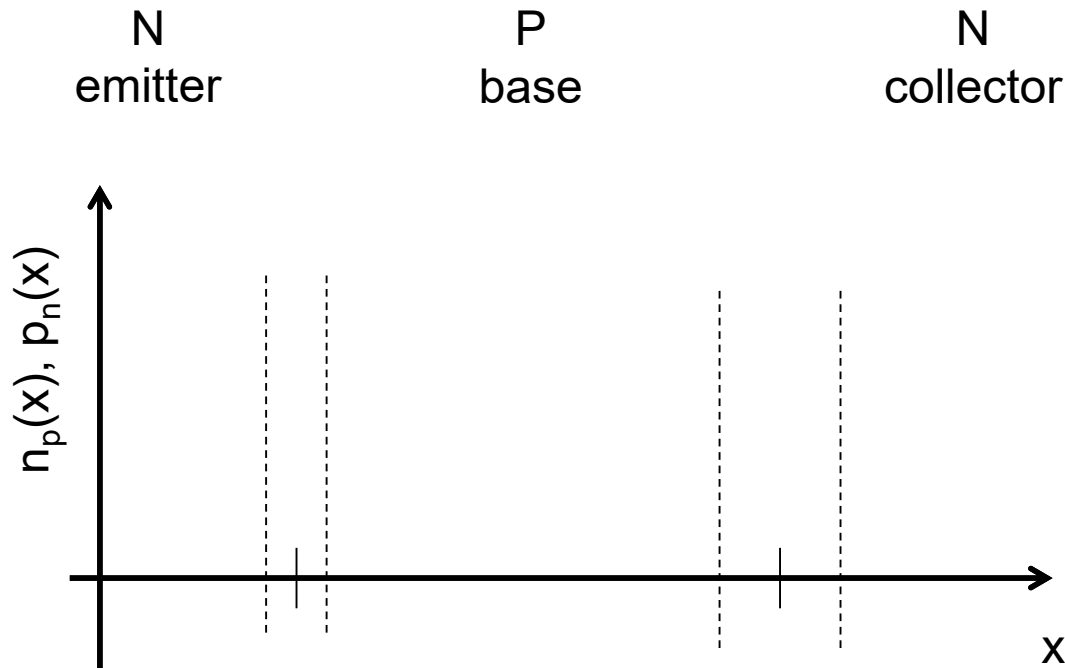
$N_A = 2 \times 10^{17} \text{ cm}^{-3}$ (base doping)
 $D_n = 15 \text{ cm}^2/\text{s}$ (electron diffusion constant)
 $\tau_b = 1 \mu\text{s}$ (minority carrier lifetime)

Calculate β and I_S

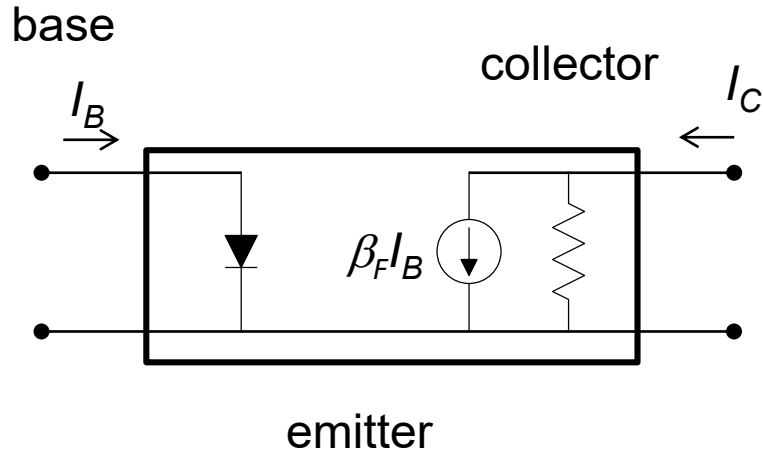
2 min exercise – Early effect (4.2.3)

Ideally I_C should not increase with V_{CE} , however the width of a pn-junction depends on applied voltage!

1. How does the base-collector depletion region change for increasing V_{CE} ?
2. Sketch the minority carrier distribution in the base with V_{CE} applied?
3. How does I_C change with V_{CE} in this case?

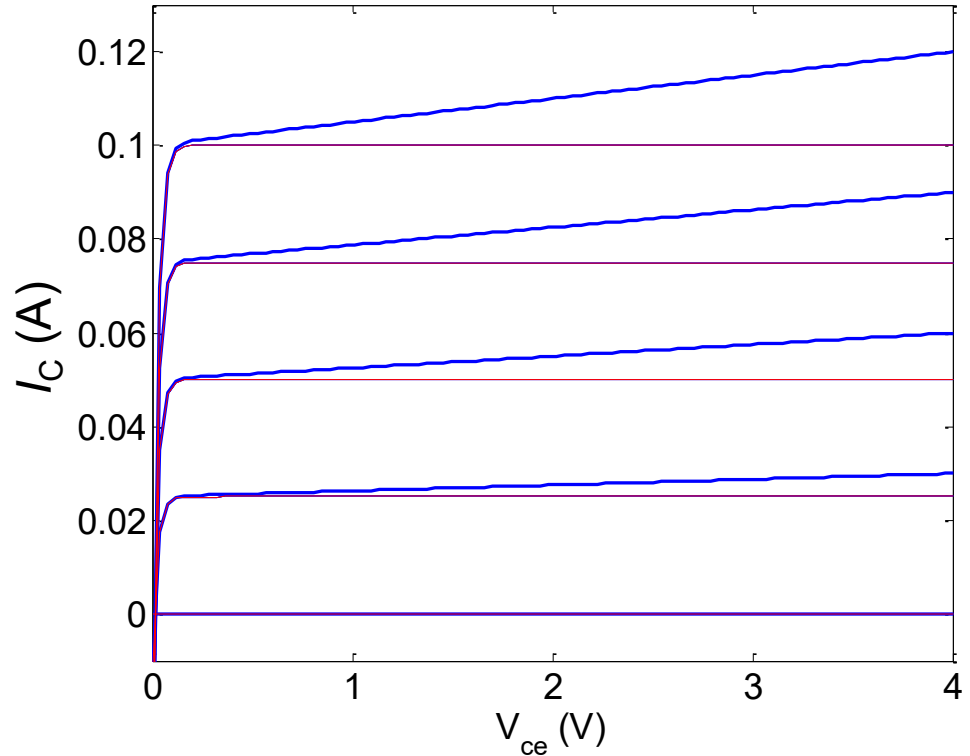


Large Signal Model – Early Effect (4.2.3)



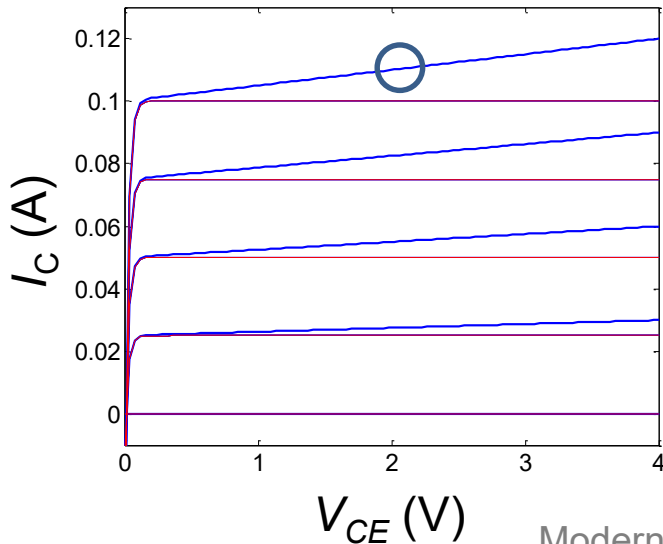
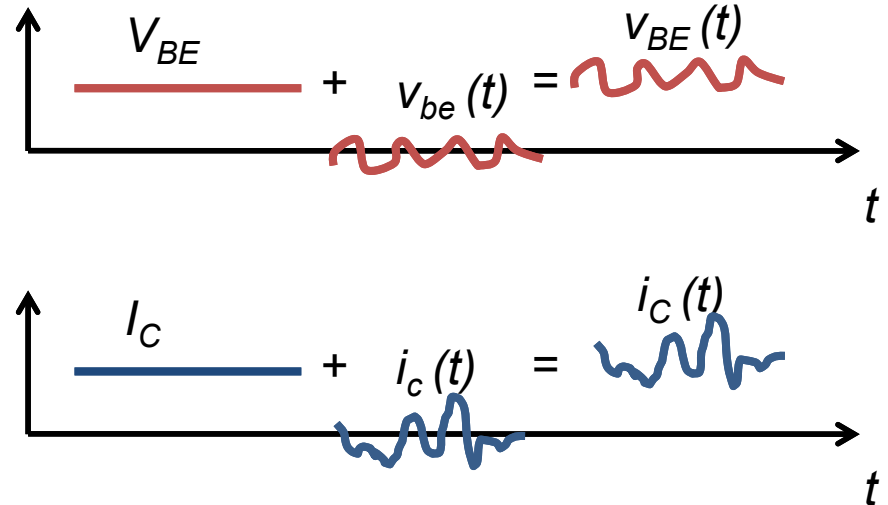
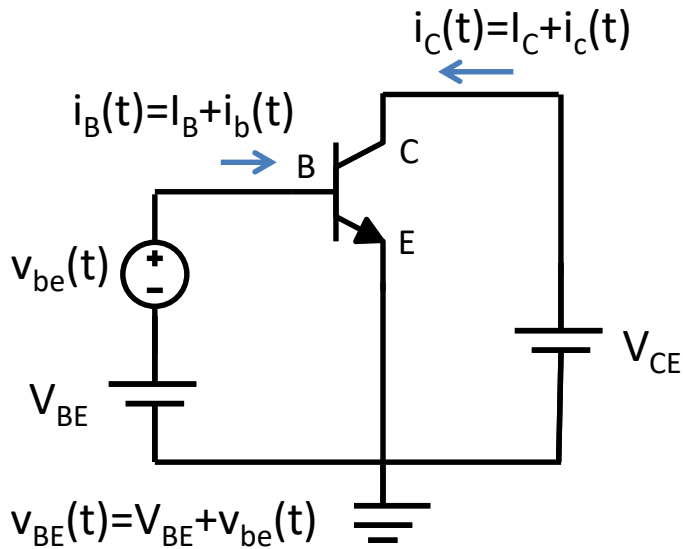
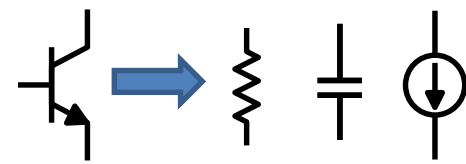
$$I_C = I_S \left(1 + \frac{V_{CE}}{V_A} \right) e^{\frac{V_{BE}}{V_T}}$$

$$I_B = \frac{I_C}{\beta_F} = \frac{I_S \left(1 + \frac{V_{CE}}{V_A} \right) e^{\frac{V_{BE}}{V_T}}}{\beta_F}$$



V_A – Early voltage
15-100V

Small Signals– Taylor Expansion



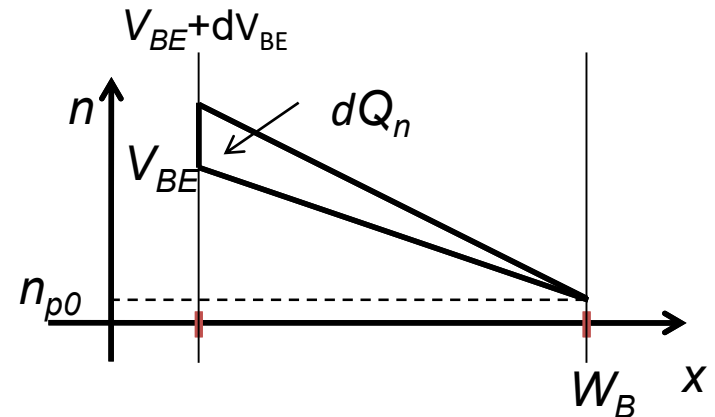
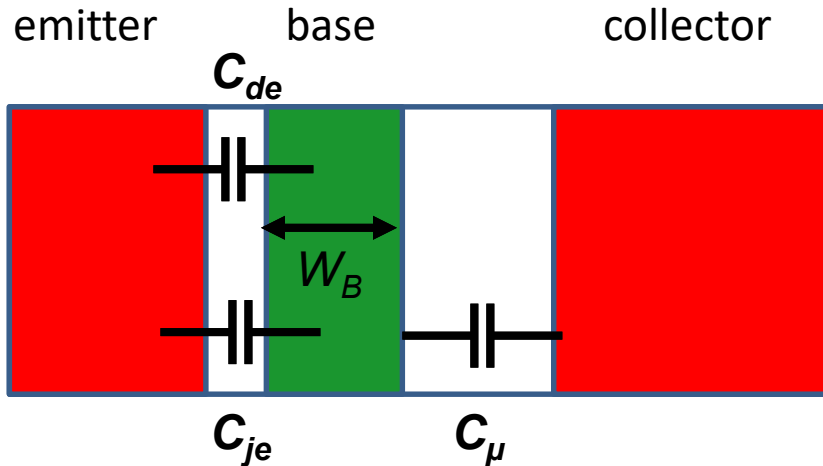
1st order Taylor expansion - linearization

$$f(x_0 + \delta x) \approx f(x_0) + \left. \frac{df(x)}{dx} \right|_{x=x_0} \delta x + \dots$$

$$i_C(V_{BE} + v_{be}) \approx i_C(V_{BE}) + g_m \cdot v_{be}$$

$$i_C(V_{CE} + v_{ce}) \approx i_C(V_{CE}) + \frac{1}{r_o} \cdot v_{ce}$$

capacitances: C_{μ} , C_{je} , C_{de} (9.2.2 (partially))



C_{je} , C_{μ} : junction capacitances (small).

C_{de} : diffusion (base charging) capacitance (only forward biased pn-junction). Change in V_{BE} give change in charge in base (Q_n) -> capacitance

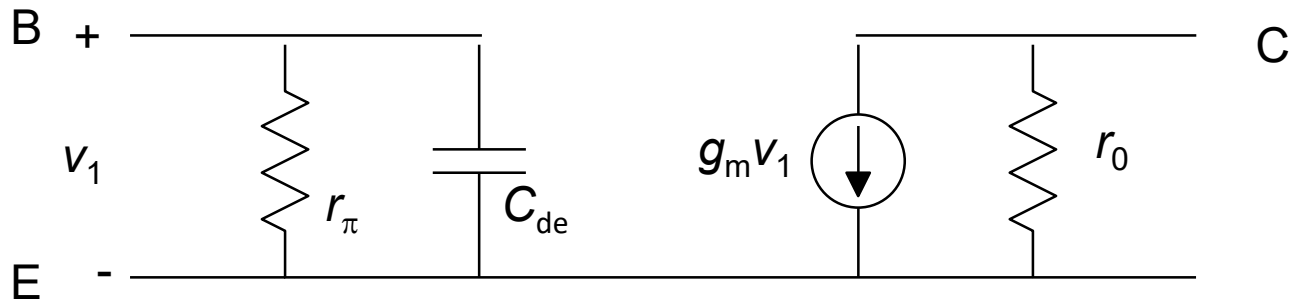
Capacitances become important for high-frequencies

$$C_{de} = \frac{dQ_n}{dv_{BE}} = \tau_F \frac{di_C}{dv_{BE}} = \tau_F g_m$$

τ_F : base transit time, average time for a carrier to cross base

$$\tau_F = W_B^2 / 2D_n$$

(simple) small Signal Model – Active Mode



Transconductance – controls the current source

$$g_m = \frac{I_C}{V_T}$$

Remember:
 $V_T = kT/q$

Input resistance – I_B change with V_{BE}

$$r_\pi = \frac{\beta_F}{g_m}$$

Output resistance – Early effect

$$r_o = \frac{V_A}{I_C} = \frac{V_A}{g_m V_T}$$

Diffusion (base charging) capacitance – forward biased BE junction

$$C_{de} = \tau_F g_m$$

Example – low f model

A BJT is biased so that $I_C = 5\text{mA}$.

Low frequency \rightarrow capacitances are “open circuit”

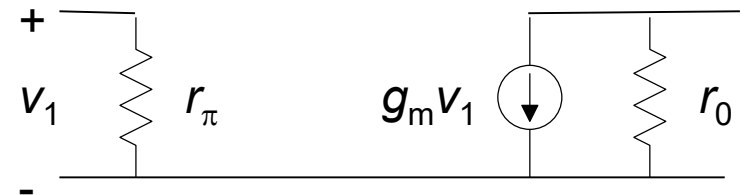
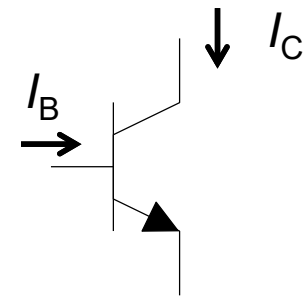
Parameters:

$$\beta = 500$$

$$V_A = 100\text{ V}$$

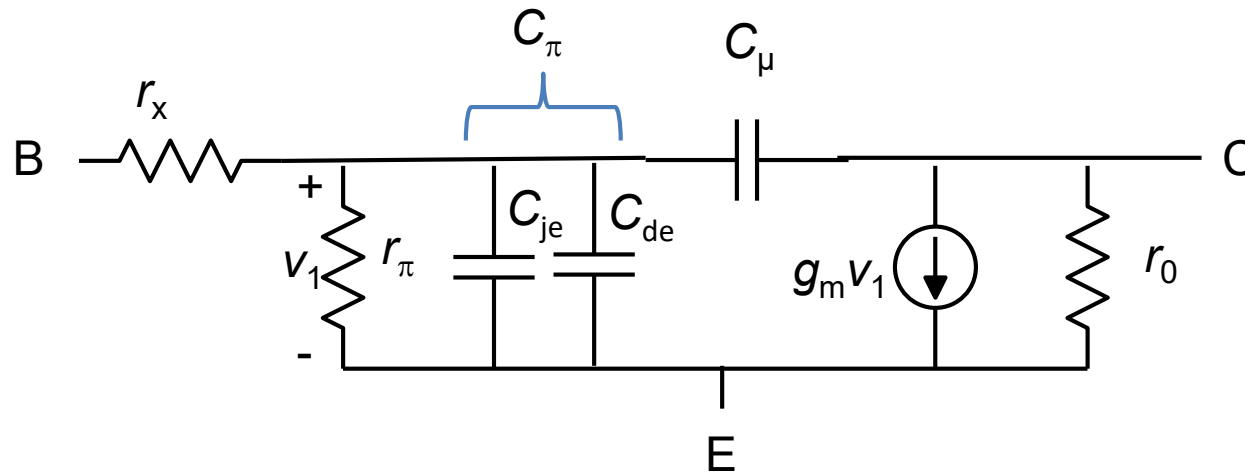
$$V_T = 25.9\text{ mV}$$

Calculate I_B and the corresponding small signal model.



Small Signal Model – more advanced

- Add junction capacitances C_μ (BC junction) and C_{je} (BE junction).
- Sum capacitances $C_\pi = C_{je} + C_{de}$.
- Add series resistances in base (r_x).



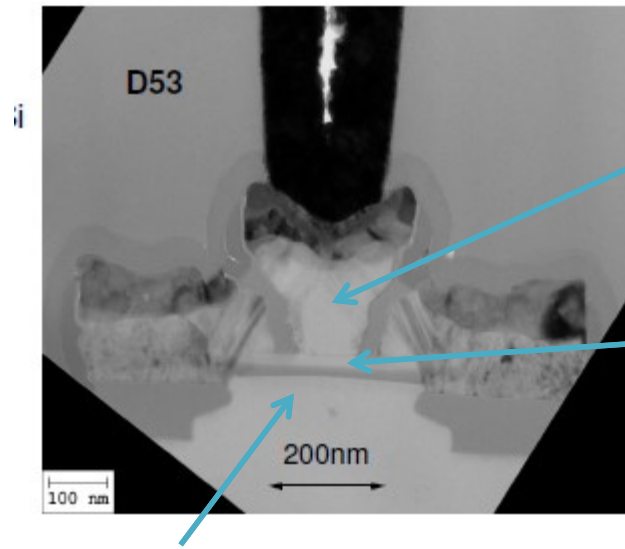
r_x , C_{je} and C_μ : Depends on exact transistor geometry.

Summary - BJTs

- NPN or PNP. NPN is faster due to higher electron mobility
- Active mode (used for amplifiers):
 - Emitter-base junction is forward biased (injects minority carriers into base)
 - Collector-base junction is reverse biased (removes minority carriers from base)
- Early effect: increasing V_{CE} extends the collector-base depletion region narrowing the base resulting in a higher diffusion current I_C .
- Small signal model:
 - Input resistance (r_{π}) due change in base current (I_B) when changing V_{BE} .
 - Output resistance (r_o) due to Early effect.
 - Base charging capacitance (C_{de}) due to change in charge in base region when changing V_{BE} .
 - Can add more capacitances and resistances to get more accurate (and complicated) model.

State-of-the Art: SiGe Bipolar Transistor

- Vertical device
- Not symmetric – high emitter doping, lower in collector
- Heterostructure with graded E_g to enable high doping in base (low resistance) without backinjection into emitter.



Emitter

Base

Collector

$V_{BR} \sim 1.5V$
 $V_{ce,sat} \sim 0.3V$
 $\beta_F \sim 600$
 $f_t = 300 \text{ GHz}$

