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A lonescu and H Riel Nature 479, 2011, p 329 "Tunnel field-effect transistors..."

III-V/Si Nanowire Implementation:

K Tomioka et al Nano Letters, 13, 2013 p.58222 " Sub 60 mV/decade Switch..."

Performance Comparison:

S Koswatta et al IEEE Trans. Electron Devices 56, 2009, p. 456 "Performance comparison ..."

III-V Nanowire TFETs:

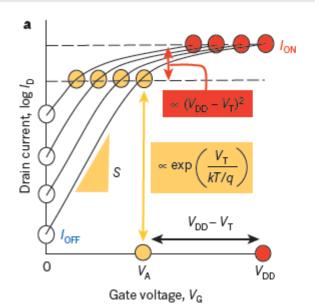
E. Memisevice et al Nano Letters, 17, 2017 p.4373 "Individual defects..."

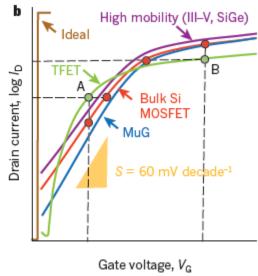


Tunnel field-effect transistors as energy-efficient electronic switches

Adrian M. Ionescu1 & Heike Rie12

Power dissipation is a fundamental problem for nanoelectronic circuits. Scaling the supply voltage reduces the energy needed for switching, but the field-effect transistors (FETs) in today's integrated circuits require at least 60 mV of gate voltage to increase the current by one order of magnitude at room temperature. Tunnel FETs avoid this limit by using quantum-mechanical band-to-band tunnelling, rather than thermal injection, to inject charge carriers into the device channel. Tunnel FETs based on ultrathin semiconducting films or nanowires could achieve a 100-fold power reduction over complementary metal-oxide-semiconductor (CMOS) transistors, so integrating tunnel FETs with CMOS technology could improve low-power integrated circuits.





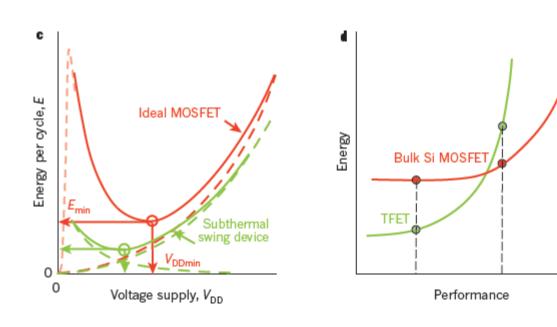
$$S = \underbrace{\frac{dV_G}{d\Psi_s}}_{m} \underbrace{\frac{d\Psi_s}{d(\log_{10} I_D)}} \cong \left(1 + \frac{C_d}{C_{ox}}\right) \ln 10 \frac{kT}{q}$$

$$\rightarrow \underbrace{kT}_{q} \ln 10 \cong 60 \text{ mV decade}^{-1} \mid T = 300 \text{ K}$$

Two a:pproaches to reduce Vdd:

- Reduce Vt to maintain (Vdd-Vt)
- Reduce S

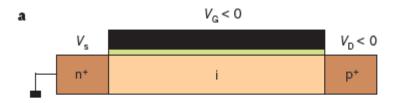
Energy Required for Switching

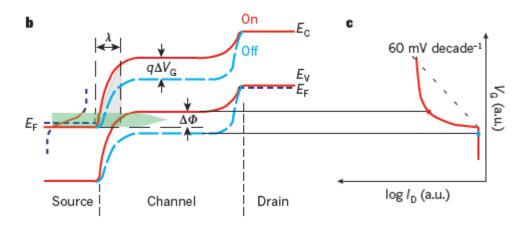


$$\begin{split} E_{\text{total}} &= E_{\text{dynamic}} + E_{\text{leakage}} = \alpha L_{\text{d}} C V_{\text{DD}}^2 + L_{\text{d}} I_{\text{OFF}} V_{\text{DD}} \tau_{\text{delay}} \\ &\approx \alpha L_{\text{d}} C V_{\text{DD}}^2 = L_{\text{d}} C V_{\text{DD}}^2 \frac{I_{\text{OFF}}}{I_{\text{ON}}} = L_{\text{d}} C V_{\text{DD}}^2 \left(\alpha + \frac{I_{\text{OFF}}}{I_{\text{ON}}}\right) \\ &\approx L_{\text{d}} C V_{\text{DD}}^2 \left(\alpha + 10^{\frac{-V_{\text{BO}}}{S}}\right) \end{split}$$

$$P = \alpha L_d C V_{DD}^2 f = I_{OFF} V_{DD} \approx K C V_{DD} = I_{OFF} V_{DD}^3$$

Principle of Operation





-On-state:

Carriers may tunnel from the source to the channel

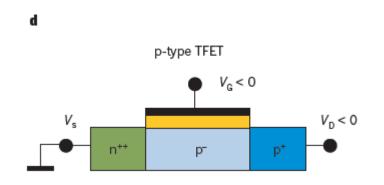
-Off-state:

Tunneling from the source is restricted due to the band gap

Tunnel injection across triangular barrier

$$T_{ ext{WKB}} pprox \exp\!\left(-rac{4\lambda\sqrt{2m^*}\!\sqrt{E_{ ext{g}}^3}}{3q\hbar\left(E_{ ext{g}}+\Delta\Phi
ight)}
ight)$$

pFET Implementation

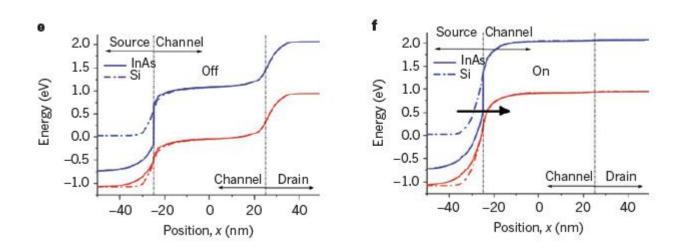


pFET operation:

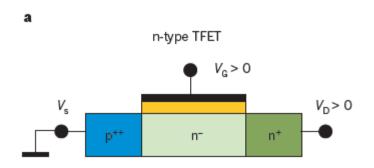
Gate is used to lift the bands in the gate region

pFET booster:

Use n++ InAs in the source



nFET Implementation



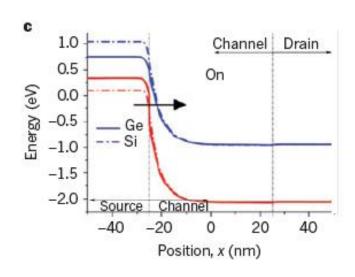
b Channel | Drain 1.0 0.5 Off Energy (eV) 0.0 Conduction -0.5Ge band -1.0-1.5Valence Source Channel band -2.0-20 20 -4040 0 Position, x (nm)

nFET operation:

Gate is used to lower the bands In the gate region

nFET booster:

Use p++ Ge in the source



Materials selection

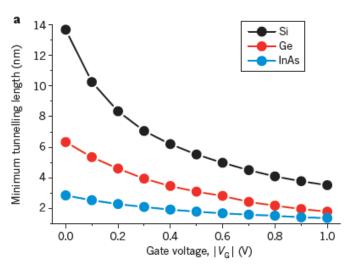
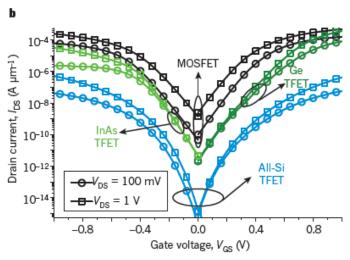


Figure 4 | Importance of the material system on TFET performance. a, Modulation of the minimum screening tunnelling length with the applied gate voltage in all-silicon (black), Ge-source (red) and InAs-source (blue) TFETs, showing the beneficial effect of a higher tunnelling rate due to the shorter tunnelling length in a heterostructure TFET with a low bandgap source material compared with silicon. By contrast, a higher ratio between



the tunnelling length in the off and the on state reflects an improved $I_{\rm ON}/I_{\rm OFF}$ b, Corresponding transfer characteristics of a state-of-the-art 65-nm CMOS transistor (black), complementary Ge/InAs TFET (green) and complementary all-Si TFET (blue). The complementary Ge/InAs TFET achieves the best trade-off between a low $I_{\rm OFF}$, a steep subthreshold swing and performance. $I_{\rm DS}$, drain-to-source current; $V_{\rm DS}$ drain-to-source voltage; $V_{\rm GS}$ gate voltage at source.

Important to reduce the tunneling barrier. Use narrow band gap material or heterostructure design.

$$T_{\text{WKB}} \approx \exp \left(-\frac{4\lambda\sqrt{2m^*}\sqrt{E_{\text{g}}^3}}{3q\hbar\left(E_{\text{g}} + \Delta\Phi\right)}\right)$$



Sub 60 mV/decade Switch Using an InAs Nanowire—Si Heterojunction and Turn-on Voltage Shift with a Pulsed Doping Technique

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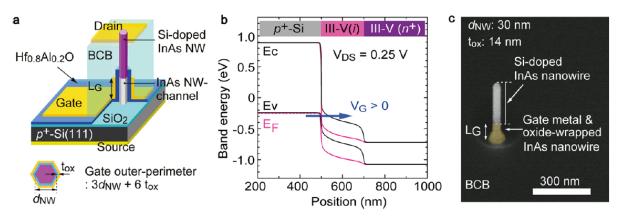


Figure 1. (a) Vertical transistor using InAs NW/Si heterojunction. Surrounding-gate configuration of InAs NW/Si heterojunction. Each InAs NW has a Si-doped InAs/undoped InAs stacked junction with a diameter of 30 nm. At the heterojunction, the diameter of the heterojunction is <30 nm because of isotropic wet etching of SiO₂. The NW is wrapped with $Hf_{0.8}Al_{0.2}$ gate oxide with a thickness of 14 nm and tungsten gate metal. The drain metal is a Ti/Al/Ti/Au multilayer. The gate length is 200 nm. The outer perimeter of the gate is 174 nm. (b) Band diagram simulated by using the one-dimensional Poisson—Schrödinger equation. Under $V_{DS} > 0$, positive V_{G} induces Zener tunneling transport from p-Si to the III—V NW. (c) Representative SEM image showing the vertical transistor.

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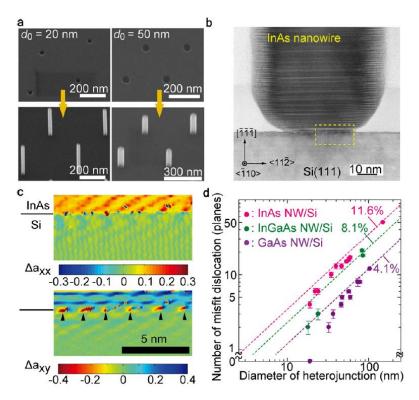


Figure 2. (a) Representative SEM images showing selective-area growth of vertical InAs NWs on Si. The NW has a hexagonal pillar shape surrounded by six-sided $\{-110\}$ planes and a (111)B top surface. (b) Representative TEM image of the heterointerface of InAs NW/Si. The incidence of electron-beam is the <-110> direction. The heterojunction has atomically flat interface. (c) Displacement of latticeconstant for xx-direction (Δa_{xx}) and xy-direction (Δa_{xy}) mappings estimated from a filtered image of dashed squared part in panel b. xx is [-1-12] direction, and xy is [001] direction. (d) Numbers of misfit dislocation at III–V NW/Si heterojunction with a variation of diameter of heterointerface. Dashed lines are calculated values from the lattice mismatch. The close squares are the experimental data counted from the mapping and the TEM images.

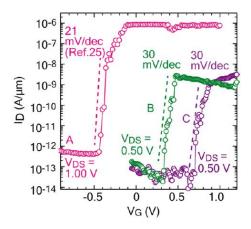


Figure 3. (a) Transfer characteristic of a steep-SS transistor using an InAs NW/Si heterojunction. The pink curve (A)²⁵ is the transfer characteristic of Figure 1 at $V_{\rm DS}=1.00$ V. Curves labeled B (as green curve) and C (as purple curve) show the steep-SS characteristic at $V_{\rm DS}=0.50$ V using an InAs NW/Si heterojunction with a Zn pulse doped InAs NW-channel segment with a 1 s pulse at 29 s intervals and Zn-pulse doped InAs NW-channel with a 2 s pulse at 28 s intervals.

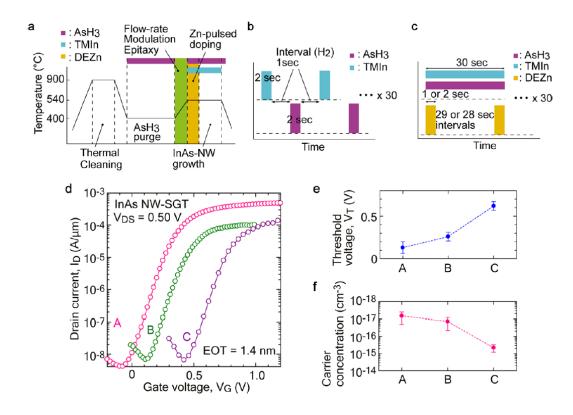


Figure 4. (a) Growth sequence for aligning vertical InAs NWs on Si with the Zn-pulse doping technique. (b) Schematic diagram of flow-rate modulation epitaxy. TMIn (2 s) and AsH_3 (2 s) are alternately supplied with an interval of H_2 (1 s). This sequence is repeated 30 times. (c) Diagram of Zn-pulse doping technique. DEZn (1 or 2 s) are alternately supplied with an interval of 29 or 28 s. This sequence is repeated six times to make pseudointrinsic InAs channel region. (d) The curve labeled A (pink plots) is the transfer characteristic of InAs NW-SGT with an undoped segment. Curves labeled B and C show transfer properties of InAs NW-SGTs with a Zn-pulse doped InAs NW-channel segment with a 1 s pulse with 29 s intervals (green plots) and with 2 s pulse with 28 s intervals (purple plots). (e) Variation of V_T of an InAs NW vertical SGT with a different InAs NW-channel segment. (f) Variation of N_D estimated from the V_T .

Performance Comparison Between p-i-n Tunneling Transistors and Conventional MOSFETs

Siyuranga O. Koswatta, Mark S. Lundstrom, Fellow, IEEE, and Dmitri E. Nikonov, Senior Member, IEEE

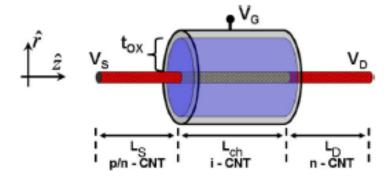
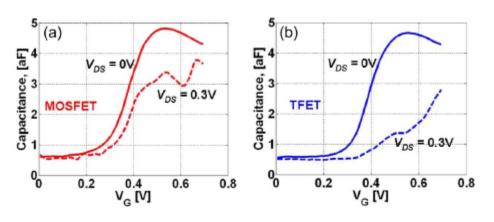


Fig. 1. Modeled device geometry used in this paper with cylindrically symmetric wrap-around gate electrode (see text for device parameters). The high-k oxide is removed from source/drain regions in order to reduce the fringing fields that adversely affect the drive current for the p-i-n TFET.

Nanotube used for the modelling

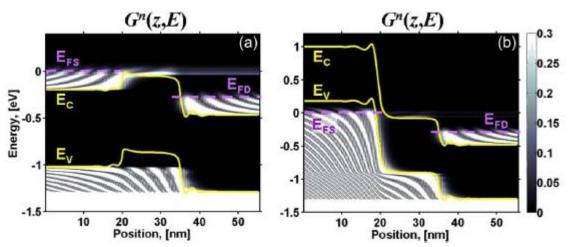
15 nm gate length

Simulated Data



Reduced capacitance

Fig. 2. Total gate capacitance (C_{tot}) versus V_{GS} comparison for (a) MOSFET and (b) TFET under dissipative transport. At small V_{DS} , both devices show similar characteristics. However, at larger V_{DS} , a fundamentally different behavior is observed; for the TFET device, capacitance remains small until larger gate biases are applied.



Less carriers in the channel

12

Fig. 3. Energy-position-resolved electron distribution [i.e., diagonal elements of G^n in (5)] for (a) MOSFET and (b) TFET under ballistic transport at $V_{GS} = 0.5 \text{ V}$, $V_{DS} = 0.3 \text{ V}$. A significantly higher occupation of channel states is observed for the former.

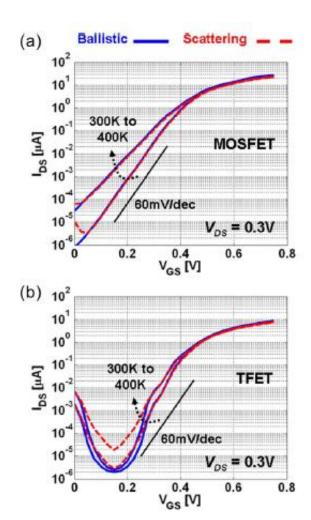


Fig. 5. $I_{\rm DS}$ – $V_{\rm GS}$ dependence on temperature for (a) MOSFET and (b) TFET under ballistic and dissipative transport. The latter has reduced temperature dependence under ballistic conditions. Phonon-assisted tunneling can, however, degrade the subthreshold characteristics.

Comparison MOSFETs and TFETs

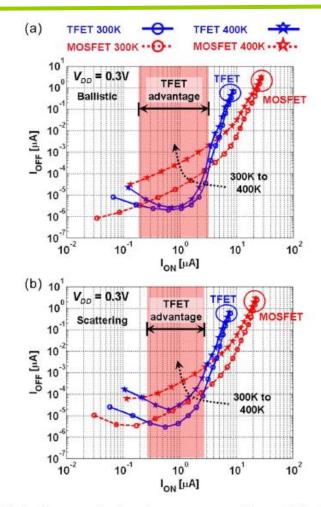
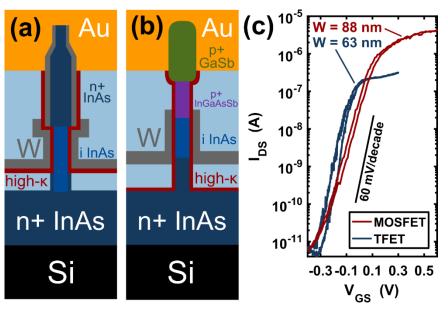
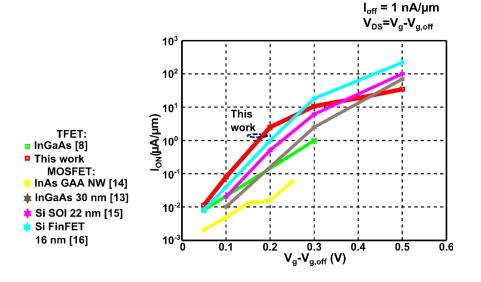


Fig. 6. $I_{\rm OFF}$ versus $I_{\rm ON}$ dependence on temperature at $V_{\rm DD}=0.3$ V under (a) ballistic and (b) dissipative transport. Shaded region is where the TFET has an advantage over the MOSFET due to larger $I_{\rm ON}$ with a smaller $I_{\rm OFF}$. Temperature dependence of $I_{\rm OFF}$ for the TFET is also smaller than that for the latter.

TFET Characteristics







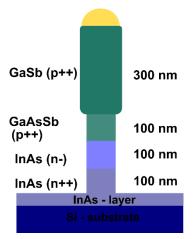
TFET benefits

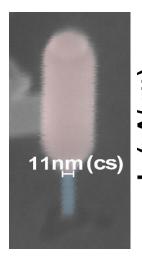
- Higher drive current at 0.2-0.3 V overdrive
- Option for power saving!

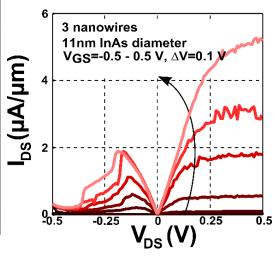
TFET Process



- VLS growth combined with top-down processing
- Digital etching
 - ozone surface oxidation
 - citric acid oxide etching





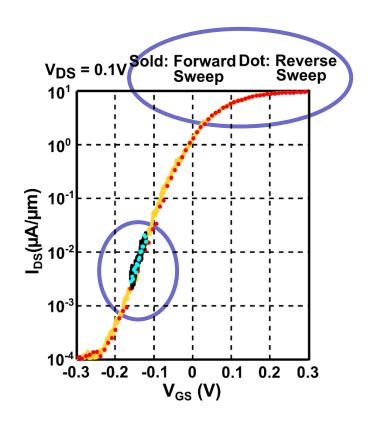


- Diameter scaling down below 10 nm
- Typical diameter 20 nm
- No etching of GaSb is observed

Hysteresis/Measurement Reliability



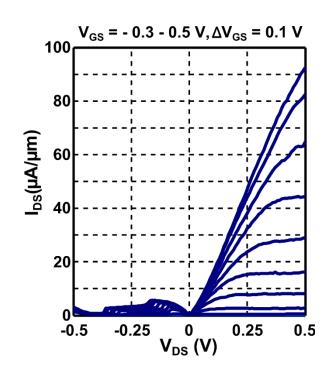
- Sweep in forward and backward direction
- Using different voltage ranges
- Small hystersis of 5.4 mV
- Subthreshold swing independent of sweep direction and range

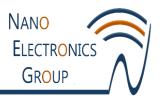


Output Characteristics



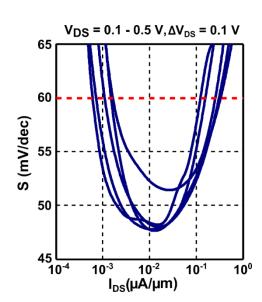
- Strong NDR in reverse bias with PVR 14.8
- High quality junction
- I_{DS} = 92 µA/µm at V_{DS} = V_{GS} = 0.5 V
- Weak superlinear behaviour

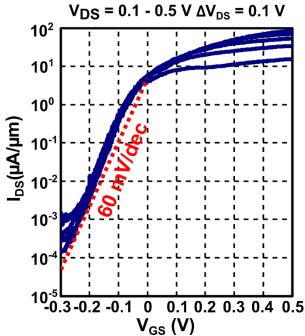




Transfer Characteristics

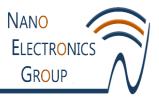
- Sub 60 mV/decade operation at I_{DS} ~ 1-300 nA/μm
- Good electrostatic control (DIBL 25 mV/V), 5 mV Hysteresis
- Gate-current < I_{DS}/100
- S_{min} = 48 mV/dec at 0.3 V
- I₆₀= 0.31 μA/μm at 0.3 V

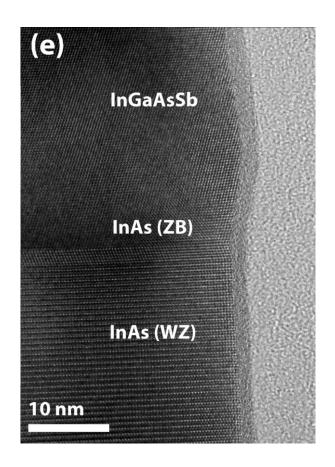


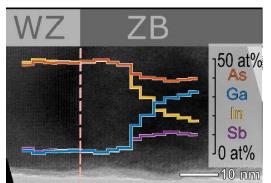


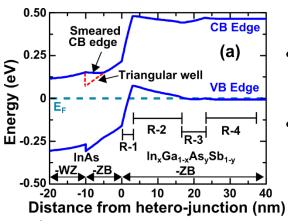
E. Memisevic et al., IEDM 2016

Materials Characterization







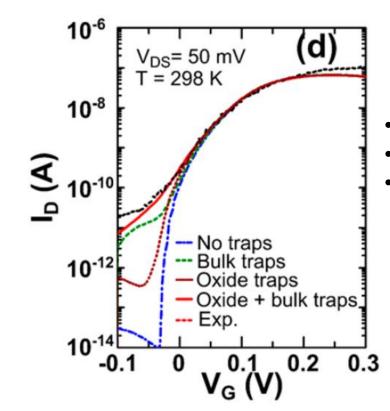


- Nanowire with WZ-ZB
 InAs transition
- Slowly varying composition gradient with transition over about 20 nm
 - Strong strain field within heterostructure (2-3%)
- 28% Sb in InGaAsSb

E. Memisevic et al., Nano Lett 2017

Where are the Defects?





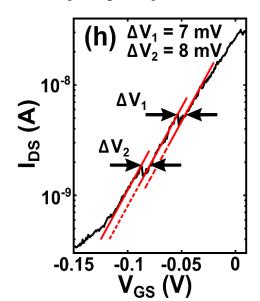
- Excellent fit by TCAD modeling (ETH)
- Bulk trap main contribution to leakage current
- Oxide traps have vanishing influence besides electrostatic effect

E. Memisevic et al., Nano Lett 2017

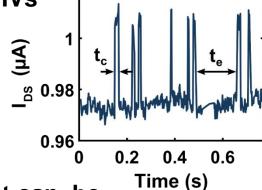
Defects within the Oxide



Multiple jumps detected in some IVs 1.02



E. Memisevic et al., Nano Letters 2017 M. Hellenbrand et al ESSDERC 2017



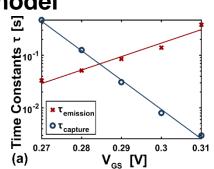
Each event can be

viewed as RTN

Time constants fit with model

of defects within oxide

$$\tau_{c/e} = \tau_0 \exp(z/\lambda)$$

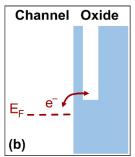


Counts (1000)

0.96

0.98

Current I_{DS} (µA)



1.02