

What have we covered thus far?

- □ Recap: Considered the effect of multipath propagation on the received amplitude and phase, as well as its temporal variations.
- Key assumption: Small system bandwidth (narrowband systems only). As a consequence, multiple directions can not be resolved by the RX and seem that they arrive almost at the same time.
- ☐ Most current and future systems however will leverage large bandwidths.
- ☐ Desirable to describe channel variations over a larger bandwidth range the topic for the current lecture.
 - Wideband characterization of channels and real world examples.

 Wireless Communication
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Propagation Impact on Wideband Systems

Impact interpreted in two different ways:

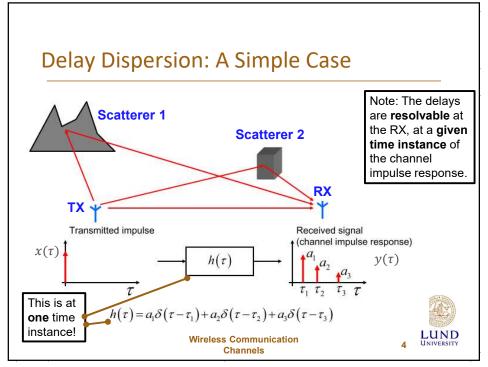
- ☐ The transfer function of the channel **varies** over the **bandwidth** of interest (a.k.a. the **frequency selectivity** of the channel).
- ☐ Impulse response of the channel is **not** a Delta function; the arriving signal has a **longer run time** than the transmitted signal (a.k.a. delay dispersion).

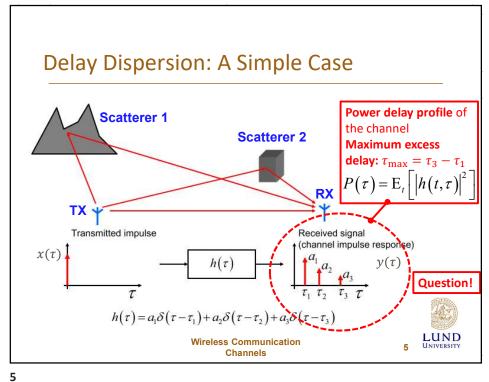
Question: What is the relationship between the above?

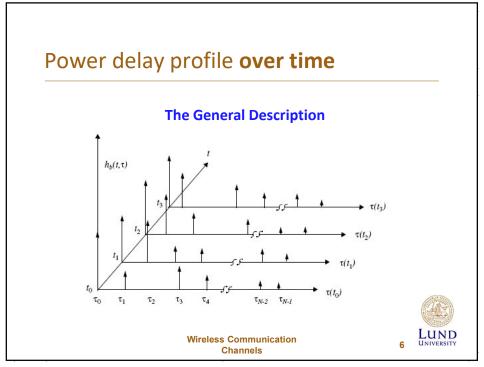


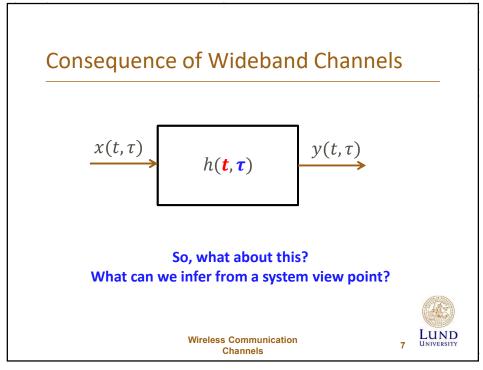
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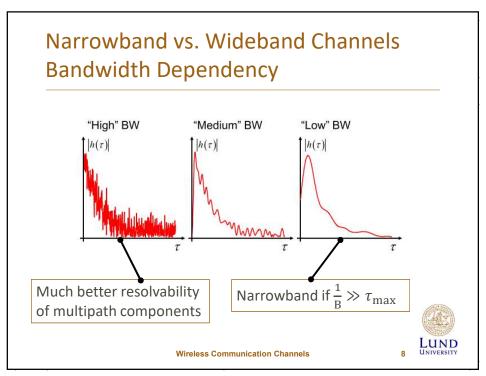
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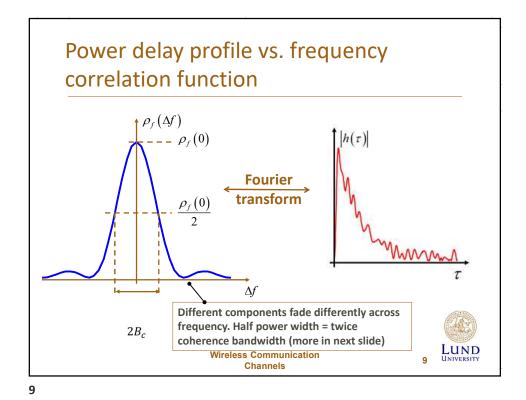












Condensed parameters Coherence bandwidth Given the frequency correlation of a channel, we can define the coherence bandwidth B_{C} : What does the coherence $\rho_f(\Delta f)$ bandwidth tell us? It shows us over how large a bandwidth we can assume so that the channel is fairly constant. Radio systems using a bandwidth much smaller than B_C will not notice the frequency selectivity Δf of the channel. $2B_c$ **Wireless Communication** Channels

Condensed parameters Power delay profile (cont.)

We can infer many useful parameters from the power delay profile Total power (time integrated):

$$P_{m} = \int_{-\infty}^{\infty} P(\tau) d\tau$$

Average mean delay (first moment of the PDP)

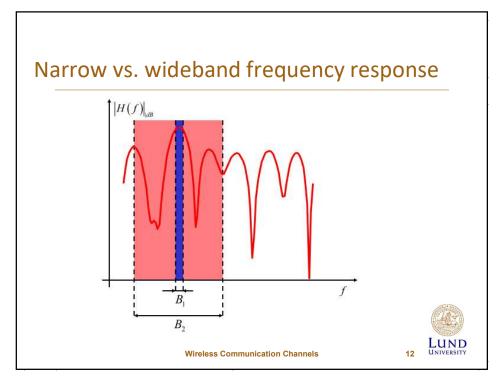
$$T_{m} = \frac{\int_{-\infty}^{\infty} \tau P(\tau) d\tau}{P_{m}}$$

 $T_{\rm m} = \frac{\int_{-\infty}^{\infty} \tau P(\tau) d\tau}{P_{\rm m}}$ Average RMS delay spread (second moment of the PDP)

$$S_{\tau} = \sqrt{\frac{\int_{-\infty}^{\infty} \tau^2 P(\tau) d\tau}{P_m} - T_m^2}$$

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Widely used "rules-of-thumb"

$$Tc \approx \frac{1}{D_s}$$

$$Bc \approx \frac{1}{S_\tau}$$

$$Tc = \frac{9}{16\pi D_s}$$

time over which the time correlation function is above 0.5

$$T_c = \frac{0.423}{D_{\rm S}} \label{eq:Tc}$$
 less restrictive and widely used

$$Bc = \frac{1}{5S_{\tau}}$$

band over which the frequency correlation function is above 0.5

$$Bc = \frac{1}{50S_{\tau}}$$

 $Bc = \frac{1}{50S_{\tau}}$ band over which the frequency correlation function is above 0.9



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