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Wireless Communications Channels

Lecture 3: Fading

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Fading – Statistical description of the wireless channel

- Why statistical description
- Large scale fading
- Small scale fading:
 - without dominant component
 - with dominant component
- Statistical models
- Measurement example

Why “statistical” description?

- ❑ Complex, unknown environment
- ❑ Can not describe everything in detail
 - ❑ Maxwell’s equations far too complex in real scenarios
- ❑ Large variations depending on the TX, RX and interacting object locations
- ❑ Need a statistical measure since we can not describe every point everywhere

“There is a x% probability that the amplitude/power will be above the level y”

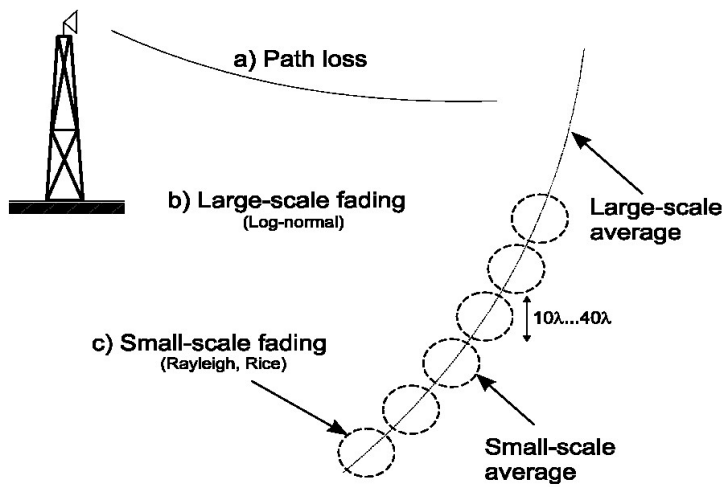
The WSSUS model: Assumptions

A very common channel model is the WSSUS-model: **Statistical properties remain the same over the considered time (or area)**

Recall: the channel is composed of a number of different contributions (incoming waves), the following is assumed:

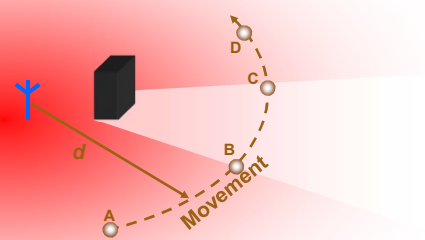
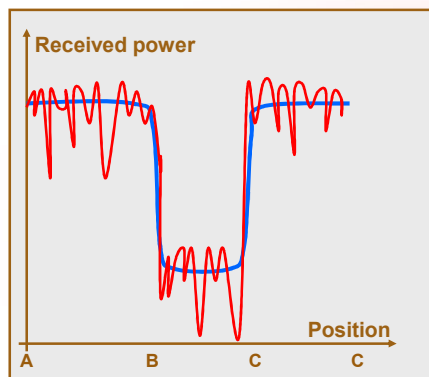
- ❑ The channel is **Wide-Sense Stationary (WSS)**, meaning that the **correlation** of the channel is **invariant** over time.
- ❑ The channel is built up by **Uncorrelated Scatterers (US)**, meaning that the **frequency** correlation of the channels is **invariant** over frequency. (Contributions with different delays are uncorrelated.)

What is large scale and small scale?



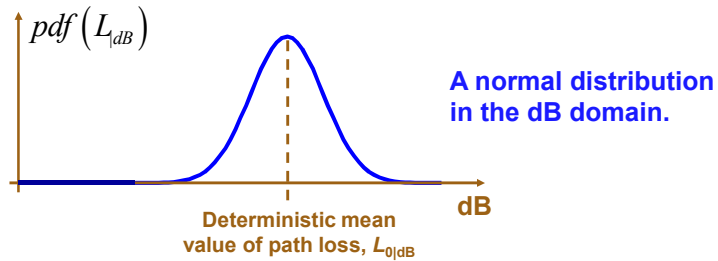
Wireless Communication Channels

Large-scale fading: Basic principle



Large-scale fading: Log normal distribution

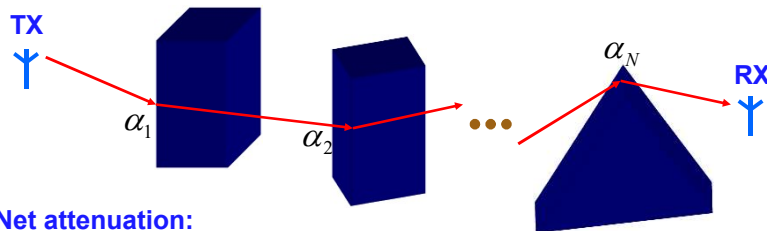
Confirmed by propagation channel measurements over the past 50 years.



$$pdf(L_{dB}) = \frac{1}{\sqrt{2\pi}\sigma_{F|dB}} \exp\left(-\frac{(L_{dB} - L_{0,dB})^2}{2\sigma_{F|dB}^2}\right)$$

Large-scale fading: Why log-normal?

Many diffraction points adding extra attenuation to the pathloss. This is, however, only one of several possible explanations.



Net attenuation:

$$L_{tot} = L(d) \times \alpha_1 \times \alpha_2 \times \dots \times \alpha_N$$

$$L_{tot|dB} = L(d)_{dB} + \alpha_{1|dB} + \alpha_{2|dB} + \dots + \alpha_{N|dB}$$

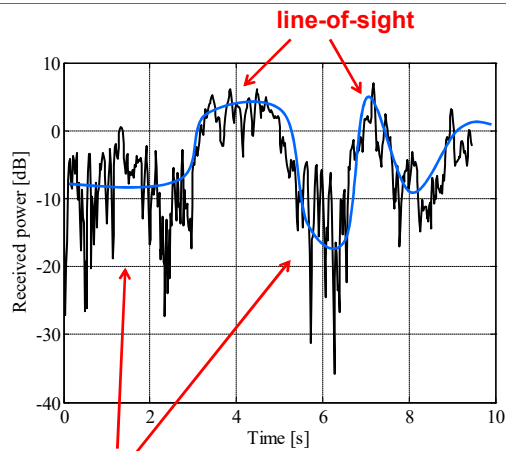
If these are considered random and independent, we should get a normal distribution in the dB domain.

Example: Shadowing from people



Two persons communicating with each other using cell phones, signal sometimes blocked by randomly moving humans

Example: Shadowing from humans



obstructed LOS

Small-scale fading: Two waves

Wave 1

$$E_{RX1}(t) = A_1 \cos(2\pi f_c t - 2\pi/\lambda * d_1)$$

$$k_0 = 2\pi/\lambda$$

$$\vec{E} = E_1 \exp(-jk_0 d_1)$$

Wave 2

$$E_{RX2}(t) = A_2 \cos(2\pi f_c t - 2\pi/\lambda * d_2)$$

$$E = E_2 \exp(-jk_0 d_2)$$

TX: $E_{TX}(t) = A \cos(2\pi f_c t)$

RX

Vector diagram showing E_1 and E_2 vectors.

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Small-scale fading: Two waves

Wave 1

Wave 2

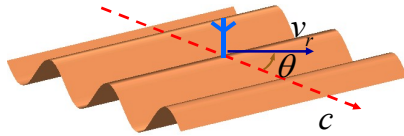
Wave 1 + Wave 2

TX

RX

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Small-scale fading: Doppler shifts



Receiving antenna moves with speed v_r at an angle θ relative to the propagation direction of the incoming wave, which has frequency f_0 .

Frequency of received signal:

$$f = f_0 + \nu$$

where the doppler shift is

$$\nu = -f_0 \frac{v_r}{c} \cos(\theta)$$

The maximal Doppler shift is

$$\nu_{\max} = f_0 \frac{v}{c}$$

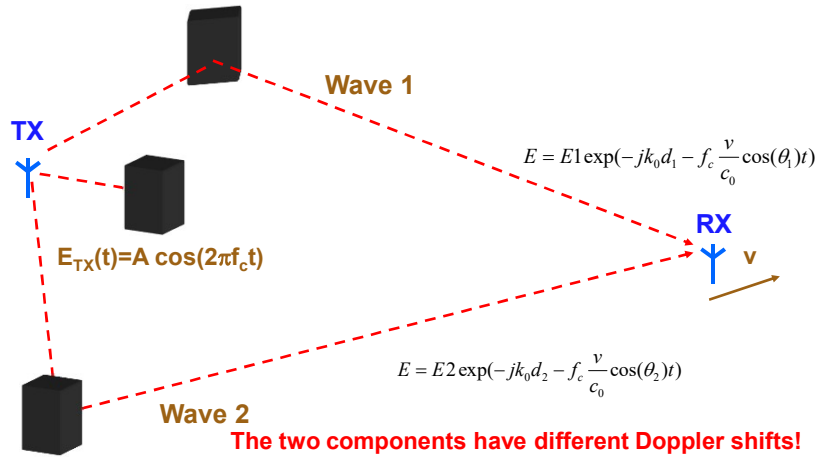
Doppler shift

How large is the maximum Doppler frequency at pedestrian speeds for 5.2 GHz WLAN and at highway speeds using GSM 900?

$$\nu_{\max} = f_0 \frac{v}{c}$$

- $f_0=5.2 \cdot 10^9$ Hz, $v=5$ km/h, (1.4 m/s) \Rightarrow 24 Hz
- $f_0=900 \cdot 10^6$ Hz, $v=110$ km/h, (30.6 m/s) \Rightarrow 92 Hz

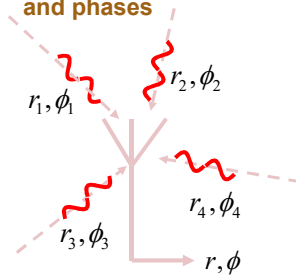
Small-scale fading Two waves with Doppler



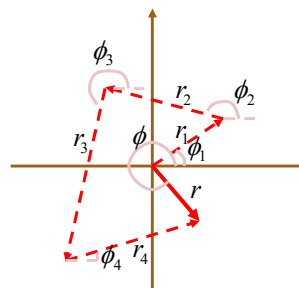
**The two components have different Doppler shifts!
The Doppler shifts will cause a random frequency modulation**

Small-scale fading Many incoming waves

Many incoming waves with independent amplitudes and phases



Add them up as phasors



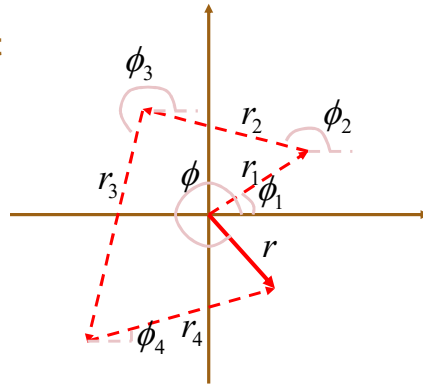
$$r \exp(j\phi) = r_1 \exp(j\phi_1) + r_2 \exp(j\phi_2) + r_3 \exp(j\phi_3) + r_4 \exp(j\phi_4)$$

Small-scale fading Many incoming waves

Re and Im components are sums of many independent equally distributed components

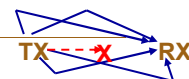
$$\text{Re}(r) \in N(0, \sigma^2)$$

$\text{Re}(r)$ and $\text{Im}(r)$ are independent

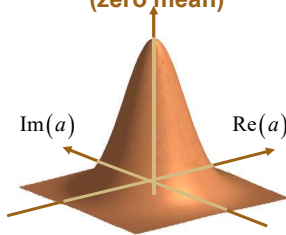


Small-scale fading Rayleigh fading

No dominant component
(non line-of-sight)

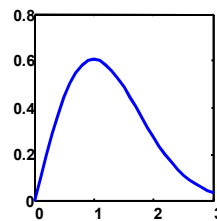


Tap distribution
2D Gaussian
(zero mean)



No line-of-sight component

Amplitude distribution
Rayleigh

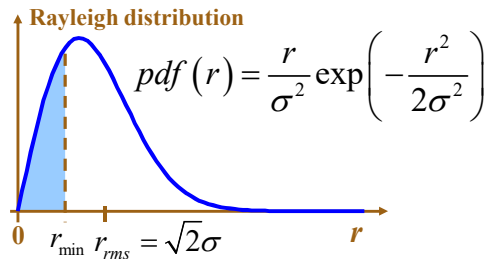


$$pdf(r) = \frac{r}{\sigma^2} \exp\left(-\frac{r^2}{2\sigma^2}\right)$$



Small-scale fading

Rayleigh fading



Probability that the amplitude is below some threshold r_{\min} :

$$\Pr(r < r_{\min}) = \int_0^{r_{\min}} pdf(r) dr = 1 - \exp\left(-\frac{r_{\min}^2}{r_{rms}^2}\right)$$



Small-scale fading

Rayleigh fading – outage probability

- What is the probability that we will receive an amplitude 20 dB below the r_{rms} ?

$$\Pr(r < r_{\min}) = 1 - \exp\left(-\frac{r_{\min}^2}{r_{rms}^2}\right) = 1 - \exp(-0.01) \approx 0.01$$

- What is the probability that we will receive an amplitude below r_{rms} ?

$$\Pr(r < r_{\min}) = 1 - \exp\left(-\frac{r_{\min}^2}{r_{rms}^2}\right) = 1 - \exp(-1) \approx 0.63$$



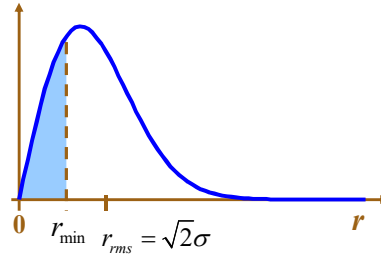
Small-scale fading

Rayleigh fading – fading margin

To ensure that in most cases we **receive** enough power, we transmit **extra** power by including the so-called “fading margin”

$$M = \frac{r_{rms}^2}{r_{min}^2}$$

$$M_{dB} = 10 \log_{10} \left(\frac{r_{rms}^2}{r_{min}^2} \right)$$



Small-scale fading

Rayleigh fading – fading margin

How many dB fading margin, against Rayleigh fading, do we need to obtain an outage probability of 1%?

$$\Pr(r < r_{min}) = 1 - \exp\left(-\frac{r_{min}^2}{r_{rms}^2}\right) = 1\% = 0.01$$

Some manipulation gives

$$1 - 0.01 = \exp\left(-\frac{r_{min}^2}{r_{rms}^2}\right) \Rightarrow \ln(0.99) = -\frac{r_{min}^2}{r_{rms}^2}$$

$$\Rightarrow \frac{r_{min}^2}{r_{rms}^2} = -\ln(0.99) = 0.01 \Rightarrow M = \frac{r_{rms}^2}{r_{min}^2} = 1 / 0.01 = 100$$

$$\Rightarrow M_{dB} = 20$$

Small-scale fading

Rayleigh fading – signal and interference

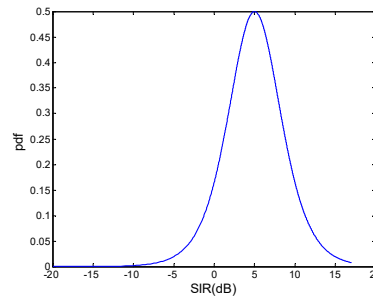
Both the desired signal and the interference undergo fading

For a single user interferer and Rayleigh fading:

$$pdf_{SIR}(r) = \frac{2\bar{\sigma}^2 r}{(\bar{\sigma}^2 + r^2)^2}$$

$$cdf_{SIR}(r) = 1 - \frac{\bar{\sigma}^2}{\bar{\sigma}^2 + r^2}$$

where $\bar{\sigma}^2 = \frac{\sigma_2^2}{\sigma_1^2}$ is the mean signal to interference ratio



pdf for 10 dB mean signal to interference ratio

Small-scale fading

Rayleigh fading – signal and interference

What is the probability that the instantaneous SIR will be below 0 dB if the mean SIR is 10 dB when both the desired signal and the interferer experience Rayleigh fading?

$$\Pr(r < r_{\min}) = 1 - \frac{\bar{\sigma}^2 r_{\min}}{\bar{\sigma}^2 + r_{\min}^2} = 1 - \frac{10}{(10+1)} \approx 0.09$$

Small-scale fading: One dominating component

In case of Line-of-Sight (LOS) one component dominates.

- Assume it is aligned with the real axis

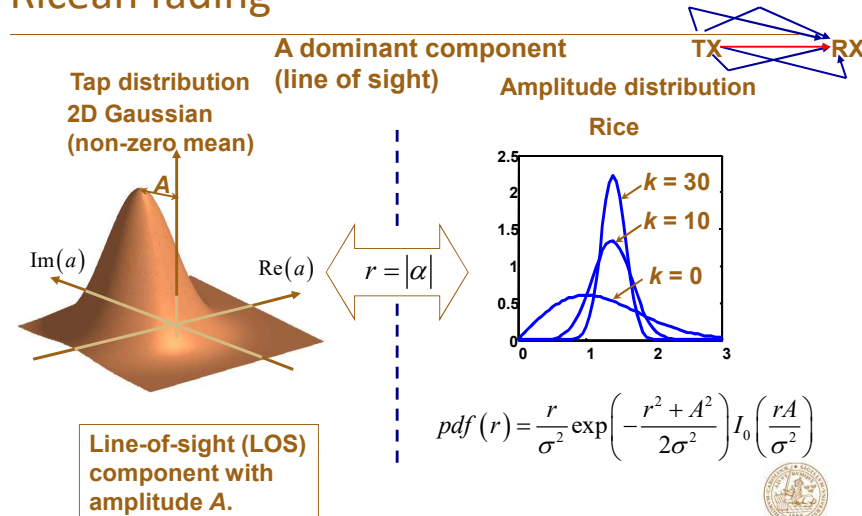
$$\text{Re}(r) \in N(A, \sigma^2) \quad \text{Im}(r) \in N(0, \sigma^2)$$

- The received amplitude has now a Ricean distribution instead of a Rayleigh
 - The fluctuations are smaller
 - The phase is dominated by the LOS component
 - In real cases the mean propagation loss is often smaller due to the LOS
- The ratio between the power of the LOS component and the diffuse components is called Ricean K-factor

$$k = \frac{\text{Power in LOS component}}{\text{Power in random components}} = \frac{A^2}{2\sigma^2}$$

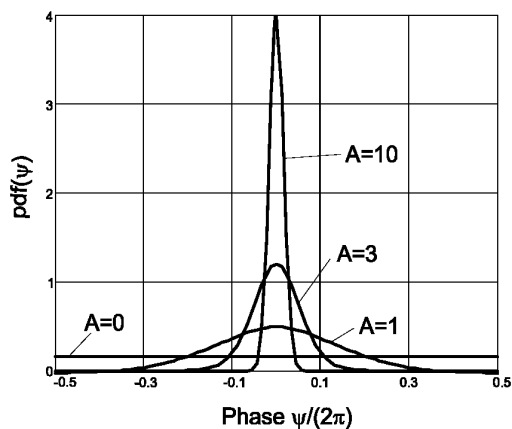


Small-scale fading Ricean fading



Small-scale fading Ricean fading, phase distribution

The distribution of the phase is dependent on the K-factor



Small-scale fading: Nakagami distribution

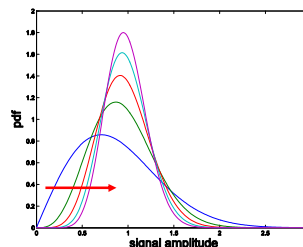
In many cases the received signal can not be described as a pure LOS + diffuse components

The Nakagami distribution is often used in such cases

$$pdf(r) = \frac{1}{\Gamma(m)} \left(\frac{m}{\sigma^2}\right)^m r^{2m-1} \exp\left(-\frac{m}{2\sigma^2} r^2\right)$$

where $\Gamma(m)$ is the Gamma function

$$m = \left(\frac{2\sigma^2}{r^2 - 2\sigma^2}\right)^2$$



increasing m
{1 2 3 4 5}

with m it is possible to adjust the dominating power

Some special cases

Rayleigh fading

$$pdf(r) = \frac{r}{\sigma^2} \exp\left(-\frac{r^2}{2\sigma^2}\right)$$

Rice fading, $K=0$

$$pdf(r) = \frac{r}{\sigma^2} \exp\left(-\frac{r^2+A^2}{2\sigma^2}\right) I_0\left(-\frac{rA}{\sigma^2}\right) = \frac{r}{\sigma^2} \exp\left(-\frac{r^2}{2\sigma^2}\right)$$

Rice fading with $K=0$ becomes Rayleigh

Nakagami, $m=1$

$$pdf(r) = \frac{1}{\Gamma(m)} \left(\frac{m}{\sigma^2}\right)^m r^{2m-1} \exp\left(-\frac{m}{2\sigma^2} r^2\right) = \frac{r}{\sigma^2} \exp\left(-\frac{r^2}{2\sigma^2}\right)$$

Nakagami with $m=1$ becomes Rayleigh



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