

Exam in Digital Communications, EITG05

August 20, 2018

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- During this exam you are allowed to use a calculator, the compendium, a printout of the lecture slides, and Tefyma (or equivalent).
- Please use a new sheet of paper for each solution. Write your anonymized assessment code + a personal identifier on each paper.
- Solutions should clearly show the line of reasoning and follow the methods presented in the course. If you use results from the compendium or lecture slides, please add a reference in your solution.
- ▶ If any data is lacking, make reasonable assumptions.

Good luck!

Determine for each of the five statements below if it is true or false. Give a motivation for each of your answers.

(a) Consider a conventional *M*-ary QAM system, using a pulse $g(t) = g_{rc}(t)$ with duration $T = T_s/2$.

"If the bandwidth efficiency is 4.21 bps/Hz, M = 256, then the 90% definition of bandwidth is used."

- (b) "For 4-ary QAM with equally likely signal alternatives and a triangular pulse $g(t) = g_{tri}(t)$ with amplitude A and duration $6T_s/10$ the average signal power is $2A^2/10$."
- (c) "With uncoded equally likely binary signals the bit error probability is always larger than $2.5 \cdot 10^{-10}$ if E_b/N_0 is 13 dB."
- (d) "8-ary PPM and 8-ary FSK have the same energy efficiency."
- (e) Consider *M*-ary PAM signaling over an AWGN channel without multipath. *"In order to avoid inter-symbol interference (ISI), the symbol rate* R_s *can never be larger than the inverse* 1/T *of the transmit pulse* g(t) *duration* T."

Consider a QAM signal constellation, with rectangular pulse shape of duration $T = T_s$,

$$s_{\ell}(t) = A_{\ell} g_{rec}(t) \cos(2\pi f_c t) - B_{\ell} g_{rec}(t) \sin(2\pi f_c t) , \quad \ell = 0, \dots, 7 ,$$

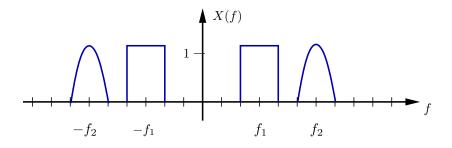
for which the M = 8 possible amplitude pairs are given as follows:

$$(A_0, B_0) = (+2, 0), (A_1, B_1) = (+1, +1), (A_2, B_2) = (0, +2), (A_3, B_3) = (-1, +1),$$

 $(A_4, B_4) = (-2, 0), (A_5, B_5) = (-1, -1), (A_6, B_6) = (0, -2), (A_7, B_7) = (+1, -1).$

- (a) For the carrier frequency $f_c = 2/T_s$, draw the transmit signal s(t) corresponding to the message sequence $\mathbf{m} = (m[0] \ m[1] \ m[2] \ m[3]) = (0 \ 6 \ 2 \ 4)$ within the time interval $0 \le t \le 4 T_s$.
- (b) Draw the constellation diagrams for both conventional 8 PSK and for the QAM constellation defined above.
- (c) For each constellation, find some Gray mapping to assign bits to the signal alternatives.
- (d) You want to scale the amplitude of 8 PSK with some factor C to achieve equal average energy per bit \mathcal{E}_b for both constellations. Determine the scaling factor C. Which constellation will then have a lower symbol error probability with an ML receiver (assuming a large signal-to-noise ratio)?

A communication system can serve two users simultaneously, at carrier frequencies $f_1 = 300 \text{ MHz}$ and $f_2 = 600 \text{ MHz}$, respectively. The frequency spectrum X(f) of the combined bandpass signal $x(t) = x_1(t) + x_2(t)$ is given in the figure below:



- (a) Describe how the receiver can reconstruct the (real-valued) baseband signal $s_2(t)$ of the second user, using a multiplier and a low-pass filter.
- (b) Draw the spectrum of the converted signal *before* the filtering.

Consider now a 2-ray multi-path channel with impulse response

$$h(t) = \sum_{i=1}^{2} \alpha_i \,\delta(t - \tau_i)$$
, where $\alpha_1 = 1, \alpha_2 = -0.5, \ \tau_1 = 0 \,\mu s, \tau_2 = 1.5 \,\mu s$.

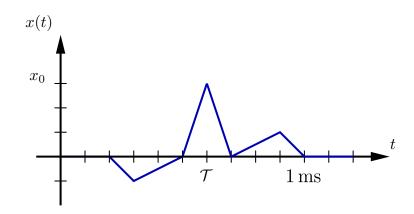
A binary PAM signal with rectangular pulse $g_{rec}(t)$ of duration $T = T_s = 3 \,\mu s$ and amplitude *A* is transmitted over this channel, namely

$$s(t) = \sum_{n=-\infty}^{\infty} A[n] g(t - nT_s)$$
 where $A[-1] = 0, A[1] = +1, A[2] = -1, A[3] = +1$.

- (c) Draw the signal z(t) at the output of the multi-path channel within the time interval $0 \le t \le 3T_s$.
- (d) What happens in this scenario, if an ML receiver is used that was designed for a channel without multipath? Which modifications to the transmitter/receiver can be done to guarantee the same error probability as for the case $h(t) = \delta(t)$?

Remark: Part (c) and (d) can be solved independently from (a) or (b).

Assume a communication system employing 4 PAM modulation with equally likely signal alternatives. The combination of the transmit pulse g(t), channel filter h(t), and receiver filter v(t) can be written as x(t) = g(t) * h(t) * v(t). The signal is sampled in the receiver at time instants $\mathcal{T} + i T_s$, i = 0, 1, 2, ...



- (a) What is the maximum possible bitrate for ISI-free reception?
- (b) Let us now assume that we can tolerate some ISI. You are allowed to choose either $T_s = 200 \,\mu\text{s}$ or $T_s = 300 \,\mu\text{s}$. Draw the discrete impulse response x[i] of the system for both cases.
- (C) Determine which of the two cases is better, by comparing their worst case intersymbol interference ISI_{wc} . What bitrate can be achieved with your selection?
- (d) Give an example of an information sequence A[i] for which the worst case ISI occurs. Is there a risk for erroneous decisions?

Your task is to specify a 8-PSK communication link that will be used to replace an existing 8-FSK system with a target symbol error rate of $P_s \leq 2 \cdot 10^{-6}$. The power at the receiver is equal to $P_z = 10^{-13} W$. The parameter of the additive white Gaussian noise is equal to $N_0 = 5 \cdot 10^{-20} [W/Hz]$ and the propagation attenuation is $\alpha = 0.02 \cdot d^{-1}$.

- (a) Determine the maximum bit rate R_b that can be achieved by the 8-FSK system.
- (b) The current distance between transmitter and receiver is d = 100 m. Your aim is to achieve the same P_s and R_b with 8-PSK modulation without increasing the transmit power. What is the maximum value of d under these conditions?
- (c) Compare the required bandwidth of the two solutions, assuming in both cases a rectangular pulse of duration $T = T_s$. The frequency separation of the FSK system is $f_{\Delta} = R_s$.
- (d) Describe some advantages and disadvantages of FSK modulation and of PSK modulation.