Exercise Lesson 11

Problems from the compendium:

none this time

Other problems:

11.1 We want to transmit the binary sequence

$$\mathbf{b} = 1 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0 \ 1 \ 1 \ 1$$

with bandpass 4-ary PAM signaling, using a rectangular pulse $g_{rec}(t)$ with amplitude 1 and duration $T = T_s$. Consider the following mapping from bits to the amplitudes $A_0 = -3$, $A_1 = -1$, $A_2 = +1$, and $A_3 = +3$:

$$00 \rightarrow A_0$$
, $10 \rightarrow A_1$, $11 \rightarrow A_2$, $01 \rightarrow A_3$.

- (a) Draw the baseband transmit signal $x_I(t)$.
- (b) Assume a homodyne receiver with phase error $\phi_{err}(t) = -67.5^{\circ}(\text{i.e.}, -3/8\pi)$ and A = 2 for an ideal noise-free channel, i.e., $y(t) = x_I(t) \cos(2\pi f_c t)$. Draw the inphase and quadrature component signals $u_I(t)$ and $u_Q(t)$ in the interval $0 \le t \le 2T_s$.
- (c) Let us now see the benefit of coherent reception. In the interval $0 \le t \le 2T_s$, draw the three signals $u_I(t)\cos(\phi_{err}(t)), u_Q(t)\sin(\phi_{err}(t))$, and $\hat{u}_I(t) = u_I(t)\cos(\phi_{err}(t)) u_Q(t)\sin(\phi_{err}(t))$.
- **11.2** Consider the same scenario as in Problem 11.1, but with 4-ary QAM signaling using amplitudes $A_0 = -1$, $A_1 = +1$, $B_0 = -1$, and $B_1 = +1$ with the mapping:

$$00 \to A_0, B_0$$
, $10 \to A_1, B_0$, $11 \to A_1, B_1$, $01 \to A_0, B_1$.

- (a) Draw the baseband transmit signals $x_I(t)$ and $x_Q(t)$.
- (b) The homodyne receiver with phase error $\phi_{err}(t) = -67.5^{\circ}(\text{i.e.}, -3/8\pi)$ and A = 2 sees now the input signal $y(t) = x_I(t) \cos(2\pi f_c t) x_Q(t) \sin(2\pi f_c t)$. Draw the inphase and quadrature component signals $u_I(t)$ and $u_Q(t)$ in the interval $0 \le t \le 2T_s$.
- (c) Is it possible to see the original data in $u_I(t)$ and $u_Q(t)$?
- **11.3** Assume again $y(t) = x_I(t) \cos(2\pi f_c t) x_Q(t) \sin(2\pi f_c t)$ at the input of a homodyne receiver.
 - (a) Use equations (3.171) and (3.172) to compute the signal

$$\hat{u}_I(t) = u_I(t)\cos(\phi_{err}(t)) - u_Q(t)\sin(\phi_{err}(t)) .$$

What can we conclude?

(b) (Optional) Can we recover the signal $u_Q(t)$ in a similar way?

11.4 For an ideal channel without noise the baseband signal produced by a homodyne receiver can be written as

$$\tilde{u}(t) = \frac{\tilde{x}(t)}{2} A \cdot e^{-j\phi_{err}(t)} = u_I(t) + j u_Q(t) ,$$

where

$$\tilde{x}(t) = x_I(t) + j x_Q(t) \; .$$

Show that this expression is equivalent to equations (3.171) and (3.172).

11.5 Let us now describe the signals in Problem 11.2 using complex baseband notation

$$\tilde{x}(t) = x_I(t) + j \ x_Q(t) = \sum_{n=0}^{4} \tilde{A}_{m[n]} g(t - nT_s)$$

- (a) Determine the sequence $\tilde{A}_{m[n]}$.
- (b) Draw the values $\tilde{A}_{m[n]}$ as points in the complex plane and label each point with the corresponding input bits.
- (c) The sequence is transmitted over a channel with $\tilde{h}(t) = 2 \tilde{\alpha} \delta(t)$ and additive noise $\tilde{w}(t)$, where $\tilde{\alpha} = 0.001 e^{-j\phi_{\alpha}}$. At the receiver the signal $\tilde{u}(t)$ is matchfiltered, sampled and multiplied with a scaling coefficient \tilde{C} , resulting in the decision variable

$$\tilde{\xi}_{m[n]} = \tilde{C} \left(\tilde{A}_{m[n]} \, \tilde{\alpha} \, e^{-j\phi_{err,n}} \frac{A \, E_g}{2} + \tilde{\mathcal{N}}_n \right) \;,$$

where $\tilde{\mathcal{N}}_n$ denotes the contribution of noise at the output of the matched filter. The coefficient \tilde{C} is chosen such that $\tilde{\xi}_{m[n]} = \tilde{A}_{m[n]}$ for $\tilde{\mathcal{N}}_n = 0$. Determine \tilde{C} .

(d) Assume now that another sequence **b** is transmitted under the same conditions. The decision variables computed by the receiver are

$$\xi_{m[n]} = (0.5 + 0.5 j), (-1.5 + 0.5 j), (1.5 + 1.5 j), (-0.5 + 0.5 j), (-1.5 - 0.5 j)$$

Draw the values $\xi_{m[n]}$ as points in the complex plane, together with the possible ideal values $\tilde{A}_{m[n]}$. For a minimum Euclidean distance receiver, determine the estimated sequences $\tilde{A}_{m[n]}$ and $\hat{\mathbf{b}}$.