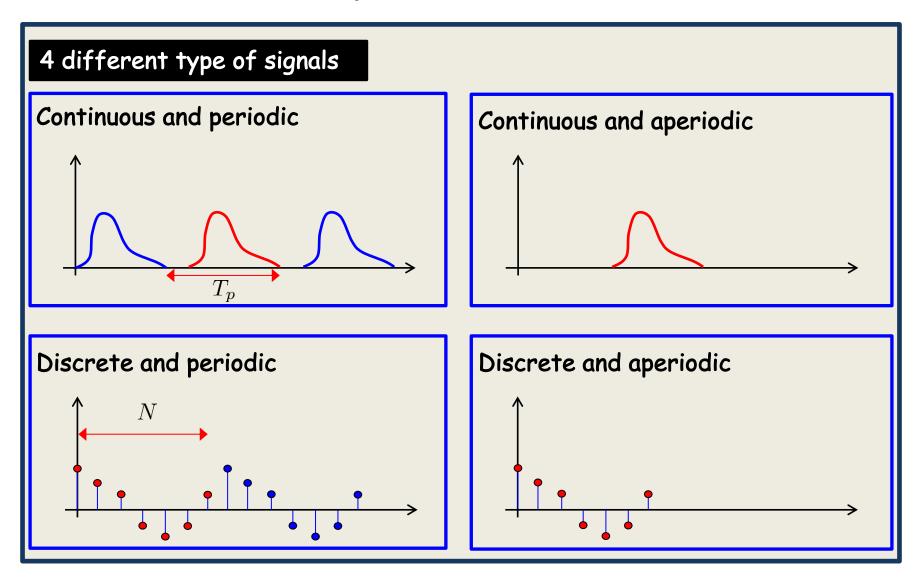
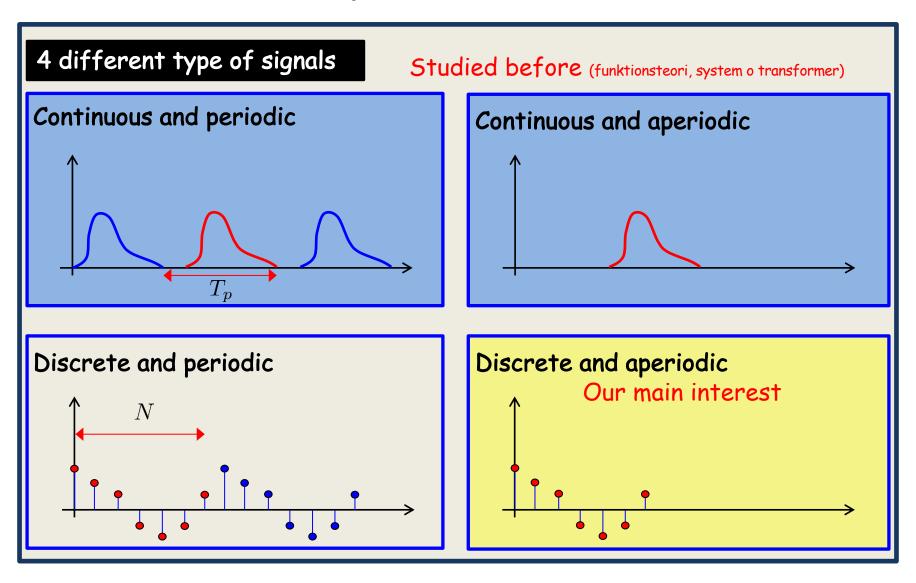
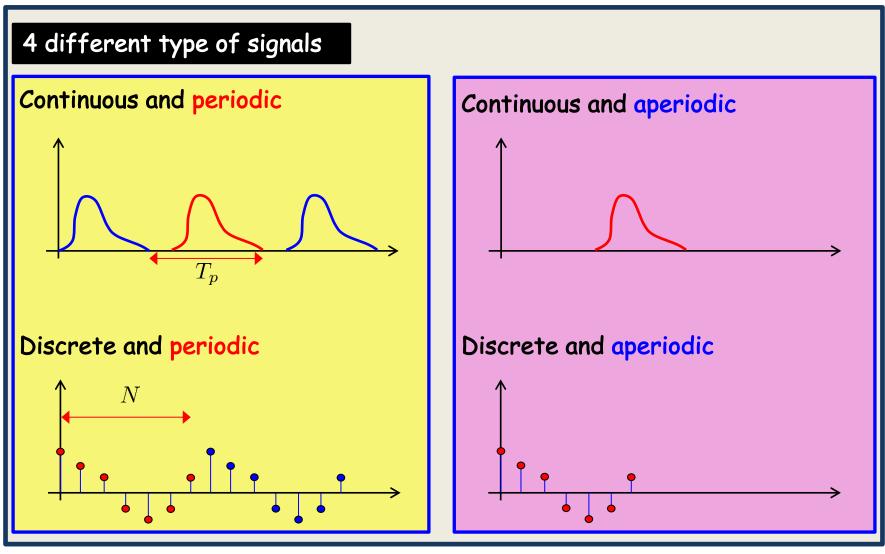
# **EITF75 Systems and Signals**

Lecture 5
The discrete-time Fourier transform

Fredrik Rusek

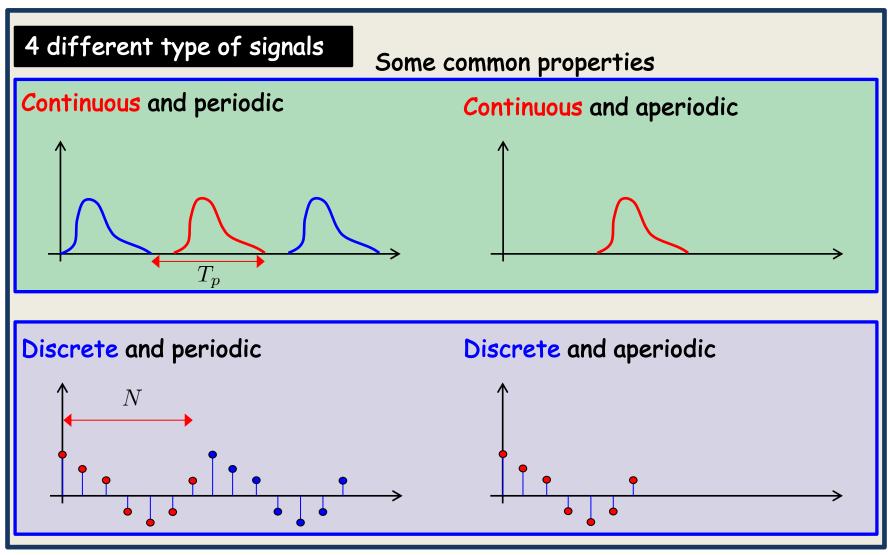




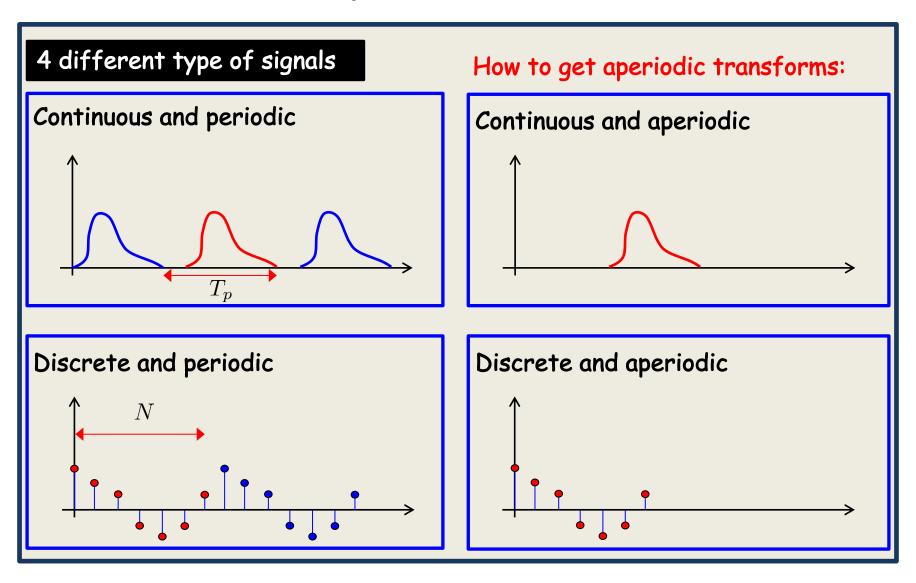


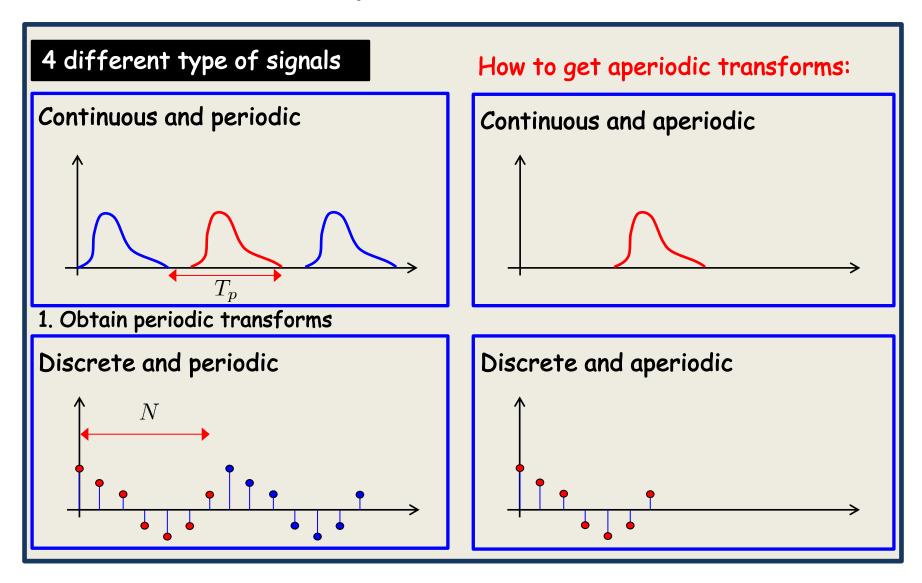
Some common properties

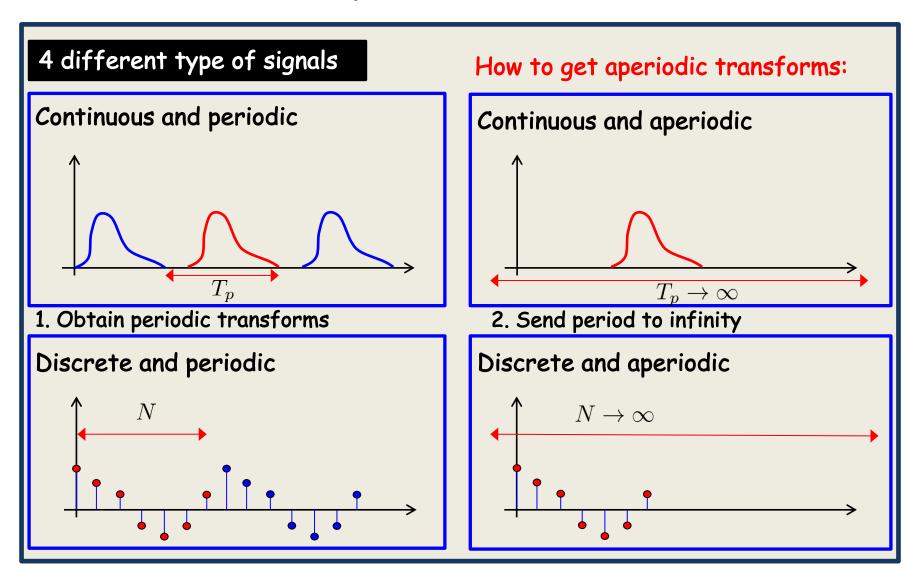
Some common properties



Some common properties





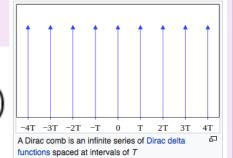




#### Dirac comb

$$ext{III}_T(t) \; riangleq \; \sum_{k=-\infty}^\infty \delta(t-kT)$$

$$=rac{1}{T}\sum_{n=-\infty}^{\infty}e^{i2\pi nrac{t}{T}}$$



$$\int_0^T x(\tau) \frac{1}{T} \sum_{k=-\infty}^{\infty} e^{i2\pi k \frac{t-\tau}{T}} d\tau = \int_0^T x(\tau) \sum_{k=-\infty}^{\infty} \delta(t-\tau-kT) d\tau$$

$$T_{p} \sum_{k=-\infty}^{\infty} |c_{k}|^{2} = T_{p} \sum_{k=-\infty}^{\infty} \left| \frac{1}{T_{p}} \int_{0}^{T_{p}} x(t) \exp(-i2\pi k F_{0} t) dt \right|^{2}$$

$$= \frac{1}{T_{p}} \sum_{k=-\infty}^{\infty} \int_{0}^{T_{p}} \int_{0}^{T_{p}} x(t) \exp(-i2\pi k F_{0} t) x^{*}(\tau) \exp(i2\pi k F_{0} \tau) dt d\tau$$

$$= \frac{1}{T_{p}} \sum_{k=-\infty}^{\infty} \int_{0}^{T_{p}} \int_{0}^{T_{p}} x(t) x^{*}(\tau) \exp(i2\pi k F_{0} (\tau - t)) dt d\tau$$

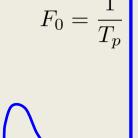
$$= \int_{0}^{T_{p}} \int_{0}^{T_{p}} x(t) x^{*}(\tau) \coprod_{T_{p}} (\tau - t) dt d\tau = \int_{0}^{T_{p}} |x(t)|^{2} dt$$

Parseval's formula

# Summary

 $\uparrow x(t)$ 

## Continuous and periodic



#### Fourier series representation

#### Analysis equation

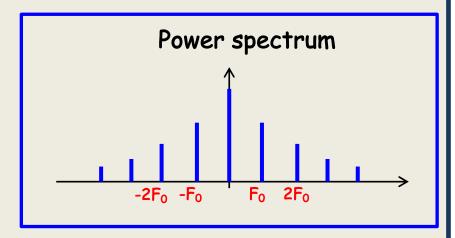
$$c_k = \frac{1}{T_p} \int_0^{T_p} x(t) \exp(-i2\pi k F_0 t) dt$$

#### Synthesis equation

$$x(t) = \sum_{k=-\infty}^{\infty} c_k \exp(i2\pi k F_0 t)$$

#### Parseval's identity

$$T_p \sum_{k=-\infty}^{\infty} |c_k|^2 = \int_0^{T_p} |x(t)|^2 dt$$



#### 4 different type of signals

#### Fourier transform

$$X(F) = \int_{-\infty}^{\infty} x(t) \exp(-i2\pi F t) dt$$

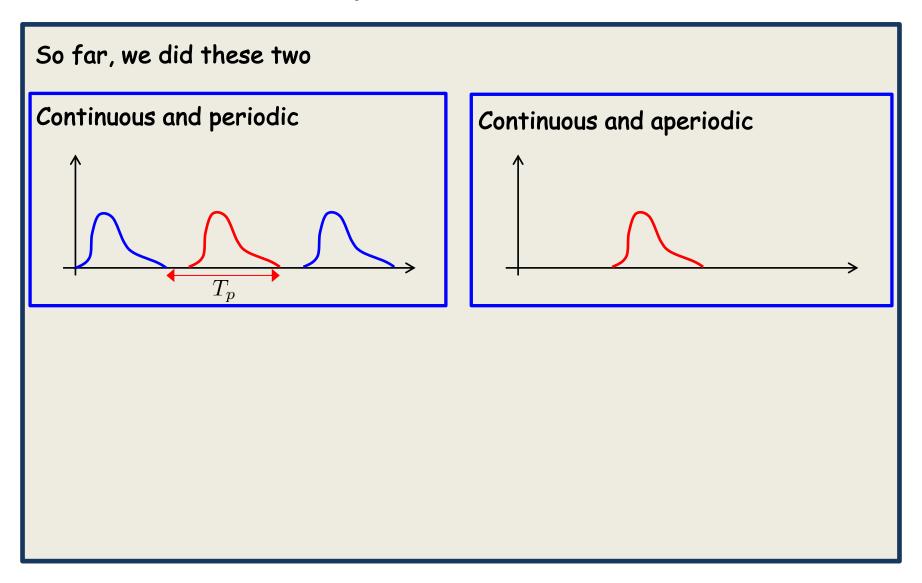
#### Parseval's identity

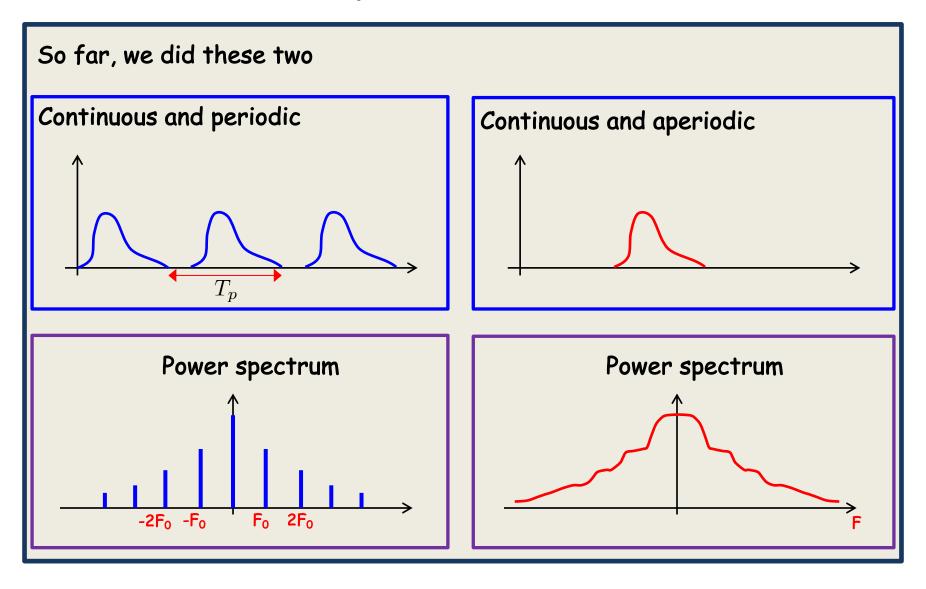
$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |x(F)|^2 dF$$

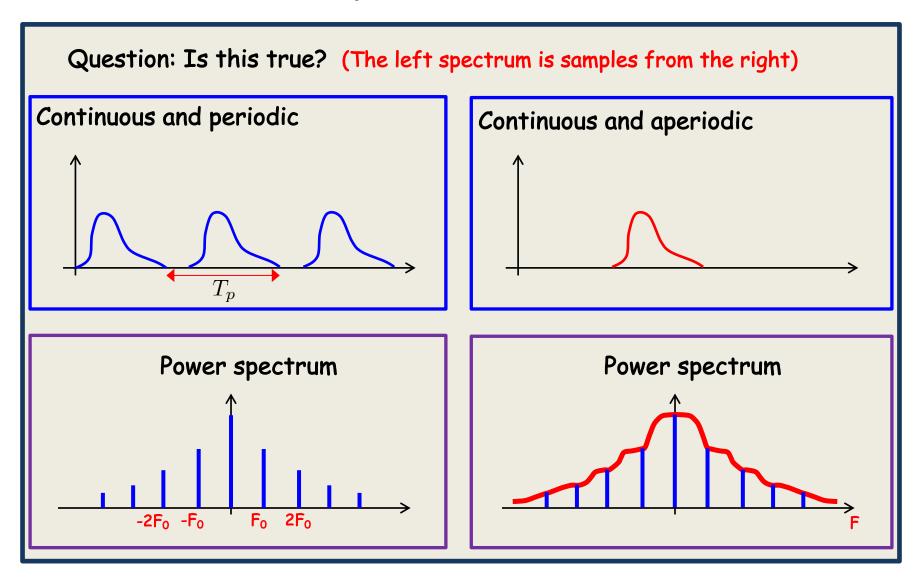
Energy in time, must be present in frequency as well

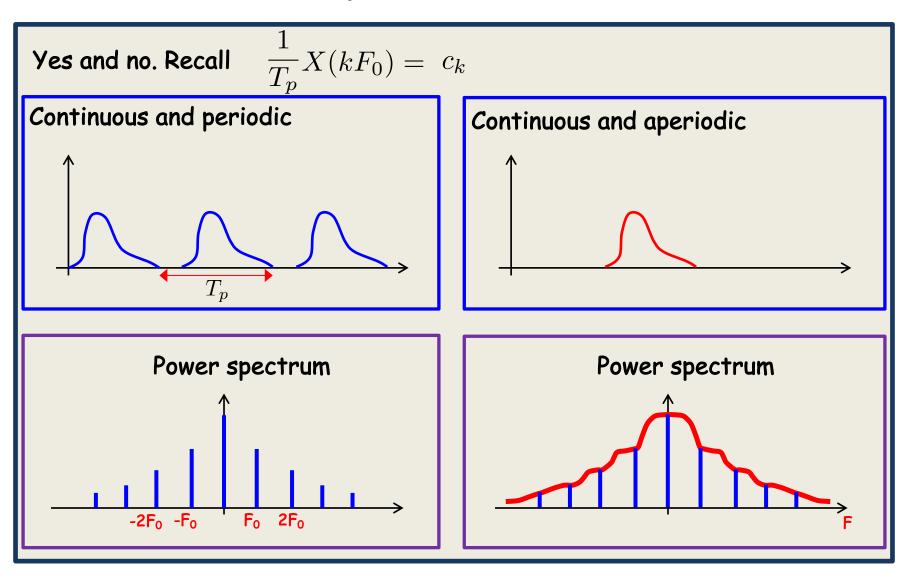
#### Inverse Fourier transform

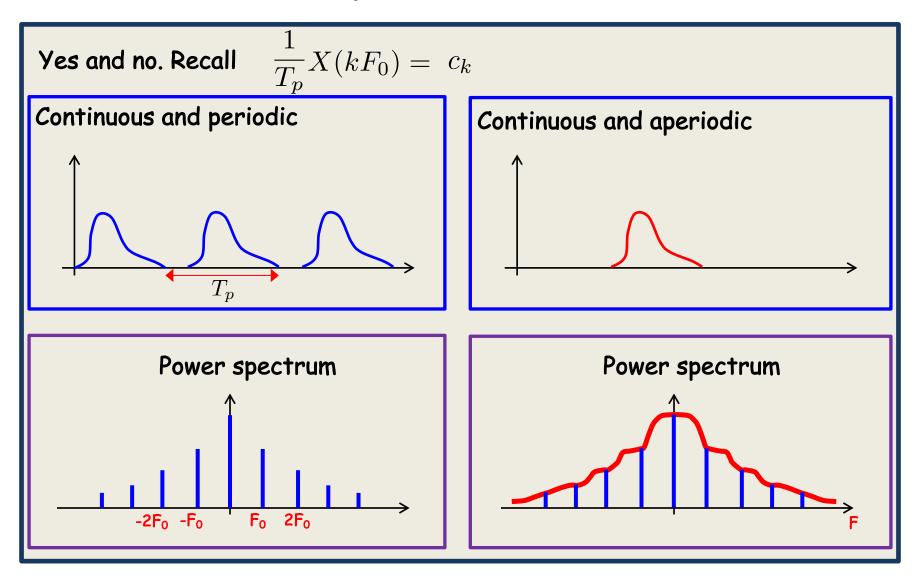
$$x(t) = \int_{-\infty}^{\infty} X(F) \exp(i2\pi F t) dF$$



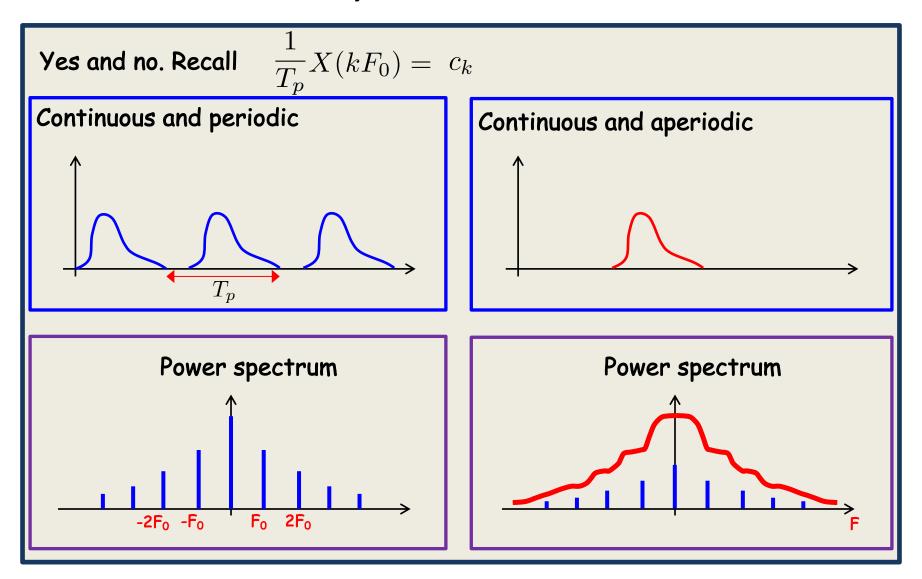


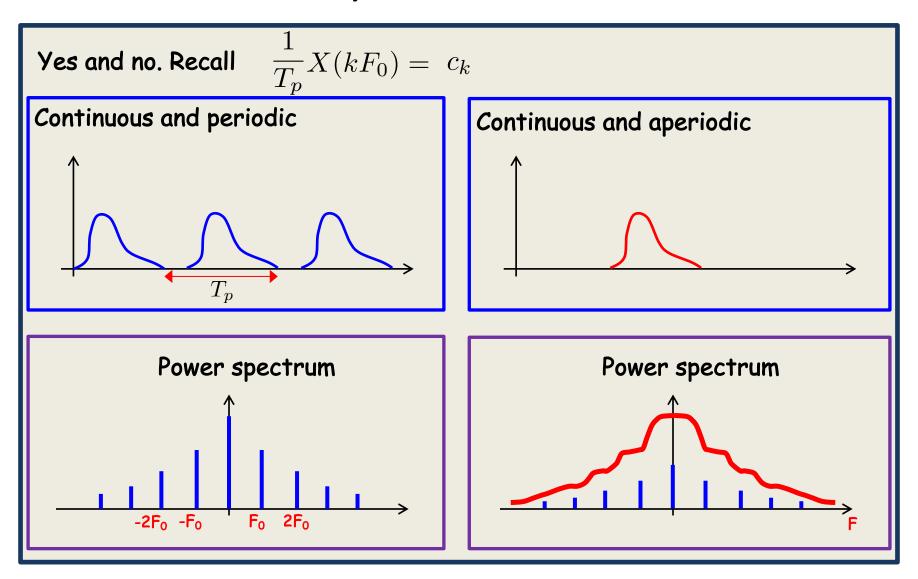


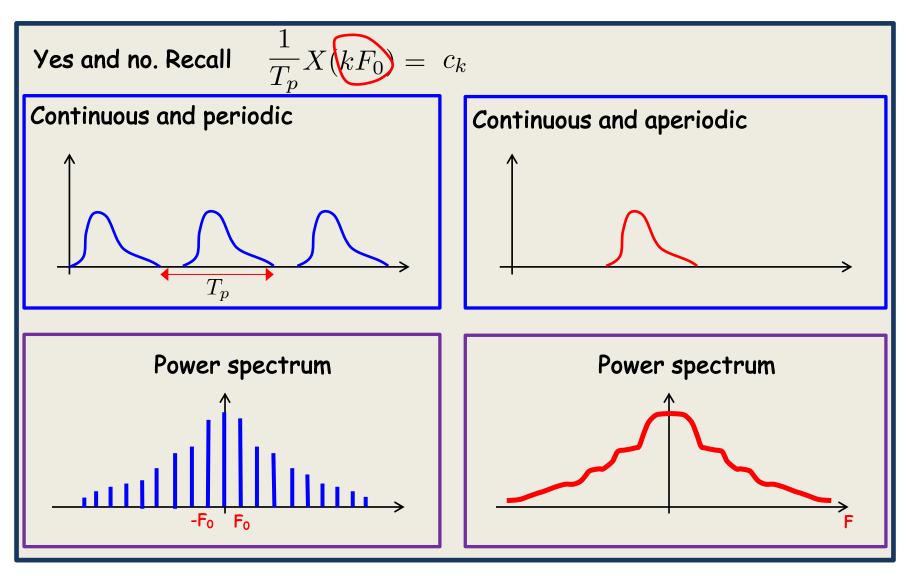




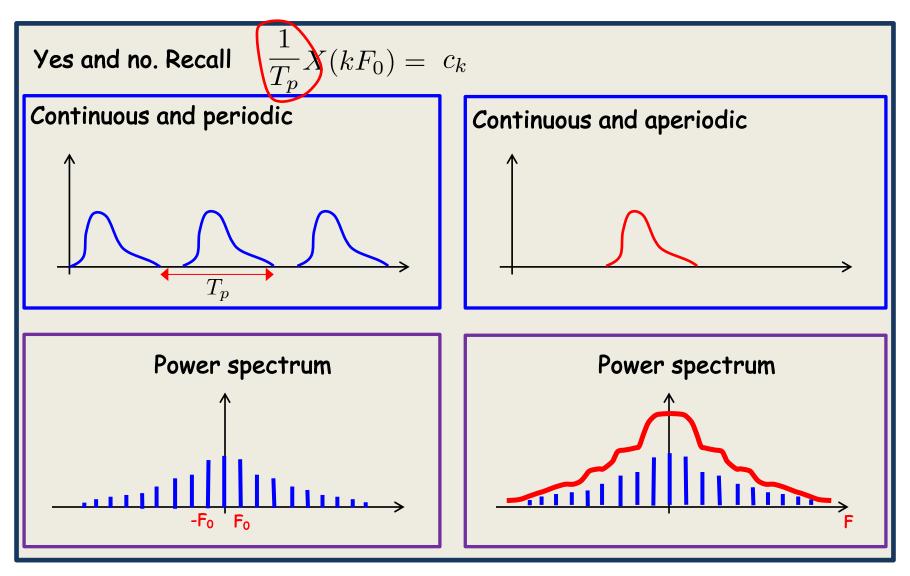
True if  $T_p=1$ 



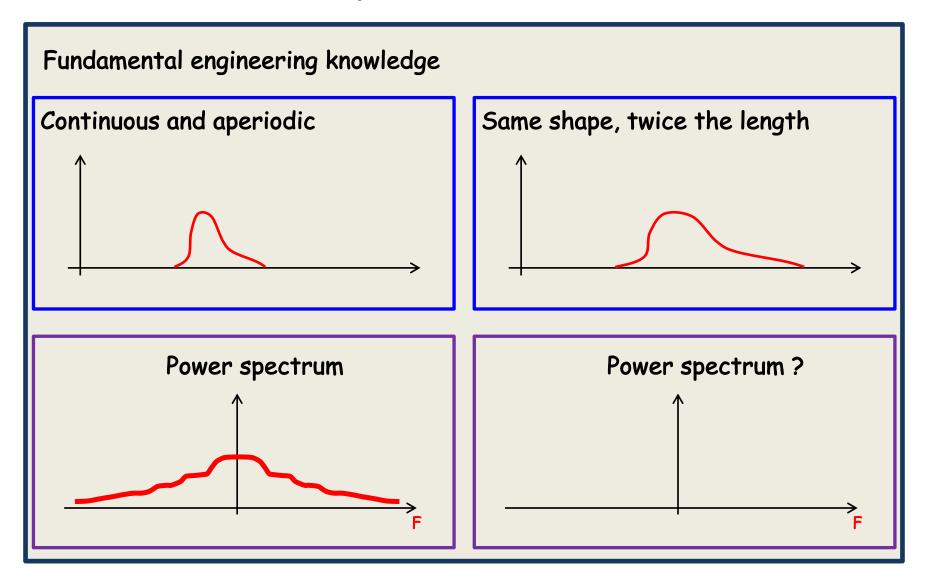


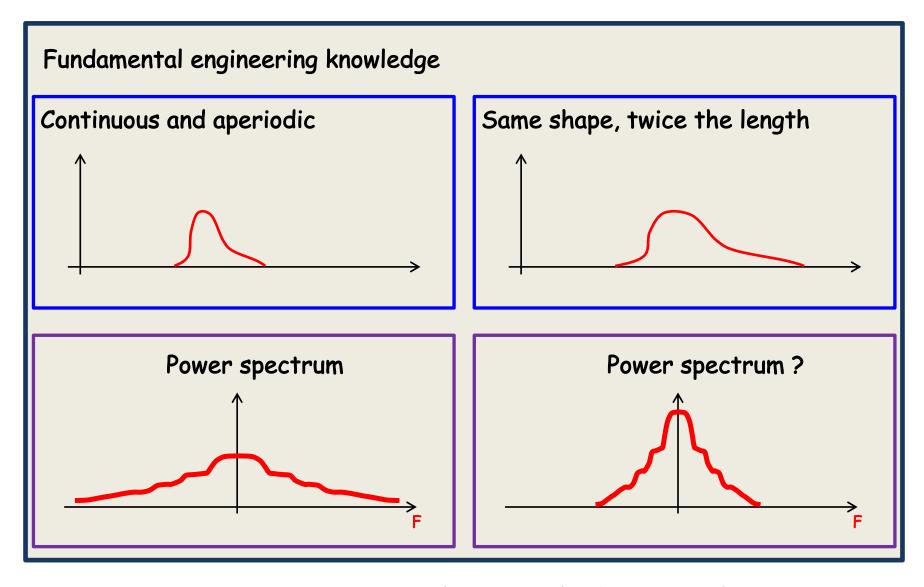


Effect 1: Denser sampling

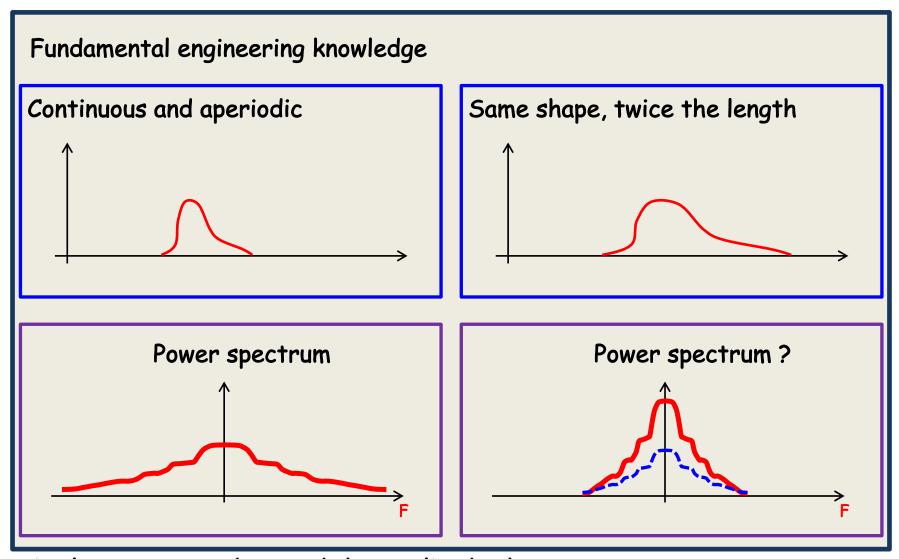


Effect 2: Scaled amplitude

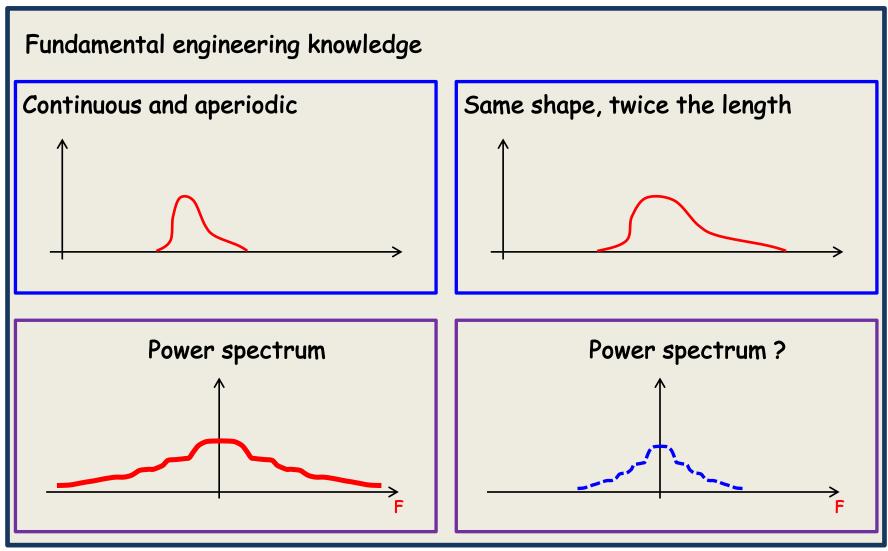




Strect in time is compression in frequency (and vica versa)

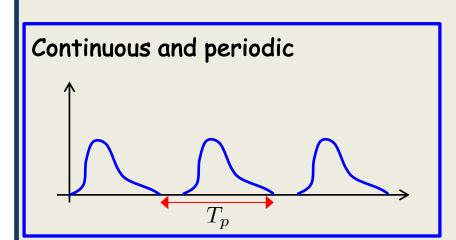


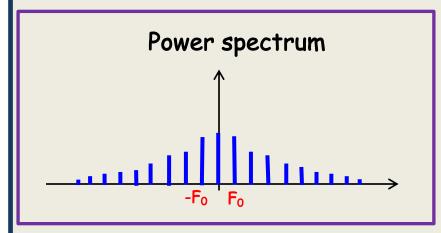
Simplest way to understand the amplitude change?



Simplest way to understand the amplitude change?

Parseval's identity: No way the integral-of-the-right-plot-squared equals the integral-of-the-left-plot-squared





Why is the spectrum discrete?

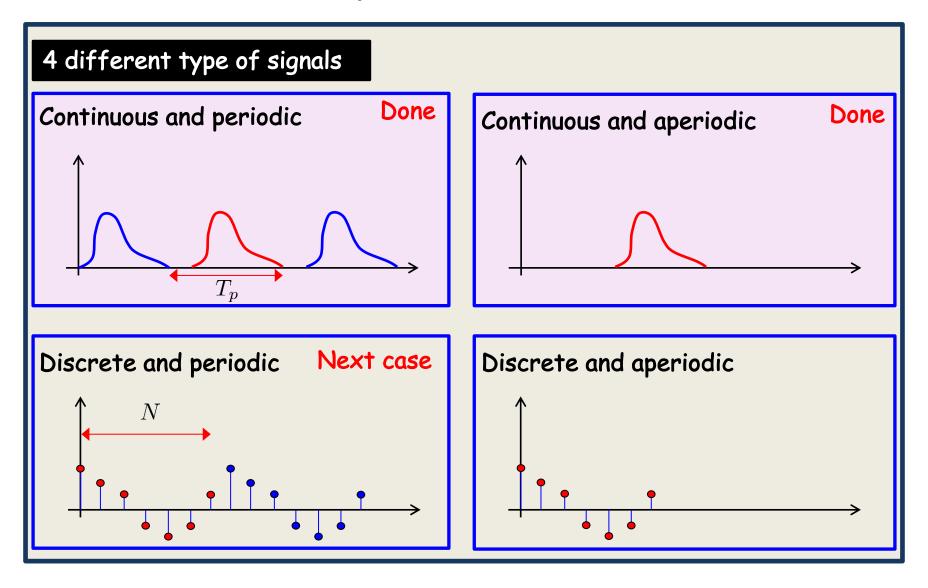
We can write the signal as

$$x(t) = \sum_{k=-\infty}^{\infty} c_k \exp(i2\pi k F_0 t)$$

However,  $\exp(i2\pi kF_0t)$ 

is not periodic with period Tp unless k is an integer.

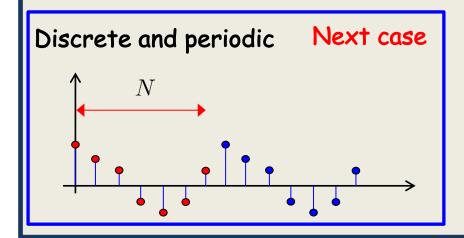
Thus, there can be no non-integer components in the spectrum



#### Recall what our goal is right now:

Given a periodic x(n), find either (we don't know which one yet)

- A set of coefficients c<sub>k</sub> representing x(n)
- A continuous function X(F)



#### Recall what our goal is right now:

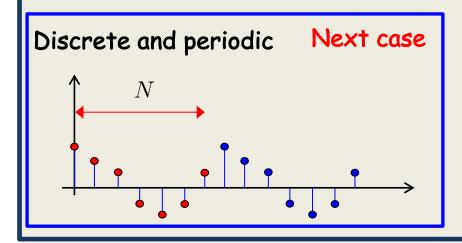
Given a periodic x(n), find either (we don't know which one yet)

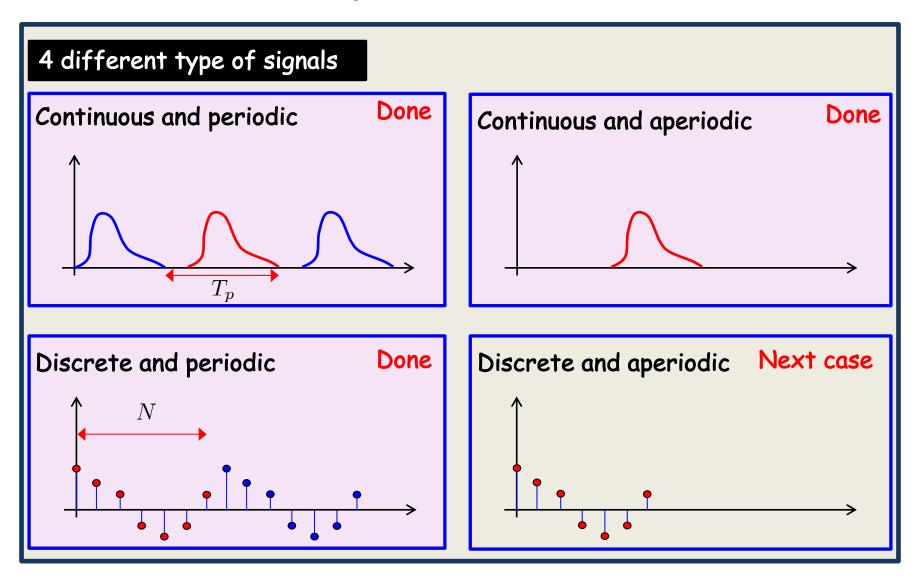
• A set of N coefficients  $c_k$  representing x(n)

Likely this should be the result

Don't forget (Lecture 1): For discrete signals, there is no difference between normalized frequency f=0.4 and f=...-0.6,1.4, 2.4,...

Therefore: Likely that the coefficients  $c_k$  are periodically extended.





#### Consider

$$X(f) = \sum_{n=0}^{N-1} x(n) \exp(-i2\pi nf)$$

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x(n) \exp(-i2\pi nk/N)$$

$$\hat{X}(f) = \sum_{n=0}^{N-1} c_n \exp(i2\pi nf)$$

What can we say about this one?

$$x(n) = \frac{1}{n+1}u(n)$$

Uniform convergence

$$\hat{X}(f) = X(f)$$

if, absolutely summable

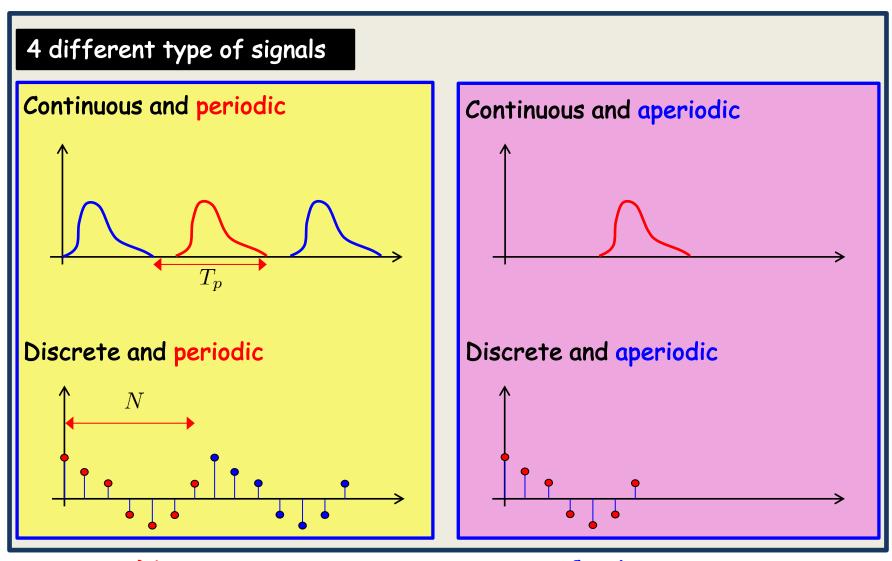
$$\sum |x(n)| < \infty$$

$$ho_{0.5}$$
 Mean square sense convergence

$$\int_{-0.5}^{0.5} |\hat{X}(f) - X(f)|^2 \mathrm{d}f = 0$$

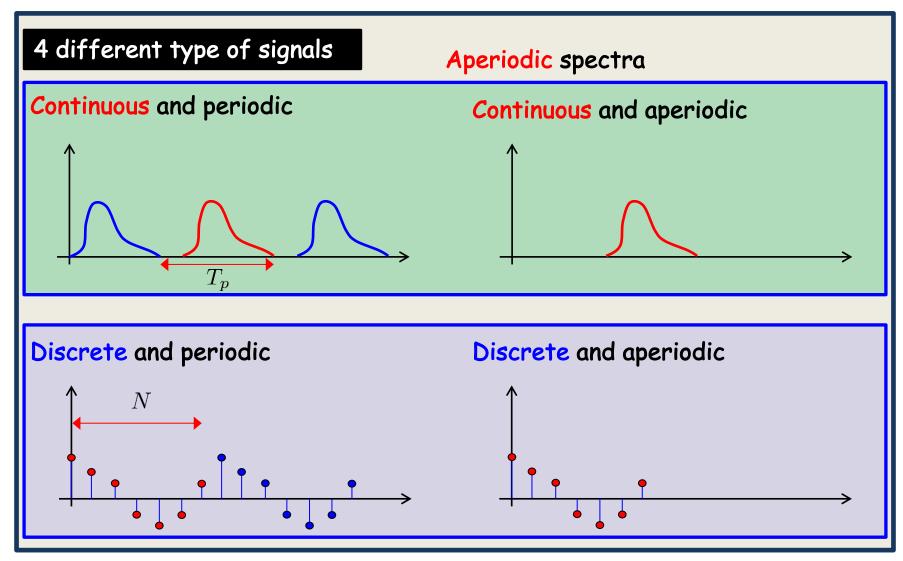
if, square summable

$$\sum |x(n)|^2 < \infty$$



Discrete spectra

Continuous spectra



#### Example

Find the Fourier transform of a square pulse

$$x(t) = \begin{cases} 1 & 0 \le t < T \\ 0 & \text{otherwise} \end{cases}$$

 $x(t) = \begin{cases} 1 & 0 \le t < T \\ 0 & \text{otherwise} \end{cases}$  Signal is continuous and aperiodic Transform is continuous and aperiodic

#### Example

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$$X(F) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi Ft} dt = \int_{0}^{T} 1 \cdot e^{-j2\pi Ft} dt$$

By definition

#### Example

#### Find the Fourier transform of a square pulse

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$$= \frac{e^{-j2\pi FT} - 1}{-j2\pi F}$$

Elementary integral

### Example

Find the Fourier transform of a square pulse

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$$= \frac{e^{-j2\pi FT} - 1}{-j2\pi F} = \frac{e^{-j2\pi \cdot \frac{T}{2} \cdot F} \left(e^{j2\pi \cdot \frac{T}{2} \cdot F} - e^{-j2\pi \cdot \frac{T}{2} \cdot F}\right)}{j2\pi F}$$

Minor manipulation

### Example

#### Find the Fourier transform of a square pulse

$$x(t) = \begin{cases} 1 & 0 \le t < T \\ 0 & \text{otherwise} \end{cases}$$
 Signal is continuous and aperiodic Transform is continuous and aperiodic

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$$= T \cdot \frac{\sin\left(2\pi \cdot \frac{T}{2} \cdot F\right)}{2\pi \cdot \frac{T}{2} \cdot F} \cdot e^{-j2\pi \cdot \frac{T}{2} \cdot F}$$

Euler's formula

#### Example

#### Find the Fourier transform of a square pulse

$$x(t) = \begin{cases} 1 & 0 \le t < T \\ 0 & \text{otherwise} \end{cases}$$

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$$= T \cdot \operatorname{sinc}\left(2 \cdot \frac{T}{2} \cdot F\right) \cdot e^{-j2\pi \cdot \frac{T}{2} \cdot F}$$

$$\operatorname{sinc}(x) = \frac{\sin(\pi x)}{\pi x}$$

$$\lim_{x \to 0} \frac{\sin(\pi x)}{\pi x} = 1$$

$$\lim_{x \to 0} \frac{\sin(\pi x)}{\pi x} = 1$$

#### Definition of a sinc-pulse

In 1948, Shannon used this pulse to derive the ultimate limit, in bits/sec, of communication. Super important pulse in EE

Emre Telatar: (In brief: Superstar)

"What Shannon's 48 paper has done for communication engineering has no parallel in any engineering field"

### Example

Find the Fourier transform of a square pulse

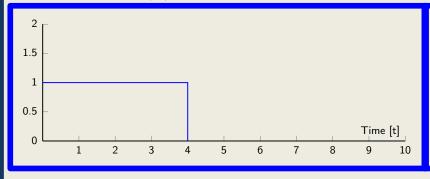
$$x(t) = \begin{cases} 1 & 0 \le t < T \\ 0 & \text{otherwise} \end{cases}$$

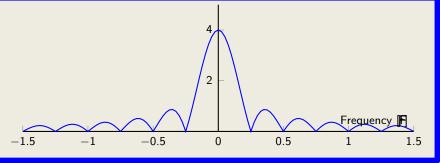
 $x(t) = \begin{cases} 1 & 0 \le t < T \\ 0 & \text{otherwise} \end{cases}$  Signal is continuous and aperiodic Transform is continuous and aperiodic

$$X(F) = T \cdot \operatorname{sinc}\left(2 \cdot \frac{T}{2} \cdot F\right) \cdot e^{-j2\pi \cdot \frac{T}{2} \cdot F} \qquad \operatorname{sinc}(x) = \frac{\sin(\pi x)}{\pi x}$$

$$x(t)$$
  $T=4$ 

|X(F)|





### Example

Find the Fourier transform of a discrete square pulse

$$x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$$

 $x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$  Signal is discrete and aperiodic Transform is continuous and periodic

### Example

Find the Fourier transform of a discrete square pulse

$$x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$$

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Note that the periodicity rules out the sincshape from the continuous case

### Example

Find the Fourier transform of a discrete square pulse

$$x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$$

 $x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$  Signal is discrete and aperiodic Transform is continuous and periodic

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n} = \sum_{n=0}^{N-1} 1 \cdot e^{-j\omega n}$$

By definition

#### Example

Find the Fourier transform of a discrete square pulse

$$x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$$

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$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n} = \sum_{n=0}^{N-1} 1 \cdot e^{-j\omega n}$$

$$= \frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}}$$
 Geometric series

### Example

Find the Fourier transform of a discrete square pulse

$$x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$$

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$$X(\omega) = \sum_{n = -\infty}^{\infty} x(n) e^{-j\omega n} = \sum_{n = 0}^{N-1} 1 \cdot e^{-j\omega n}$$
$$= \frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}}$$

$$=\frac{e^{j\omega N/2}\left(e^{j\omega\cdot\frac{N}{2}}-e^{-j\omega\cdot\frac{N}{2}}\right)}{e^{j\omega/2}\left(e^{j\omega\cdot\frac{1}{2}}-e^{-j\omega\cdot\frac{1}{2}}\right)}$$

Manipulation to reach Euler's

### Example

Find the Fourier transform of a discrete square pulse

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$$= \frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}}$$

$$= \frac{e^{j\omega N/2} \left( e^{j\omega \cdot \frac{N}{2}} - e^{-j\omega \cdot \frac{N}{2}} \right)}{e^{j\omega/2} \left( e^{j\omega \cdot \frac{1}{2}} - e^{-j\omega \cdot \frac{1}{2}} \right)}$$

$$= N \cdot \frac{\sin\left(\omega \cdot \frac{N}{2}\right)}{N \sin\left(\omega \cdot \frac{1}{2}\right)} \cdot e^{-j\omega(N-1)/2}$$

Period = 1

#### Example

Find the Fourier transform of a discrete square pulse

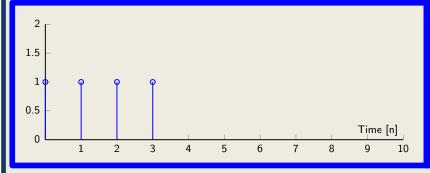
$$x(n) = \begin{cases} 1 & 0 \le n < N \\ 0 & \text{otherwise} \end{cases}$$

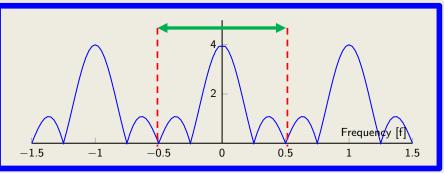
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$$x(n)$$
  $N = 4$ 





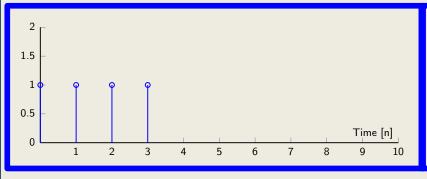


Period = 1

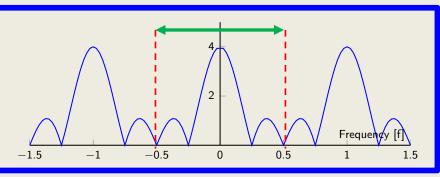
#### Remember?

$$\int_0^1 X(f) \exp(i2\pi nf) df = \int_{-0.5}^{0.5} X(f) \exp(i2\pi nf) df$$

$$x(n)$$
  $N = 4$ 



### |X(f)|



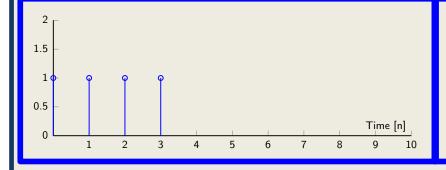
Period = 1

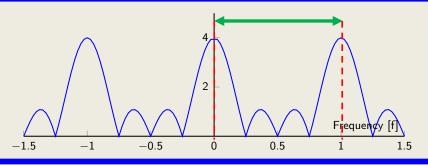
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$$x(n)$$
  $N=4$ 







Same content, different order