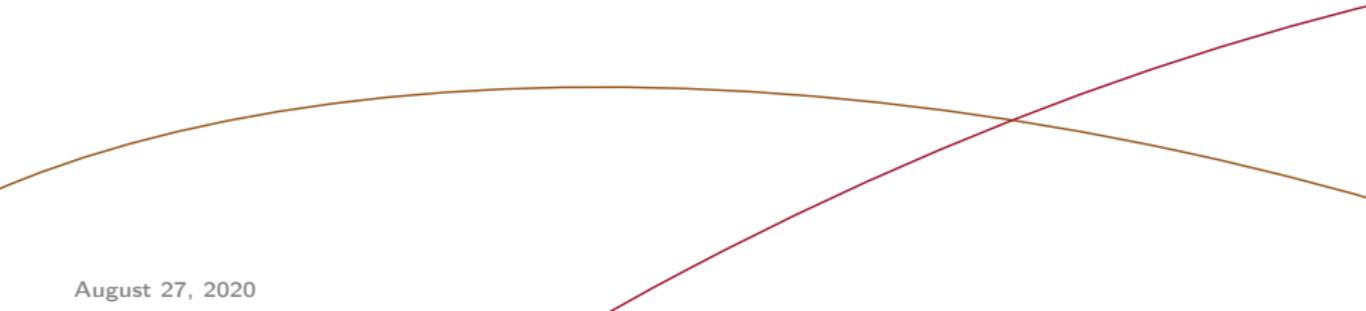




# Digitalteknik EITF65

Lecture 13: Linear Sequential Circuits, Part 1



# Linear Boolean functions

## Definition (Linearity)

A function  $f$  is said to be linear if

$$f(\mathbf{x} \oplus \mathbf{y}) = f(\mathbf{x}) \oplus f(\mathbf{y}) \quad (\text{L}_1)$$

$$f(\alpha \mathbf{x}) = \alpha f(\mathbf{x}) \quad (\text{L}_2)$$

## In other words

- ▶ The sum of the arguments gives the same as the sum of the functions.
- ▶ Multiplying the argument with a scalar is the same as multiplying the function.

# Linear Boolean functions

## Theorem

A Boolean function is linear if and only if it can be written as

$$\begin{aligned}f(\mathbf{x}) &= a_1x_1 \oplus \cdots \oplus a_nx_n \\&= (a_1 \dots a_n) \begin{pmatrix} x_1 \\ \vdots \\ x_n \end{pmatrix} = \mathbf{ax} \quad a_i \in \{0, 1\}\end{aligned}$$

- ▶ Any linear Boolean function can be realized with an  $n$ -input modulo 2 adder.

## Realisation of linear Boolean functions

- ▶ A combinational circuit containing only modulo 2 adders realizes a linear Boolean function.
- ▶ A linear Boolean function can be implemented with modulo 2 adders.

## Linearity test

With the **ring sum expansion (RSE)** all Boolean functions can (uniquely) be written with Boolean ring operations ( $\oplus, \otimes$ ) instead of Boolean algebra operations ( $\vee, \wedge, '$ ).

$$\begin{aligned}f(x) = & a_0 \oplus a_1 x_1 \oplus \cdots \oplus a_n x_n \\& \oplus a_{n+1} x_1 x_2 \oplus \cdots \oplus a_{2^n-1} x_1 \dots x_n\end{aligned}$$

In a **linear Boolean function**  $a_i = 0$ ,  $i = 0$  and  $i > n$ ,

$$f_a(x) = a_1 x_1 \oplus \cdots \oplus a_n x_n$$

## Theorem (4.6)

All Boolean functions  $f(x) \in B_n$  can be written with the (ring) operations  $\oplus$  and  $\otimes$ , by the *ring sum expansion (RSE/RMF)*,

$$f(x) = \bigoplus_{j=0}^{2^n-1} a_j \bigotimes_{i \in I_n(j)} x_i$$

where  $a_j \in B$  and  $I_n(j)$  is an index function.

The RSE/RMF expression is unique for the function.

Remark: Often the name Reed-Muller form is used.

## Derivation of RSE/RMF

Four ways to derive the RSE/RMF from a Boolean expression:

- ▶ Use the definition of the Boolean operations:

$$a \wedge b = a \cdot b$$

$$a \vee b = a \oplus b \oplus ab$$

$$a' = 1 \oplus a$$

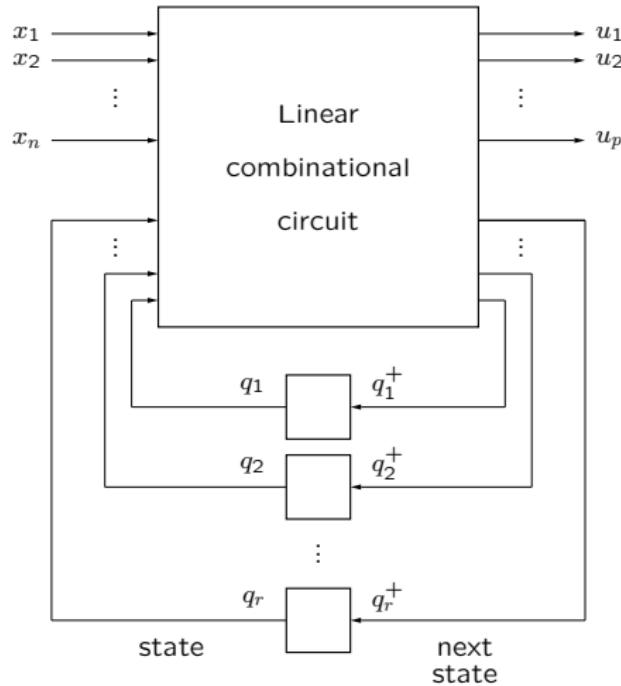
- ▶ Use deMorgan's law to get rid of  $\vee$ , then use  $a' = 1 \oplus a$ .
- ▶ Write the function in DNF. Then use that

$$m_i \vee m_j = m_i \oplus m_j \oplus \underbrace{m_i \cdot m_j}_{=0, i \neq j} = m_i \oplus m_j$$

and  $a' = 1 \oplus a$ .

- ▶ Reed-Muller transform

# Linear sequential circuit



# Linear sequential circuit

## Linear sequential circuit

A **linear sequential circuit** is a sequential circuit with only modulo 2 adders and delay elements.

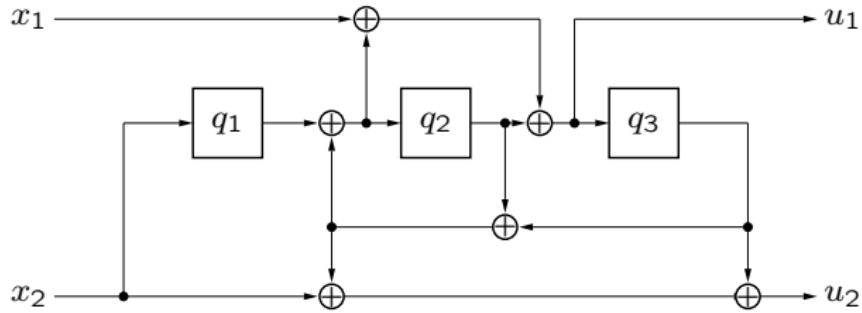
## Theorem

*A linear sequential circuit can be described by*

$$\begin{cases} \mathbf{q}^+ = A\mathbf{q} \oplus B\mathbf{x} \\ \mathbf{u} = C\mathbf{q} \oplus H\mathbf{x} \end{cases}$$

# A linear sequential circuit

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## An example

### Example

The equations can be expressed in matrix form as

$$\begin{pmatrix} q_1^+ \\ q_2^+ \\ q_3^+ \end{pmatrix} = \begin{pmatrix} x_2 \\ q_1 \oplus q_2 \oplus q_3 \\ x_1 \oplus q_1 \oplus q_3 \end{pmatrix} = \begin{pmatrix} 0 & 0 & 0 \\ 1 & 1 & 1 \\ 1 & 0 & 1 \end{pmatrix} \begin{pmatrix} q_1 \\ q_2 \\ q_3 \end{pmatrix} \oplus \begin{pmatrix} 0 & 1 \\ 0 & 0 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

$$\begin{pmatrix} u_1 \\ u_2 \end{pmatrix} = \begin{pmatrix} x_1 \oplus q_1 \oplus q_3 \\ x_2 \oplus q_2 \end{pmatrix} = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \end{pmatrix} \begin{pmatrix} q_1 \\ q_2 \\ q_3 \end{pmatrix} \oplus \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

Hence,

$$A = \begin{pmatrix} 0 & 0 & 0 \\ 1 & 1 & 1 \\ 1 & 0 & 1 \end{pmatrix} \quad B = \begin{pmatrix} 0 & 1 \\ 0 & 0 \\ 1 & 0 \end{pmatrix} \quad C = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \end{pmatrix} \quad H = \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

# Rank

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## Definition (Matrix Rank)

The **rank** of a matrix  $A$  is the maximum number of linearly independent rows (or columns) in  $A$ .

For an  $m \times n$  matrix the rank is bounded by  $\text{Rank}(A) \leq \min\{m, n\}$

## Inverse matrix

- ▶ if  $m = n$  then  $\text{Rank}(A) = m \Leftrightarrow A^{-1}$  exists,  
 $AA^{-1} = A^{-1}A = I$ .
- ▶ if  $m < n$  then  $\text{Rank}(A) = m \Leftrightarrow A_R^{-1}$  exists,  
 $AA_R^{-1} = I$ .
- ▶ if  $m > n$  then  $\text{Rank}(A) = n \Leftrightarrow A_L^{-1}$  exists,  
 $A_L^{-1}A = I$ .

## Derivation of rank

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The rank of a matrix  $A$  can be derived by Gauss elimination. After elimination the number of **pivot elements** is the rank.

Consider the matrix (over the binary Boolean ring)

$$A = \begin{pmatrix} 1 & 0 & 1 & 0 \\ 1 & 1 & 0 & 0 \\ 1 & 0 & 0 & 1 \\ 1 & 1 & 1 & 1 \end{pmatrix}$$

Gauss elimination

$$\rightarrow \begin{pmatrix} \textcircled{1} & 0 & 1 & 0 \\ 0 & 1 & 1 & 0 \\ 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 \end{pmatrix} \rightarrow \begin{pmatrix} \textcircled{1} & 0 & 1 & 0 \\ 0 & \textcircled{1} & 1 & 0 \\ 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1 \end{pmatrix} \rightarrow \begin{pmatrix} \textcircled{1} & 0 & 1 & 0 \\ 0 & \textcircled{1} & 1 & 0 \\ 0 & 0 & \textcircled{1} & 1 \\ 0 & 0 & 0 & 0 \end{pmatrix}$$

The rank is 3.

**Remark:** The same derivation over the real numbers will give rank 4.

# Diagnostic Matrix

## Diagnostic Matrix

The **diagnostic matrix** is the  $pr \times r$  matrix

$$K = \begin{pmatrix} C \\ CA \\ CA^2 \\ \vdots \\ CA^{r-1} \end{pmatrix}$$

with maximum rank,  $\text{Rank}(K) = r$ .

If  $\text{Rank}(K) < r$ , then there are equivalent states.

## Reduced form

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### Theorem (7.4)

*Let  $A$ ,  $B$ ,  $C$ , and  $H$  determine a linear sequential circuit where  $\text{Rank}(K) = k$ . Then, the reduced form is given by*

$$A_{\text{red}} = TAR$$

$$B_{\text{red}} = TB$$

$$C_{\text{red}} = CR$$

$$H_{\text{red}} = H$$

*where  $T$  consists of the first  $k$  linearly independent rows of the diagnostic matrix  $K$ , and  $R$  is a right inverse of  $T$ .*

## The example continued (I)

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The diagnostic matrix:

$$K = \begin{pmatrix} C \\ CA \\ CA^2 \end{pmatrix} = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \\ 1 & 0 & 1 \\ 1 & 1 & 1 \\ 1 & 0 & 1 \\ 0 & 1 & 0 \end{pmatrix}$$

The rank:  $\text{Rank}(K) = 2$ . Thus,

$$T = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \end{pmatrix} \quad R = T_R^{-1} = \begin{pmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{pmatrix}$$

where  $R$  is a right inverse of  $T$ .

## The example continued (II)

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How to find the right inverse of  $T$ ?

Since the first 2 columns of  $T$  are linearly independent we write  $T$  as

$$T = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 1 & 0 \end{pmatrix} = (M \ N)$$

where  $M$  is the first 2 (linearly independent) columns of  $T$ , and  $N$  is the rest. An inverse of  $M$  can be found in

$$M^{-1} = \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

Since

$$(M \ N) \begin{pmatrix} M^{-1} \\ 0 \end{pmatrix} = MM^{-1} = I$$

we have found a right inverse in

$$R = \begin{pmatrix} M^{-1} \\ 0 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{pmatrix}$$

## The example continued (III)

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The matrices are

$$A_{\text{red}} = TAR = \begin{pmatrix} 1 & 0 \\ 1 & 1 \end{pmatrix}$$

$$B_{\text{red}} = TB = \begin{pmatrix} 1 & 1 \\ 0 & 0 \end{pmatrix}$$

$$C_{\text{red}} = CR = \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

$$H_{\text{red}} = H = \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

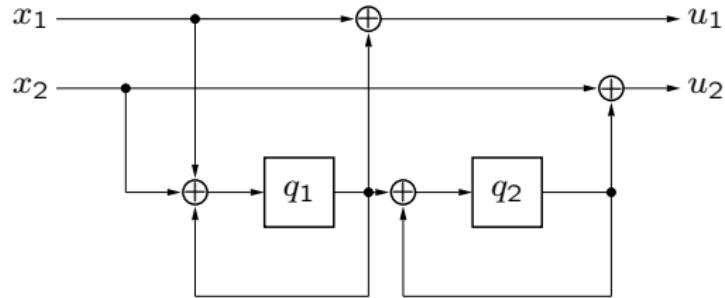
Hence,

$$\begin{pmatrix} q_1^+ \\ q_2^+ \end{pmatrix} = A_{\text{red}}\mathbf{q} \oplus B_{\text{red}}\mathbf{x} = \begin{pmatrix} q_1 \oplus x_1 \oplus x_2 \\ q_1 \oplus q_2 \end{pmatrix}$$

$$\begin{pmatrix} u_1 \\ u_2 \end{pmatrix} = C_{\text{red}}\mathbf{q} \oplus H_{\text{red}}\mathbf{x} = \begin{pmatrix} q_1 \oplus x_1 \\ q_2 \oplus x_2 \end{pmatrix}$$

## Reduced form

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## Definition

The sequence

$$x = \dots x_{-1} x_0 x_1 x_2 \dots$$

can be represented by

$$x(D) = \dots \oplus x_{-1} D^{-1} \oplus x_0 \oplus x_1 D \oplus x_2 D^2 \oplus \dots = \sum_i x_i D^i$$

where  $D$  is a delay operator.

## $\mathcal{D}$ -transform (Ex)

### Example

#### Examples of $\mathcal{D}$ -transform

- ▶  $1011010111 \dots \xrightarrow{\mathcal{D}} 1 \oplus D^2 \oplus D^3 \oplus D^5 \oplus D^7 \oplus D^8 \oplus D^9 \oplus \dots$   
 $\uparrow$   
 $t=0$
- ▶  $110\textcolor{red}{1}101 \dots \xrightarrow{\mathcal{D}} D^{-3} \oplus D^{-2} \oplus 1 \oplus D \oplus D^3 \oplus \dots$   
 $\uparrow$   
 $t=0$   
 $= D^{-3}(1 \oplus D \oplus D^3 \oplus D^4 \oplus D^6 \oplus \dots)$
- ▶  $D^2 \oplus D^6 \oplus D^7 \oplus D^8 \oplus \dots \xrightarrow{\mathcal{D}^{-1}} \textcolor{red}{0}01000111 \dots$   
 $\uparrow$   
 $t=0$   
 $D^2(1 \oplus D^4 \oplus D^5 \oplus D^6 \oplus \dots) \xrightarrow{\mathcal{D}^{-1}} \textcolor{red}{1}000111 \dots$   
 $\uparrow$   
 $t=2$

## $\mathcal{D}$ -transform of periodic sequences

### Theorem

*The (infinite) periodic sequence*

$$\mathbf{s} = [s_0 s_1 \dots s_{T-1}]^\infty = s_0 \dots s_{T-1} s_0 \dots s_{T-1} \dots$$

*has the  $\mathcal{D}$ -transform*

$$s(D) = \frac{P(D)}{1 \oplus D^T}$$

*where  $P(D)$  is the  $\mathcal{D}$ -transform of one period,*

$$P(D) = \mathcal{D}(s_0 s_1 \dots s_{T-1})$$

*and  $T$  the period.*

## $\mathcal{D}$ -transform of periodic sequences

Proof.

$\mathcal{D}$ -transform of  $[s_0 \dots s_{T-1}]^\infty$ :

- ▶ 1st position

$$[10 \dots 0]^\infty \xrightarrow{\mathcal{D}} 1 \oplus D^T \oplus D^{2T} \oplus \dots = \frac{1}{1 \oplus D^T}$$

- ▶  $i + 1$ st position

$$[0 \dots 1 \dots 0]^\infty \xrightarrow{\mathcal{D}} D^i \oplus D^{T+i} \oplus D^{2T+i} \oplus \dots = \frac{D^i}{1 \oplus D^T}$$

Super positioning gives

$$\begin{aligned}[s_0 s_1 \dots s_{T-1}]^\infty &\xrightarrow{\mathcal{D}} s_0 \frac{1}{1 \oplus D^T} \oplus \dots \oplus s_{T-1} \frac{D^{T-1}}{1 \oplus D^T} \\&= \frac{s_0 \oplus s_1 D \oplus \dots \oplus s_{T-1} D^{T-1}}{1 \oplus D^T} \\&= \frac{\mathcal{D}(s_0 s_1 \dots s_{T-1})}{1 \oplus D^T}\end{aligned}$$



## $\mathcal{D}$ -transform of periodic sequences

### Example

- ▶  $[01]^\infty \xrightarrow{\mathcal{D}} \frac{D}{1 \oplus D^2}$
- ▶  $[110]^\infty \xrightarrow{\mathcal{D}} \frac{1 \oplus D}{1 \oplus D^3} = \frac{1 \oplus D}{(1 \oplus D)(1 \oplus D \oplus D^2)} = \frac{1}{1 \oplus D \oplus D^2}$
- ▶  
$$\begin{aligned} \underset{t=0}{\overset{\uparrow}{11}}[110]^\infty &\xrightarrow{\mathcal{D}} 1 \oplus D \oplus D^2 \frac{1 \oplus D}{1 \oplus D^3} = \frac{(1 \oplus D)(1 \oplus D^3)}{1 \oplus D^3} \oplus D^2 \frac{1 \oplus D}{1 \oplus D^3} \\ &= \frac{1 \oplus D \oplus D^3 \oplus D^4 \oplus D^2 \oplus D^3}{1 \oplus D^3} = \frac{1 \oplus D \oplus D^2 \oplus D^4}{1 \oplus D^3} \\ &= \frac{1 \oplus D^2 \oplus D^3}{1 \oplus D \oplus D^2} \end{aligned}$$