



LUND  
UNIVERSITY

# Chapter 2: Linear systems & sinusoids

OVE EDFORS

DEPT. OF EIT, LUND UNIVERSITY



# Learning outcomes

---

After this lecture, the student should

- understand what a linear system is, including linearity conditions, their consequences, and the superposition principle,
- understand what time-invariance is, and its connection to linear systems – linear and time-invariant (LTI) systems,
- understand what impulses and step functions are and how they are defined,
- have a basic understanding of what generalized functions are,
- understand what the impulse response of an LTI system is and how it can be used to calculate output signals,
- be able to calculate the output signal of an LTI system, when given an input signal and the system's impulse response,
- understand causality and why it is an important property of real, physical, systems, and
- be able to find out if a certain given LTI system is causal or not.

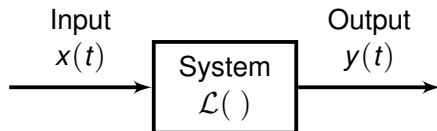
# Linear systems, what's that?

---



# Linear systems – Definition

---



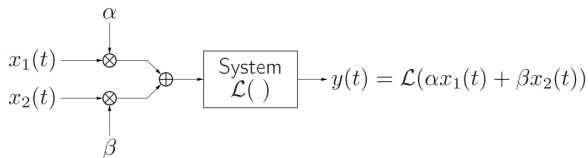
## Definition (Linear system)

A system  $\mathcal{L}$  is said to be **linear** if, whenever an input  $x(t) = x_1(t)$  yields an output  $y(t) = \mathcal{L}(x_1(t))$  and an input  $x(t) = x_2(t)$  yields an output  $y(t) = \mathcal{L}(x_2(t))$  this also leads to that we have  $\mathcal{L}(\alpha x_1(t) + \beta x_2(t)) = \alpha \mathcal{L}(x_1(t)) + \beta \mathcal{L}(x_2(t))$ , where  $\alpha$  and  $\beta$  are real or complex valued constants.

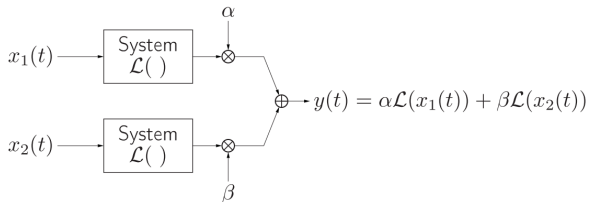
*Short version for memory:* Weighted sum of inputs leads to same weighted sum of outputs.

# Linear systems – Property illustration

This system with two superpositioned input signals



is **equivalent** to these two parallel systems, each with its own input, but outputs superpositioned the same way (same  $\alpha$  and  $\beta$ )



# The linearity condition – infinite sums

---

In order to extend the linearity condition

$$\mathcal{L}(\alpha x_1(t) + \beta x_2(t)) = \alpha \mathcal{L}(x_1(t)) + \beta \mathcal{L}(x_2(t))$$

to infinite sums (integrals – think Riemann sum) we must assume that the system  $\mathcal{L}$  is “sufficiently smooth or continuous”.

*In other words:* “Small” changes in input must give “small” changes in output.

# The linearity condition – equivalent formulation

---

The linearity condition

$$\mathcal{L}(\alpha x_1(t) + \beta x_2(t)) = \alpha \mathcal{L}(x_1(t)) + \beta \mathcal{L}(x_2(t))$$

can be replaced by two simpler conditions

$$\begin{aligned} \mathcal{L}(x_1(t) + x_2(t)) &= \mathcal{L}(x_1(t)) + \mathcal{L}(x_2(t)) && \text{[Additivity]} \\ \mathcal{L}(\alpha x(t)) &= \alpha \mathcal{L}(x(t)) && \text{[Homogeneity]} \end{aligned}$$

**Important note:** Both simplified conditions must be true for the first (linearity) condition to be true!

# Linear system properties

---





# A consequence of the linearity condition

---

Assuming that both input  $x_1(t)$  and  $x_2(t)$  are zero, i.e.,

$$x_1(t) = x_2(t) = 0$$

we obtain [additivity]

$$\mathcal{L}(0 + 0) = \mathcal{L}(0) + \mathcal{L}(0),$$

which can **only** be true if  $\mathcal{L}(0) = 0$ .

## An important consequence

**Zero input** results in **zero output** for ALL linear systems!

**Note:** This property is not true for nonlinear systems, which may e.g. “oscillate” despite a zero-input.

# Superposition principle

---

A system is said to be linear if *the output resulting from an input that is a weighted sum of signals is the same as the weighted sum of the outputs obtained when the input signals are acting separately*. This means

$$\mathcal{L}(\underbrace{\alpha x_1(t) + \beta x_2(t)}_{\text{Weighted sum of input signals}}) = \underbrace{\alpha \mathcal{L}(x_1(t)) + \beta \mathcal{L}(x_2(t))}_{\text{Corresponding weighted sum of output signals}}$$

**In other words:** Per definition, linear systems obey the superposition principle. (See definition of *linear system*.)

# Linear and time invariant systems

---



# Time invariance

---

A system  $\mathcal{I}$  is said to be *time-invariant* if a delay of the input by any amount of time only causes the same delay of the output. More precisely:

## Definition (Time invariance)

A system  $\mathcal{I}$  with output

$$y(t) = \mathcal{I}(x(t))$$

is time-invariant if

$$y(t - \tau) = \mathcal{I}(x(t - \tau))$$

holds for all delays (shifts)  $\tau$ .

**What it means:** The system as such does not change its behavior with time. Its components do not “grow old”.

# Linear and time-invariant (LTI) systems

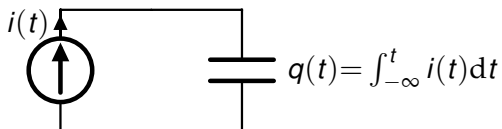
---

A system that is both **linear** and **time-invariant** is called an **LTI system**.

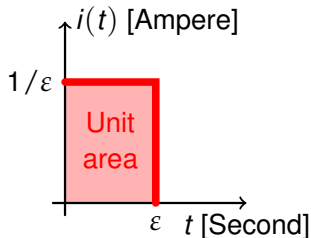
LTI systems are an important class of systems which have many applications in real life!

# A circuit example: Impulses and step functions

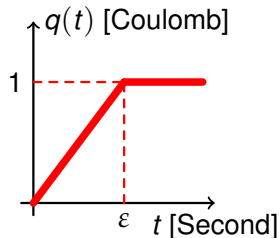
Consider the output when charging a capacitor with a current pulse:



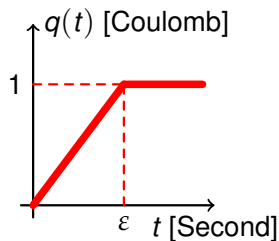
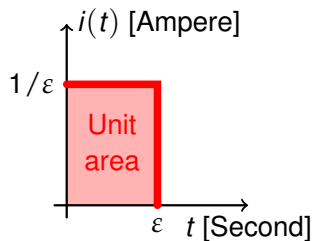
**Input signal**



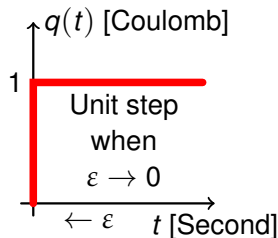
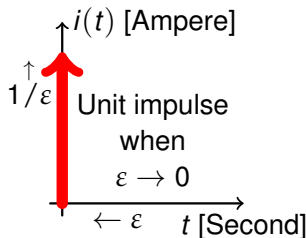
**Out signal**



# A circuit example: Impulses and step functions



Now, let the current pulse become shorter and shorter, approaching zero-length in time.



# Impulses, steps and impulse responses

---





# The delta/unit-impulse function

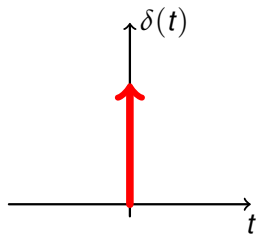
When the length of our pulse (with unit area) in the circuit example approaches zero, the pulse approaches the *delta-function*  $\delta(t)$  – also called the *unit impulse* or *Dirac's delta-function*<sup>1</sup>.

## Definition (Delta function)

The delta function  $\delta(t)$  (unit impulse) is a function that has the following two properties:

$$\delta(t) = 0 \text{ for all } t \neq 0$$

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$



**This is a very important function. We will use it throughout the course!**

<sup>1</sup>Named after Paul Dirac (1902-1984), one of the founders of quantum mechanics.

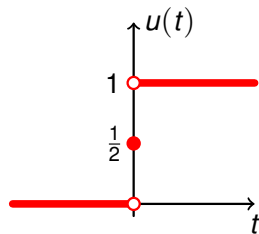
# The unit-step function

If a delta/unit-impulse function acts as input in our circuit example, the charge in the capacitor will rise immediately from 0 to 1 Coulomb, at  $t = 0$ , and stay there forever. We say that when the pulse length on the input approaches zero, the output (charge) approaches the *unit-step function*  $u(t)$  – also called Heaviside's step-function<sup>2</sup>.

## Definition (Unit-step function)

The unit-step function  $u(t)$  is defined as

$$u(t) = \begin{cases} 0 & , t < 0 \\ 1/2 & , t = 0 \\ 1 & , t > 0 \end{cases}$$



This is another very important function. We will use it throughout the course!

<sup>2</sup>Named after Oliver Heaviside (1850–1925), a famous English mathematician.

# Generalized functions and LTI systems

---



# Generalized functions / distributions

---

The delta function is not a function in the ordinary sense – it a *generalized function* or a *distribution*. These need to be treated in a special way.

The derivative of a generalized function/distribution is indirectly defined:

## Definition (Derivative of generalized function/distribution)

The derivative  $g'(t)$  of a generalized function/distribution  $g(t)$  is given by

$$\int_{-\infty}^{\infty} g'(t)\varphi(t)dt \stackrel{\text{def}}{=} - \int_{-\infty}^{\infty} g(t)\varphi'(t)dt$$

where  $\varphi(t)$  is a test function which is *infinitely differentiable* (smooth) and has *compact support* (zero outside some bounded interval).

# Important property of the delta function

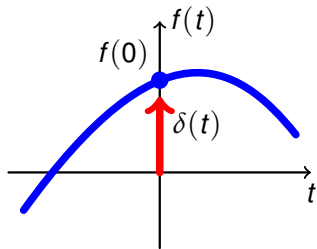
---

Using the definition of the delta function  $\delta(t)$  we can conclude that

$$\int_{-\infty}^{\infty} \delta(t)f(t)dt = f(0)$$

if  $f(t)$  is a function which is continuous at  $t = 0$ .

We can see this as the above integral *sampling* (picking up the value of) the function  $f(t)$  at  $t = 0$ .



# QUIZ: Sampling at a certain time

---

As we have seen, this expression

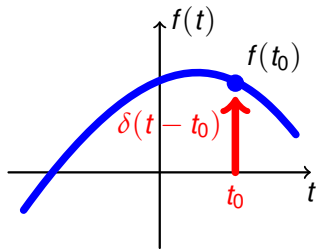
$$\int_{-\infty}^{\infty} \delta(t) f(t) dt = f(0)$$

accomplishes a sampling of  $f(t)$  at  $t = 0$ .

What would the corresponding expression sampling  $f(t)$  at an arbitrary time  $t = t_0$ ?

**Solution:** By moving (delaying) the  $\delta(t)$  function by  $t_0$  units along the  $t$ -axis, we get

$$\int_{-\infty}^{\infty} \delta(t - t_0) f(t) dt = f(t_0)$$



# Derivatives and integrals

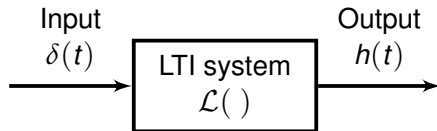
---

Between delta and step functions,  $\delta(t)$  and  $u(t)$ , we have the following relationships (derivative defined for generalized functions):

$$u'(t) = \delta(t)$$
$$\int_{-\infty}^t \delta(t) dt = u(t)$$

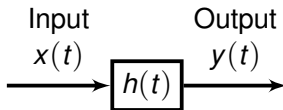
# LTI system impulse response

---



The **impulse response**  $h(t)$  of an LTI system is defined as the output when an (Dirac/unit) impulse at  $t = 0$  acts as input.

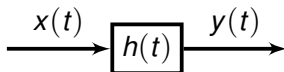
We often draw the block diagram of such as system:





# LTI system input-output relation

---



The input-output relation for an LTI system with impulse response  $h(t)$  is

$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau = \int_{-\infty}^{\infty} h(\tau)x(t - \tau)d\tau$$

where  $x(t)$  is the input and  $y(t)$  the corresponding output.

This particular integral is called a **convolution** and is *one of the most important mathematical concepts in the course*. Convolution is commutative and denoted by  $*$ .

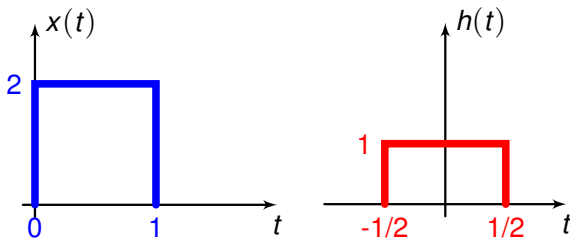
We write

$$y(t) = x(t) * h(t) = h(t) * x(t).$$

# Example 1

---

Assume we have an LTI system with input  $x(t)$  and impulse response  $h(t)$ , according to the figures below.



What will the output  $y(t)$  be?

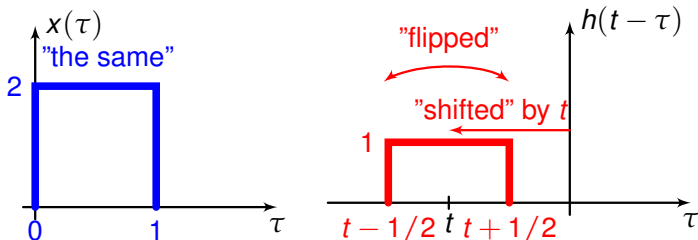
# Solution 1

The output of an LTI system with input  $x(t)$  and impulse response  $h(t)$  is determined by the convolution

$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau.$$

Let's start with performing the convolution by using the integral.

First, what does the factors of the integrand,  $x(\tau)$  and  $h(t - \tau)$ , look like on the  $\tau$ -axis?

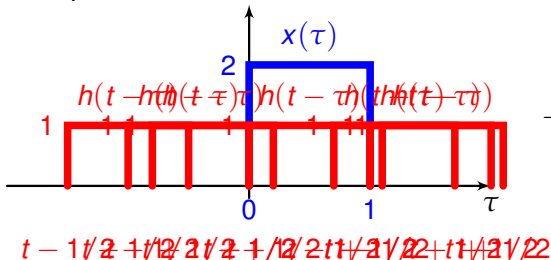


# Solution 1 (cont.)

Result of

$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$

Let's perform the convolution:



$$t < -1/2:$$

$$y(t) = 0$$

$$-1/2 \leq t < 1/2:$$

$$y(t) = 2(t + 1/2)$$

$$1/2 \leq t < 3/2:$$

$$y(t) = 2(3/2 - t)$$

$$t \geq 3/2:$$

$$y(t) = 0$$

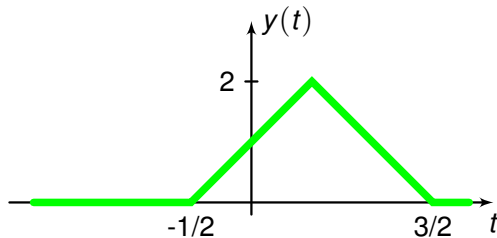
# Solution 1 (cont.)

---

Plotting the graph of the output signal

$$y(t) = \begin{cases} 0, & t < -1/2 \\ 2(t + 1/2), & -1/2 \leq t < 1/2 \\ 2(3/2 - t), & 1/2 \leq t < 3/2 \\ 0, & t \geq 3/2 \end{cases}$$

we get



## Example 2

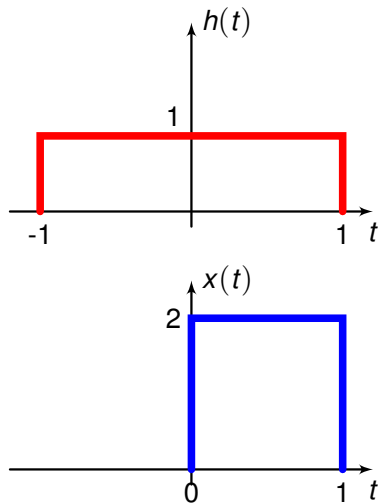
---

Consider an LTI system with impulse response

$$h(t) = \begin{cases} 1, & -1 \leq t \leq 1 \\ 0, & \text{otherwise} \end{cases}$$

and input signal

$$x(t) = \begin{cases} 2, & 0 \leq t \leq 1 \\ 0, & \text{otherwise} \end{cases}$$

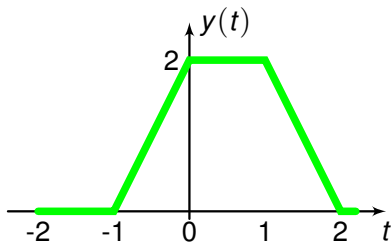


## Solution 2

---

Following the same type of reasoning as in Example 1, we get a piece-wise linear output (since  $x(t)$  and  $h(t)$  are piece-wise constant), but now with four corners rather than three (since  $x(t)$  and  $h(t)$  are of different length):

$$y(t) = \begin{cases} 0, & t < -1 \\ 2t + 2, & -1 \leq t < 0 \\ 2, & 0 \leq t < 1 \\ -2t + 4, & 1 \leq t < 2 \\ 0, & t \geq 2 \end{cases}$$



## Example 3

---

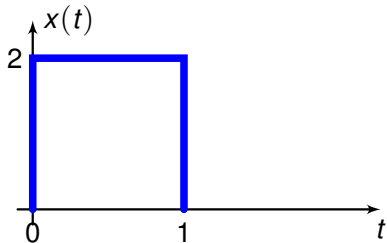
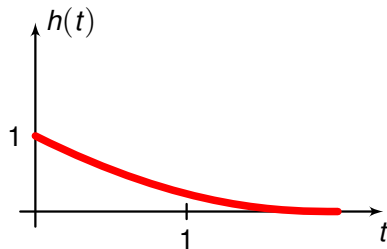
Consider an LTI system with impulse response

$$h(t) = e^{-t}u(t)$$

and the same input signal

$$x(t) = \begin{cases} 2, & 0 \leq t \leq 1 \\ 0, & \text{otherwise} \end{cases}$$

as in Example 2.

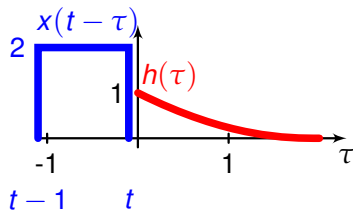




## Solution 3

With the given impulse response and input signal, we get three cases when calculating the convolution  $y(t) = x(t) * h(t) = \int h(\tau)x(t-\tau)d\tau$ :

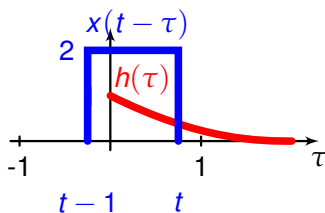
When  $t < 0$ :



the output becomes

$$y(t) = 0$$

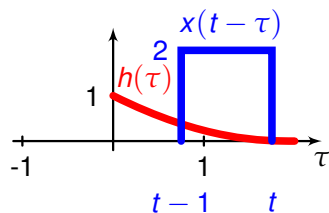
When  $0 \leq t < 1$ :



the output becomes

$$y(t) = \int_0^t 2e^{-\tau} d\tau$$

When  $t \geq 1$ :



the output becomes

$$y(t) = \int_{t-1}^t 2e^{-\tau} d\tau$$

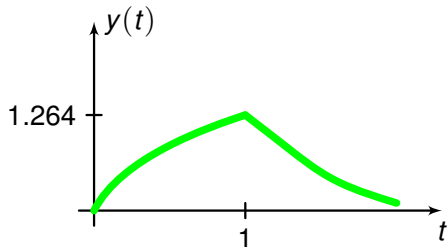
## Solution 3 (cont.)

---

Performing the integrations on the previous slide, we get the following output from our system:

$$y(t) = \begin{cases} 0, & t < 0 \\ 2(1 - e^{-t}), & 0 \leq t < 1 \\ 2(e - 1)e^{-t}, & t \geq 1 \end{cases}$$

which looks something like this in a plot:



# Causal systems

---

A system is said to be **causal** if the reaction always comes after the action<sup>3</sup>, i.e., if the output  $y(t_0)$  at any given time  $t_0$  is only influenced by inputs up to this time  $t_0$ , i.e., influenced only by  $x(t)$  for  $t \leq t_0$ .

This means that for causal systems we have

$$y(t) = x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau = \int_{-\infty}^t x(\tau)h(t - \tau)d\tau$$

which implies that  $\tau$  in the second integral never exceeds  $t$ .

For the last equality to hold, it is necessary that all causal systems have impulse responses that fulfill

$$h(t) = 0 \text{ for } t < 0.$$

## Important property to remember for causal systems!

<sup>3</sup>All real, physical, systems have this property and what we know as speed of light,  $c$  m/s, is actually the speed of causality. If input and output are separated in space by  $d$  m, the delay is  $\geq d/c$  s.

# LECTURE SUMMARY

---

- Linear and time invariant (LTI) systems
  - Linearity
  - Superposition and linearity condition
  - Time invariance
- Impulses and step functions
  - Dirac delta function/unit impulse and its properties
  - Step function and its properties
  - Generalized functions/distributions
  - Relation between delta function and step function
- LTI systems and impulse response
  - What is the impulse response
  - How can we use it to calculate outputs
  - Convolution
  - Causality and causal systems



LUND  
UNIVERSITY