

# **Estimation Theory**

## **Fredrik Rusek**

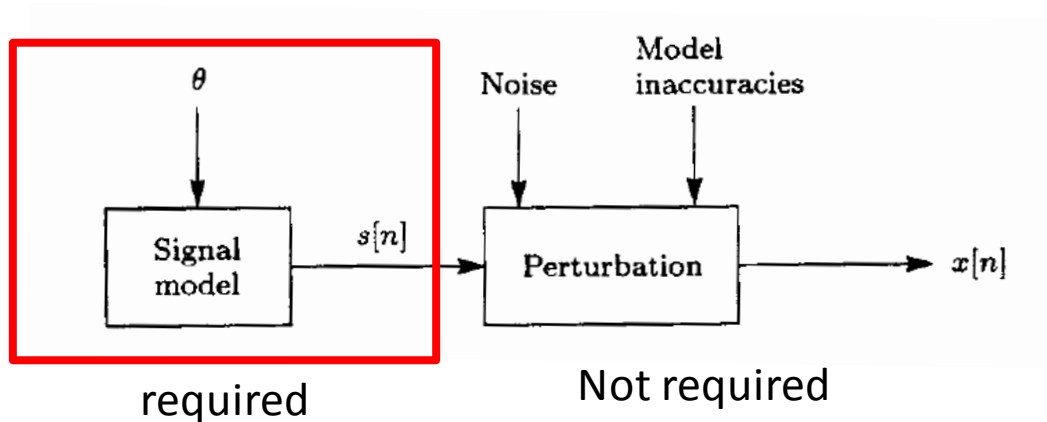
Chapters 8-9 + a bit more

# Chapter 8 – Least squares

Previous chapters: Probabilistic model of observations  $x[n]$

Least squares:

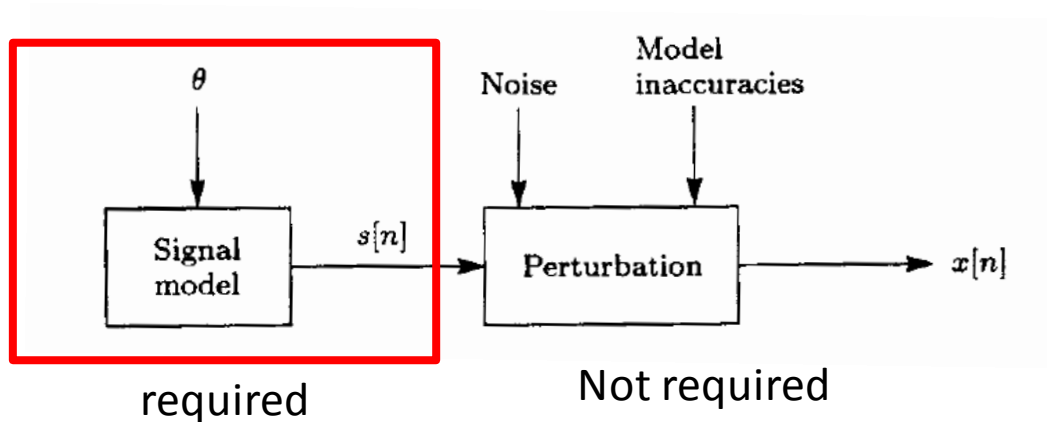
- No probabilistic assumption
- Signal model needed



# Chapter 8 – Least squares

## Definition of least squares

Choose  $\theta$  so that we minimize  $J(\theta) = \sum_{n=0}^{N-1} (x[n] - s[n])^2$



# Chapter 8 – Least squares

## Application areas of least squares

Observations are: deterministic signal + zero mean noise

# Chapter 8 – Least squares

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### Linear Least squares vs non-linear least squares

A linear LS problem is one where the  $s[n]$  is linear in the parameter  $\theta$  to be estimated

# Chapter 8 – Least squares

## Application areas of least squares

Observations are: deterministic signal + zero mean noise

### Linear Least squares vs non-linear least squares

A linear LS problem is one where the  $s[n]$  is linear in the parameter  $\theta$  to be estimated

$$s[n] = A \cos 2\pi f_0 n$$

Estimation of:

- $A$  is a linear LS problem
- $f_0$  is a non-linear estimation problem

# Chapter 8 – Least squares

## Application areas of least squares

Observations are: deterministic signal + zero mean noise

### Linear Least squares vs non-linear least squares

A linear LS problem is one where the  $s[n]$  is linear in the parameter  $\theta$  to be estimated

$$s[n] = A \cos 2\pi f_0 n$$

Estimation of:

- $A$  is a linear LS problem
- $f_0$  is a non-linear estimation problem
- Both  $A$  and  $f_0$  is linear in  $A$ , but non-linear in  $f_0$  (More about this later)

# Chapter 8 – Least squares

## Linear least squares: scalar case

Signal model

$$s[n] = \theta h[n]$$

Optimization problem

$$\min J(\theta) = \sum_{n=0}^{N-1} (x[n] - \theta h[n])^2$$



# Chapter 8 – Least squares

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Solution

$$\hat{\theta} = \frac{\sum_{n=0}^{N-1} x[n]h[n]}{\sum_{n=0}^{N-1} h^2[n]} \quad J_{\min} = J(\hat{\theta}) = \sum_{n=0}^{N-1} x^2[n] - \hat{\theta} \sum_{n=0}^{N-1} x[n]h[n]$$

# Chapter 8 – Least squares

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$$s[n] = \theta h[n]$$

Optimization problem

$$\min J(\theta) = \sum_{n=0}^{N-1} (x[n] - \theta h[n])^2$$

Clearly:

$$0 \leq J_{\min} \leq \sum_{n=0}^{N-1} x^2[n]$$

Solution

$$\hat{\theta} = \frac{\sum_{n=0}^{N-1} x[n]h[n]}{\sum_{n=0}^{N-1} h^2[n]} \quad J_{\min} = J(\hat{\theta}) = \sum_{n=0}^{N-1} x^2[n] - \hat{\theta} \sum_{n=0}^{N-1} x[n]h[n]$$

# Chapter 8 – Least squares

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Lots of noise

No noise

Clearly:

$$0 \leq J_{\min} \leq \sum_{n=0}^{N-1} x^2[n]$$

# Chapter 8 – Least squares

## Linear least squares: multivariate case

Signal model

$$\mathbf{s} = \mathbf{H}\boldsymbol{\theta}$$

Optimization problem

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

# Chapter 8 – Least squares

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Optimization problem

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

Gradient  $\frac{\partial J(\boldsymbol{\theta})}{\partial \boldsymbol{\theta}} = -2\mathbf{H}^T \mathbf{x} + 2\mathbf{H}^T \mathbf{H}\boldsymbol{\theta}$

Solution  $\hat{\boldsymbol{\theta}} = (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$

# Chapter 8 – Least squares

## Linear least squares: multivariate case

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Optimization problem

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When is the LS  
The BLUE?  
Efficient?

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# Chapter 8 – Least squares

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**When is the LS**  
**The BLUE?** Noise-cov is identity  
**Efficient?** Noise is Gaussian

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# Chapter 8 – Least squares

## Linear least squares: multivariate case

Minimum value

$$\begin{aligned} J_{\min} &= J(\hat{\boldsymbol{\theta}}) \\ &= (\mathbf{x} - \mathbf{H}\hat{\boldsymbol{\theta}})^T (\mathbf{x} - \mathbf{H}\hat{\boldsymbol{\theta}}) \\ &= (\mathbf{x} - \mathbf{H}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x})^T (\mathbf{x} - \mathbf{H}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}) \\ &= \mathbf{x}^T (\mathbf{I} - \mathbf{H}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T) (\mathbf{I} - \mathbf{H}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T) \mathbf{x} \\ &= \mathbf{x}^T (\mathbf{I} - \mathbf{H}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T) \mathbf{x}. \end{aligned}$$

Optimization problem

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

**When is the LS**  
**The BLUE?** Noise-cov is identity  
**Efficient?** Noise is Gaussian

Gradient

$$\frac{\partial J(\boldsymbol{\theta})}{\partial \boldsymbol{\theta}} = -2\mathbf{H}^T \mathbf{x} + 2\mathbf{H}^T \mathbf{H} \boldsymbol{\theta}$$

$$\text{Solution } \hat{\boldsymbol{\theta}} = (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$$



# Chapter 8 – Least squares

## Weighted Linear least squares: multivariate case

Optimization problem

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T \mathbf{W}(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

solution

$$\hat{\boldsymbol{\theta}} = (\mathbf{H}^T \mathbf{W} \mathbf{H})^{-1} \mathbf{H}^T \mathbf{W} \mathbf{x}$$

Is the BLUE if  $\mathbf{W}$  is taken as the noise covariance matrix

# Chapter 8 – Least squares

## Summary of linear LS

- Needs no probabilistic knowledge
- Needs observations that are **linear in parameter + zero mean noise**
- Is the BLUE if noise is uncorrelated
- If efficient if noise is white Gaussian
- Weighted linear LS estimators are BLUE if the weights are properly selected

Almost nothing new

# Chapter 8 – Least squares

## Tikhonov regularization

- If the matrix  $H$  is badly conditioned, the estimator tends to “blow up”
- To penalize large values of  $\theta$  (x below) Tikhonov regularization is commonly applied

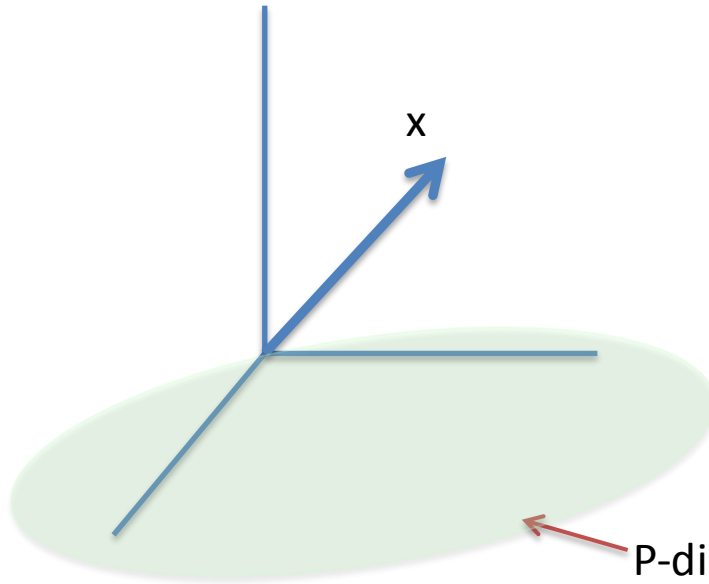
$$\|Ax - \mathbf{b}\|^2 + \|\Gamma\mathbf{x}\|^2$$

# Chapter 8 – Least squares

## Geometrical interpretation

$H$  is  $(N \times p)$  matrix

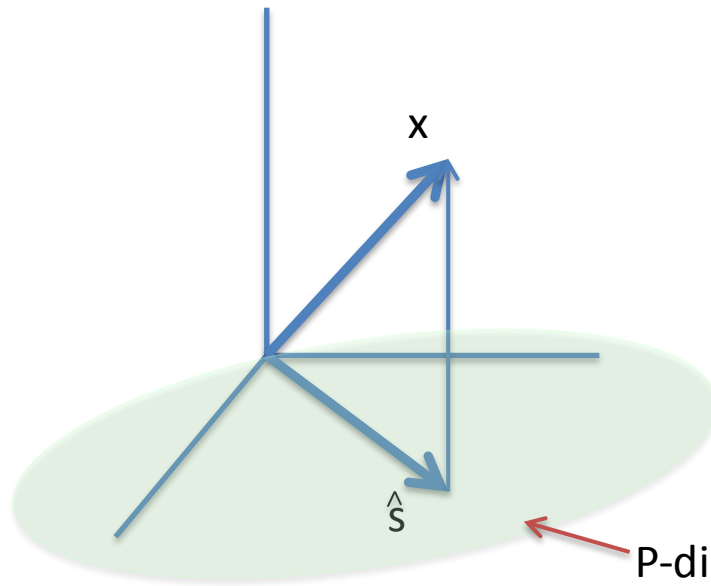
How to best approximate the  
N-dim observation  $x$  in the  
 $p$ -dimensional space



P-dimensional Subspace spanned by  $H$   
 $X$  (observations) lie in an  $N$ -dimensional space

# Chapter 8 – Least squares

## Geometrical interpretation



How to best approximate the  
N-dim observation  $x$  in the  
 $p$ -dimensional space

P-dimensional Subspace spanned by  $H$   
 $X$  (observations) lie in an N-dimensional space

# Chapter 8 – Least squares

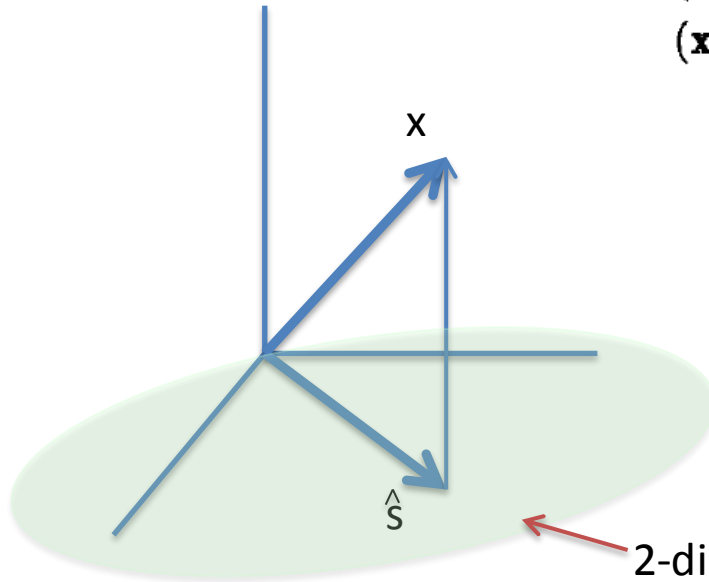
## Geometrical interpretation

$$(\mathbf{x} - \hat{\mathbf{s}}) \perp \mathbf{h}_1$$

$$(\mathbf{x} - \hat{\mathbf{s}}) \perp \mathbf{h}_2$$

$$(\mathbf{x} - \hat{\mathbf{s}})^T \mathbf{h}_1 = 0$$

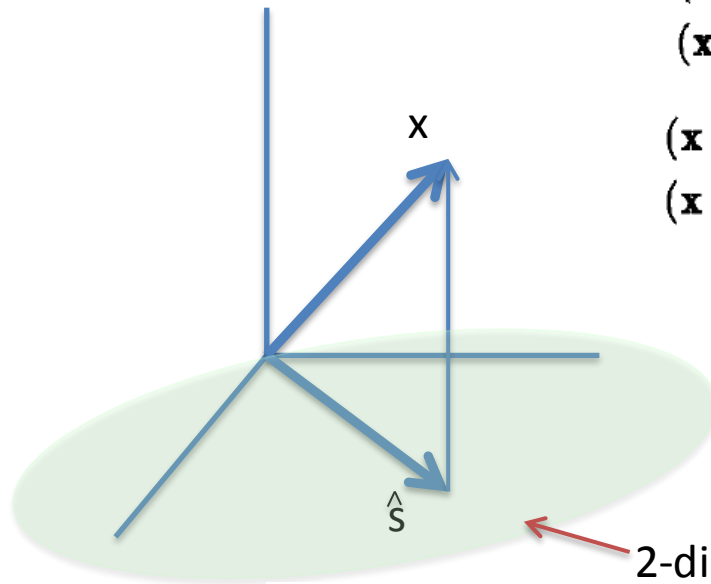
$$(\mathbf{x} - \hat{\mathbf{s}})^T \mathbf{h}_2 = 0$$



2-dimensional Subspace spanned by H  
X (observations) lie in an 3-dimensional space

# Chapter 8 – Least squares

## Geometrical interpretation



$$\hat{\mathbf{s}} = \theta_1 \mathbf{h}_1 + \theta_2 \mathbf{h}_2$$

$$(\mathbf{x} - \hat{\mathbf{s}}) \perp \mathbf{h}_1$$

$$(\mathbf{x} - \hat{\mathbf{s}}) \perp \mathbf{h}_2$$

$$(\mathbf{x} - \hat{\mathbf{s}})^T \mathbf{h}_1 = 0$$

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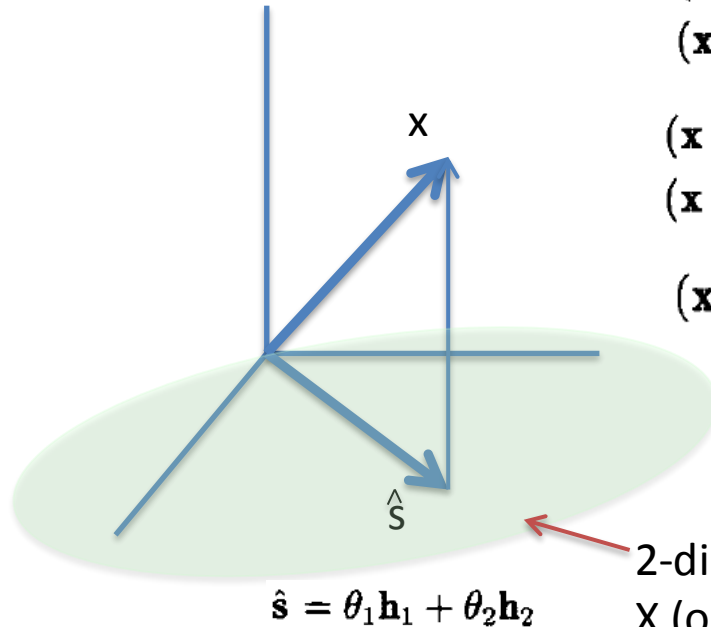
$$(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T \mathbf{h}_1 = 0$$

$$(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T \mathbf{h}_2 = 0$$

2-dimensional Subspace spanned by  $\mathbf{H}$   
 $\mathbf{X}$  (observations) lie in an 3-dimensional space

# Chapter 8 – Least squares

## Geometrical interpretation



$$(\mathbf{x} - \hat{\mathbf{s}}) \perp \mathbf{h}_1$$

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$$(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T \mathbf{h}_1 = 0$$

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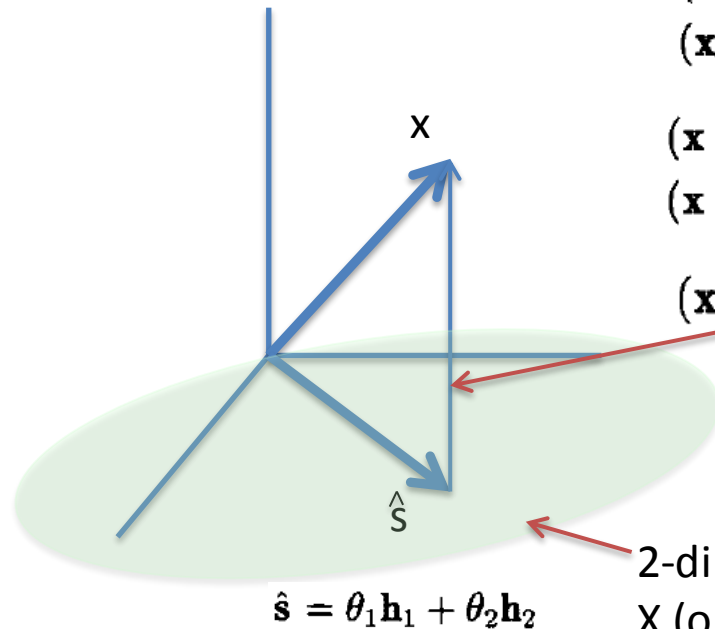
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# Chapter 8 – Least squares

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$$(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T \mathbf{h}_1 = 0$$

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$$\boldsymbol{\epsilon}^T \mathbf{H} = \mathbf{0}^T$$

2-dimensional Subspace spanned by H

X (observations) lie in an 3-dimensional space

**Orthogonality principle**

# Chapter 8 – Least squares

## Geometrical interpretation

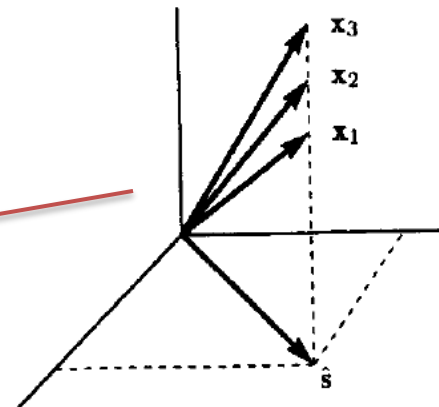
Estimator  $\hat{\boldsymbol{\theta}} = (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$

Projected signal  $\hat{\mathbf{s}} = \mathbf{H} \hat{\boldsymbol{\theta}} = \mathbf{H} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$

Projection matrix  $\mathbf{P} = \mathbf{H} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T$   
(projects a signal onto subspace spanned by H)

Properties:

1.  $\mathbf{P}^T = \mathbf{P}$
2.  $\mathbf{P}^2 = \mathbf{P}$
3. P is singular



# Chapter 8 – Least squares

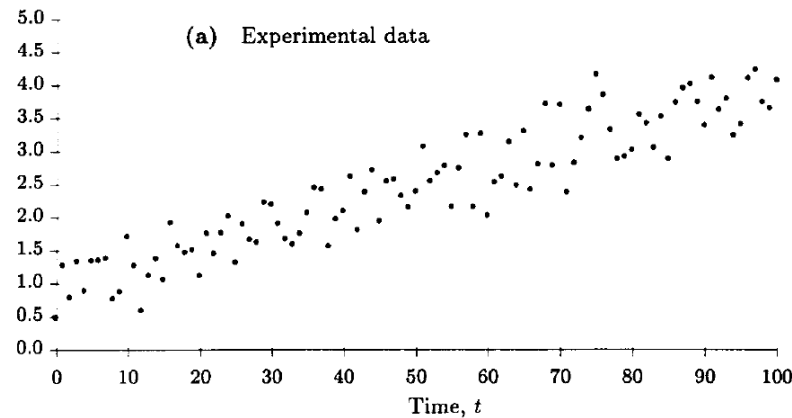
Section 8.6 and 8.7: good to know about, but not much more

# Chapter 8 – Least squares

Section 8.6 and 8.7: good to know about, but not much more

## Section 8.6

Fit this data to some  
Polynomial



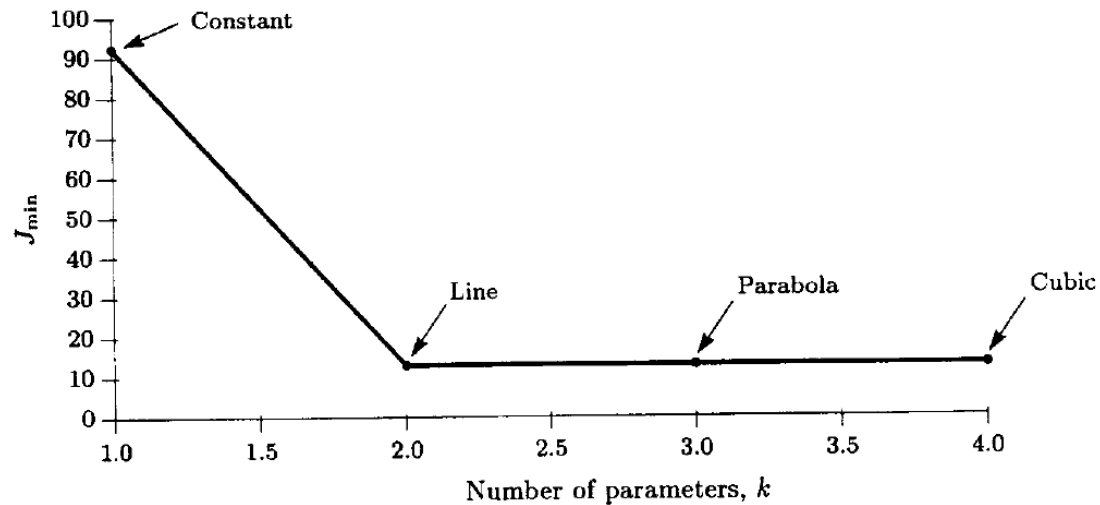
Should it be 1st, 2nd, 3d, or even higher order??

# Chapter 8 – Least squares

Section 8.6 and 8.7: good to know about, but not much more

## Section 8.6

Result for some  
Polynomial orders



There is a recursive method that finds the solution for order  $n$  given that  $n-1$  is already done

Hence, one can reuse the computations...not much more in this section

# Chapter 8 – Least squares

**Section 8.6 and 8.7: good to know about, but not much more**

## Section 8.7

Given the LS estimate for  $N-1$  observations, how to find the LS estimate with one additional observation?

A toolbox for this exist. This is in Section 8.7....not much more...

Scalar case (DC level in white noise)

$$\hat{A}[N] = \hat{A}[N - 1] + \frac{1}{N + 1} (x[N] - \hat{A}[N - 1])$$

$$J_{\min}[N] = J_{\min}[N - 1] + \frac{N}{N + 1} (x[N] - \hat{A}[N - 1])^2$$

# Chapter 8 – Least squares

## Section 8.8: constrained least squares

Suppose that the parameter vector is constrained to  $r < p$  constraints (A is  $r \times p$  matrix, rank= $r$ )

$$\mathbf{A}\boldsymbol{\theta} = \mathbf{b}$$

Optimization problem

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

$$\text{such that } \mathbf{A}\boldsymbol{\theta} = \mathbf{b}$$

# Chapter 8 – Least squares

## Section 8.8: constrained least squares

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

such that  $\mathbf{A}\boldsymbol{\theta} = \mathbf{b}$

Lagrangian

$$J_c = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T(\mathbf{x} - \mathbf{H}\boldsymbol{\theta}) + \boldsymbol{\lambda}^T(\mathbf{A}\boldsymbol{\theta} - \mathbf{b})$$



# Chapter 8 – Least squares

## Section 8.8: constrained least squares

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$$\frac{\partial J_c}{\partial \boldsymbol{\theta}} = -2\mathbf{H}^T \mathbf{x} + 2\mathbf{H}^T \mathbf{H}\boldsymbol{\theta} + \mathbf{A}^T \boldsymbol{\lambda}$$

# Chapter 8 – Least squares

## Section 8.8: constrained least squares

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

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$$\begin{aligned} \hat{\boldsymbol{\theta}}_c &= (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x} - \frac{1}{2} (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T \boldsymbol{\lambda} \\ &= \hat{\boldsymbol{\theta}} - (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T \frac{\boldsymbol{\lambda}}{2} \end{aligned}$$

# Chapter 8 – Least squares

## Section 8.8: constrained least squares

$$\min J(\boldsymbol{\theta}) = (\mathbf{x} - \mathbf{H}\boldsymbol{\theta})^T(\mathbf{x} - \mathbf{H}\boldsymbol{\theta})$$

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$$\mathbf{A}\boldsymbol{\theta}_c = \mathbf{A}\hat{\boldsymbol{\theta}} - \mathbf{A}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T \frac{\boldsymbol{\lambda}}{2} = \mathbf{b}$$

$$\frac{\boldsymbol{\lambda}}{2} = [\mathbf{A}(\mathbf{H}^T \mathbf{H})^{-1} \mathbf{A}^T]^{-1} (\mathbf{A}\hat{\boldsymbol{\theta}} - \mathbf{b}).$$

# Chapter 8 – Least squares

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# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

If the signal is not linear in  $\theta$  a non-quadratic optimization results, termed non-linear LS

$$J = (\mathbf{x} - \mathbf{s}(\boldsymbol{\theta}))^T (\mathbf{x} - \mathbf{s}(\boldsymbol{\theta}))$$

**Check-point: When is the non-linear LS also the MLE?**

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

Apart from numerical search, two methods are presented

### 1. Transformation of parameters

We try to find a function  $\alpha = \mathbf{g}(\theta)$  such that

$$\mathbf{s}(\theta(\alpha)) = \mathbf{H}\alpha$$

Solve the linear LS problem  $\hat{\alpha} = (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$

and plug back  $\hat{\theta} = \mathbf{g}^{-1}(\hat{\alpha})$

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

Example 8.9

Find  $A$  and  $\phi$ ,  $f_0$  is known

$$s[n] = A \cos(2\pi f_0 n + \phi) \quad n = 0, 1, \dots, N - 1$$

Problem is non-linear in the phase  $J = \sum_{n=0}^{N-1} (x[n] - A \cos(2\pi f_0 n + \phi))^2$

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

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Problem is non-linear in the phase  $J = \sum_{n=0}^{N-1} (x[n] - A \cos(2\pi f_0 n + \phi))^2$

but  $A \cos(2\pi f_0 n + \phi) = A \cos \phi \cos 2\pi f_0 n - A \sin \phi \sin 2\pi f_0 n$

$$\alpha_1 = A \cos \phi$$

$$\alpha_2 = -A \sin \phi$$

$$s[n] = \alpha_1 \cos 2\pi f_0 n + \alpha_2 \sin 2\pi f_0 n$$



# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

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$$\begin{aligned} \alpha_1 &= A \cos \phi \\ \alpha_2 &= -A \sin \phi \end{aligned} \quad s[n] = \alpha_1 \cos 2\pi f_0 n + \alpha_2 \sin 2\pi f_0 n$$

Signal is linear in  $\alpha_1 \alpha_2$

$$\mathbf{s} = \mathbf{H}\boldsymbol{\alpha}$$

$$\mathbf{H} = \begin{bmatrix} 1 & 0 \\ \cos 2\pi f_0 & \sin 2\pi f_0 \\ \vdots & \vdots \\ \cos 2\pi f_0(N-1) & \sin 2\pi f_0(N-1) \end{bmatrix}$$

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

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Estimator is  $\hat{\alpha} = (\mathbf{H}^T \mathbf{H})^{-1} \mathbf{H}^T \mathbf{x}$

$$A = \sqrt{\alpha_1^2 + \alpha_2^2}$$

Inverse mapping is  $\phi = \arctan\left(\frac{-\alpha_2}{\alpha_1}\right)$

$$\hat{\theta} = \begin{bmatrix} \sqrt{\hat{\alpha}_1^2 + \hat{\alpha}_2^2} \\ \arctan\left(\frac{-\hat{\alpha}_2}{\hat{\alpha}_1}\right) \end{bmatrix}$$

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

Apart from numerical search, two methods are presented

1. Transformation of parameters
2. Separable problems (half-linear, half non-linear)

With a signal model according to

$$\mathbf{s} = \mathbf{H}(\boldsymbol{\alpha})\boldsymbol{\beta}$$

Matrix, non-linear in  $\boldsymbol{\alpha}$

We get a linear problem in  $\boldsymbol{\beta}$ , but non-linear in  $\boldsymbol{\alpha}$ .

$$J(\boldsymbol{\alpha}, \boldsymbol{\beta}) = (\mathbf{x} - \mathbf{H}(\boldsymbol{\alpha})\boldsymbol{\beta})^T (\mathbf{x} - \mathbf{H}(\boldsymbol{\alpha})\boldsymbol{\beta})$$

# Chapter 8 – Least squares

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$$\hat{\boldsymbol{\beta}} = (\mathbf{H}^T(\boldsymbol{\alpha})\mathbf{H}(\boldsymbol{\alpha}))^{-1} \mathbf{H}^T(\boldsymbol{\alpha})\mathbf{x}$$

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$$\hat{\boldsymbol{\beta}} = (\mathbf{H}^T(\boldsymbol{\alpha})\mathbf{H}(\boldsymbol{\alpha}))^{-1} \mathbf{H}^T(\boldsymbol{\alpha})\mathbf{x}$$

$$\max \mathbf{x}^T \mathbf{H}(\boldsymbol{\alpha}) (\mathbf{H}^T(\boldsymbol{\alpha})\mathbf{H}(\boldsymbol{\alpha}))^{-1} \mathbf{H}^T(\boldsymbol{\alpha})\mathbf{x}$$

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

Example 8.10

Estimate  $\{A_1, A_2, A_3, r\}$

$$s[n] = A_1 r^n + A_2 r^{2n} + A_3 r^{3n}$$

# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

Example 8.10

Estimate  $\{A_1, A_2, A_3, r\}$

$$s[n] = A_1 r^n + A_2 r^{2n} + A_3 r^{3n}$$

Signal is linear in  $A_1, A_2, A_3$ , non-linear in  $r$



# Chapter 8 – Least squares

## Section 8.9: non-linear Least squares

Example 8.10

Estimate  $\{A_1, A_2, A_3, r\}$

$$s[n] = A_1 r^n + A_2 r^{2n} + A_3 r^{3n}$$

Signal is linear in  $A_1, A_2, A_3$ , non-linear in  $r$

$$\mathbf{H}(r) = \begin{bmatrix} 1 & 1 & 1 \\ r & r^2 & r^3 \\ \vdots & \vdots & \vdots \\ r^{N-1} & r^{2(N-1)} & r^{3(N-1)} \end{bmatrix}$$

LSE for the amplitudes

$$\hat{\boldsymbol{\beta}} = (\mathbf{H}^T(\hat{r})\mathbf{H}(\hat{r}))^{-1} \mathbf{H}^T(\hat{r})\mathbf{x}$$

LSE for  $r$

$$\max \mathbf{x}^T \mathbf{H}(r) (\mathbf{H}^T(r)\mathbf{H}(r))^{-1} \mathbf{H}^T(r)\mathbf{x}$$

# Chapter 9 – method of moments

Consider a Gaussian mixture pdf

$$p(x[n]; \epsilon) = \frac{1 - \epsilon}{\sqrt{2\pi\sigma_1^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_1^2}\right) + \frac{\epsilon}{\sqrt{2\pi\sigma_2^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_2^2}\right)$$

$$p(x[n]; \epsilon) = (1 - \epsilon)\phi_1(x[n]) + \epsilon\phi_2(x[n])$$

$$\phi_i(x[n]) = \frac{1}{\sqrt{2\pi\sigma_i^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_i^2}\right).$$

$\mathcal{N}(0, \sigma_1^2)$  PDF with probability  $1 - \epsilon$  and from a  $\mathcal{N}(0, \sigma_2^2)$  PDF with probability  $\epsilon$ .

# Chapter 9 – method of moments

Consider a Gaussian mixture pdf

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$\mathcal{N}(0, \sigma_1^2)$  PDF with probability  $1 - \epsilon$  and from a  $\mathcal{N}(0, \sigma_2^2)$  PDF with probability  $\epsilon$ .

Our methods to find MVU estimator for  $\epsilon$  will all fail

Options:

1. MLE
2. Method of moments (much simpler)

# Chapter 9 – method of moments

Consider a Gaussian mixture pdf

$$p(x[n]; \epsilon) = \frac{1 - \epsilon}{\sqrt{2\pi\sigma_1^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_1^2}\right) + \frac{\epsilon}{\sqrt{2\pi\sigma_2^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_2^2}\right)$$

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## Second moment

$$\begin{aligned} E(x^2[n]) &= \int_{-\infty}^{\infty} x^2[n] [(1 - \epsilon)\phi_1(x[n]) + \epsilon\phi_2(x[n])] dx[n] \\ &= (1 - \epsilon)\sigma_1^2 + \epsilon\sigma_2^2 \end{aligned}$$

# Chapter 9 – method of moments

Consider a Gaussian mixture pdf

$$p(x[n]; \epsilon) = \frac{1 - \epsilon}{\sqrt{2\pi\sigma_1^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_1^2}\right) + \frac{\epsilon}{\sqrt{2\pi\sigma_2^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_2^2}\right)$$

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**Second moment. Replace with its estimator**

$$\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] = (1 - \epsilon)\sigma_1^2 + \epsilon\sigma_2^2$$

$$\hat{\epsilon} = \frac{\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] - \sigma_1^2}{\sigma_2^2 - \sigma_1^2}$$

Not optimal, not unbiased in general, but simple to find

# Chapter 9 – method of moments

Consider a Gaussian mixture pdf

$$p(x[n]; \epsilon) = \frac{1 - \epsilon}{\sqrt{2\pi\sigma_1^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_1^2}\right) + \frac{\epsilon}{\sqrt{2\pi\sigma_2^2}} \exp\left(-\frac{1}{2} \frac{x^2[n]}{\sigma_2^2}\right)$$

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$$\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] = (1 - \epsilon)\sigma_1^2 + \epsilon\sigma_2^2$$

$$\hat{\epsilon} = \frac{\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] - \sigma_1^2}{\sigma_2^2 - \sigma_1^2}$$

As  $N \rightarrow \infty$ ,  $\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] \rightarrow E(x^2[n])$  so  $\hat{\epsilon} \rightarrow \epsilon$  as  $N \rightarrow \infty$ .

# Chapter 9 – method of moments

Method of moments, scalar case

***k*th moment**

$$\mu_k = E(x^k[n]) = h(\theta)$$

# Chapter 9 – method of moments

Method of moments, scalar case

***k*th moment**

$$\mu_k = E(x^k[n]) = h(\theta)$$

**Find inverse function**

$$\theta = h^{-1}(\mu_k)$$



# Chapter 9 – method of moments

Method of moments, scalar case

***k*th moment**

$$\mu_k = E(x^k[n]) = h(\theta)$$

**Find inverse function**

$$\theta = h^{-1}(\mu_k)$$

**Replace moment with estimate**

$$\hat{\mu}_k = \frac{1}{N} \sum_{n=0}^{N-1} x^k[n]$$

# Chapter 9 – method of moments

Method of moments, scalar case

***k*th moment**

$$\mu_k = E(x^k[n]) = h(\theta)$$

**Find inverse function**

$$\theta = h^{-1}(\mu_k)$$

**Replace moment with estimate**

$$\hat{\mu}_k = \frac{1}{N} \sum_{n=0}^{N-1} x^k[n]$$

**conclude**

$$\hat{\theta} = h^{-1} \left( \frac{1}{N} \sum_{n=0}^{N-1} x^k[n] \right)$$

# Chapter 9 – method of moments

Example 9.2: exponential pdf

$$\textit{Exponential pdf} \quad p(x[n]; \lambda) = \begin{cases} \lambda \exp(-\lambda x[n]) & x[n] > 0 \\ 0 & x[n] < 0. \end{cases}$$

$$\mu_1 = E(x[n]) = \frac{1}{\lambda}$$

# Chapter 9 – method of moments

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$$\mu_1 = E(x[n]) = \frac{1}{\lambda}$$

$$\hat{\lambda} = \frac{1}{\frac{1}{N} \sum_{n=0}^{N-1} x[n]}$$

*According to discussion yesterday, this is not MVU*

# Chapter 9 – method of moments

Method of moments, vector parameter

***Find  $p$  moments***

$$\mu_1 = h_1(\theta_1, \theta_2, \dots, \theta_p)$$

$$\mu_2 = h_2(\theta_1, \theta_2, \dots, \theta_p)$$

$\vdots$   $\vdots$   $\vdots$

$$\mu_p = h_p(\theta_1, \theta_2, \dots, \theta_p)$$

$$\boldsymbol{\mu} = \mathbf{h}(\boldsymbol{\theta})$$

***find inverse***

$$\boldsymbol{\theta} = \mathbf{h}^{-1}(\boldsymbol{\mu})$$

***conclude***

$$\hat{\boldsymbol{\theta}} = \mathbf{h}^{-1}(\hat{\boldsymbol{\mu}})$$

# Chapter 9 – method of moments

Method of moments, Performance

**Definition of method of moments**  $\hat{\theta} = g(\mathbf{T})$  where  $E(\mathbf{T}) = \boldsymbol{\mu}$

**Taylor expand around  $\boldsymbol{\mu}$**

$$\hat{\theta} = g(\mathbf{T}) \approx g(\boldsymbol{\mu}) + \sum_{k=1}^r \left. \frac{\partial g}{\partial T_k} \right|_{\mathbf{T}=\boldsymbol{\mu}} (T_k - \mu_k)$$

# Chapter 9 – method of moments

Method of moments, Performance

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**To the precision of the linearization**  $E(\hat{\theta}) = g(\boldsymbol{\mu})$

$$\begin{aligned} \text{var}(\hat{\theta}) &= E \left\{ \left[ g(\boldsymbol{\mu}) + \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T (\mathbf{T} - \boldsymbol{\mu}) - E(\hat{\theta}) \right]^2 \right\} = E \left\{ \left[ g(\boldsymbol{\mu}) + \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T (\mathbf{T} - \boldsymbol{\mu}) - g(\boldsymbol{\mu}) \right]^2 \right\} \\ &= E \left\{ \left[ \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T (\mathbf{T} - \boldsymbol{\mu}) \right]^2 \right\} = \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T \mathbf{C}_T \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}} \end{aligned}$$

# Chapter 9 – method of moments

Method of moments, exponential distribution continued

*Recall estimator*  $\hat{\lambda} = \frac{1}{\frac{1}{N} \sum_{n=0}^{N-1} x[n]}$

$$\text{var}(\hat{\theta}) = \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T \mathbf{C}_T \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}} \quad E(\hat{\theta}) = g(\boldsymbol{\mu})$$



# Chapter 9 – method of moments

Method of moments, exponential distribution continued

**Recall estimator**  $\hat{\lambda} = \frac{1}{\frac{1}{N} \sum_{n=0}^{N-1} x[n]}$

**Identify**  $\hat{\lambda} = g(T_1)$

$$T_1 = \frac{1}{N} \sum_{n=0}^{N-1} x[n] \quad g(T_1) = \frac{1}{T_1}$$

$$\text{var}(\hat{\theta}) = \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T \mathbf{C}_T \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}} \quad E(\hat{\theta}) = g(\boldsymbol{\mu})$$

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**We should do T.S.E at mean of T**

$$\mu_1 = E(T_1) = \frac{1}{\lambda}$$

$$\text{var}(\hat{\theta}) = \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T \mathbf{C}_T \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}} \quad E(\hat{\theta}) = g(\boldsymbol{\mu})$$

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$$\mu_1 = E(T_1) = \frac{1}{\lambda}$$

**Now, to the precision of the linear approximation**

$$E(\hat{\lambda}) = g(\mu_1) = \frac{1}{\frac{1}{\lambda}} = \lambda$$

$$\text{var}(\hat{\theta}) = \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}}^T \mathbf{C}_T \left. \frac{\partial g}{\partial \mathbf{T}} \right|_{\mathbf{T}=\boldsymbol{\mu}} \quad E(\hat{\theta}) = g(\boldsymbol{\mu})$$

# Chapter 9 – method of moments

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# Chapter 9 – method of moments

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$$\text{var}(\hat{\lambda}) = (-\lambda^2) \frac{1}{N\lambda^2} (-\lambda^2) = \frac{\lambda^2}{N}$$

$$\text{var}(\hat{\theta}) = \frac{\partial g}{\partial \mathbf{T}} \Big|_{\mathbf{T}=\boldsymbol{\mu}}^T \mathbf{C}_T \frac{\partial g}{\partial \mathbf{T}} \Big|_{\mathbf{T}=\boldsymbol{\mu}} \quad E(\hat{\theta}) = g(\boldsymbol{\mu})$$

# Chapter 9 – method of moments

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Fisher information  
Per sample:  $\lambda^2$

# Chapter 9 – method of moments

In practice it is common to use cumulants rather than moments.

Cumulants are compressing the moments effectively

Cumulant generating function is log of moment generating function, thus, a bit smaller in size and fluctuation” than the moments

<b>Cumulants</b>	$\kappa_2 = \mu_2$		
	$\kappa_3 = \mu_3$		
	$\kappa_4 = \mu_4 - 3\mu_2^2$		<b>moments</b>
	$\kappa_5 = \mu_5 - 10\mu_3\mu_2$		
	$\kappa_6 = \mu_6 - 15\mu_4\mu_2 - 10\mu_3^2 + 30\mu_2^3$		

# Chapter 9 – method of moments

Example, PIMRC 2014

System model:  $y=x+n$ ,  
X is M-QAM, n is noise.

Task: estimate M

Solution: linear combination of cumulants

Used cumulants

$$C_{20} = M_{20}$$

$$C_{40} = M_{40} - 3M_{20}^2$$

$$C_{41} = M_{40} - 3M_{20}M_{21}$$

$$C_{42} = M_{42} - \text{abs}(M_{20})^2 - 2M_{21}^2$$

$$C_{60} = M_{60} - 15M_{20}M_{40} + 30M_{20}^3$$

$$C_{61} = M_{61} - 5M_{21}M_{40} - 10M_{20}M_{41} + 30M_{20}^2M_{21}$$

$$C_{62} = M_{62} - 6M_{20}M_{42} - 8M_{21}M_{41} - M_{22}M_{40} + 6M_{20}^2M_{22} + 24M_{21}^2M_{20}$$

$$C_{63} = M_{63} - 9M_{21}M_{42} + 12M_{21}^3 - 3M_{20}M_{43} - 3M_{22}M_{41} + 18M_{20}M_{21}M_{22}$$

(3)

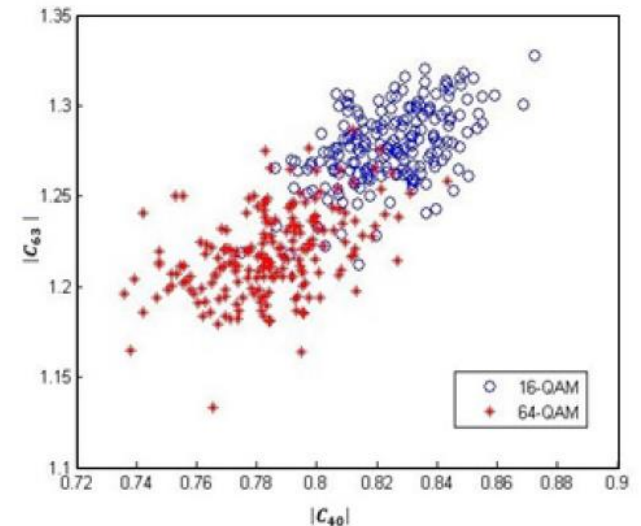


Figure 3: 200 realizations of 16-QAM and 64-QAM



# Generalized method of moments

Fairly new method. Nobel prize in economics 2013.

Find function  $g(\cdot, \cdot)$  so that  $m(\theta_0) \equiv E[g(Y_t, \theta_0)] = 0$

Replace function by its estimator  $\hat{m}(\theta) \equiv \frac{1}{T} \sum_{t=1}^T g(Y_t, \theta)$

Minimize the norm of  $\hat{m}(\theta)$   $\|\hat{m}(\theta)\|_W^2 = \hat{m}(\theta)' W \hat{m}(\theta)$ ,

Example: reciprocity calibration of MaMIMO.....

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Minimize the norm of  $\hat{m}(\theta)$   $\|\hat{m}(\theta)\|_W^2 = \hat{m}(\theta)' W \hat{m}(\theta)$ ,

Optimal weights:  $W \propto \Omega^{-1}$

$$\Omega = E[g(Y_t, \theta_0)g(Y_t, \theta_0)']$$