# Estimation Theory Fredrik Rusek

Extra lecture





Figure 14.1 Decision-making process in estimator selection



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Consider the following complex-valued estimation problem

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

Where all quantities are complex-valued (~ means complex in Kay)

In the real-valued case we can take the differential with respect to A, set to 0, and solve

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In the real-valued case we can take the differential with respect to A, set to 0, and solve

To take a differential, we must go back to the definition of a differential

$$f'(z_0) = \lim_{z \to z_0} \frac{f(z) - f(z_0)}{z - z_0}$$

Note: z is assumed to be complex valued

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Or, in other words: An analytical function allows one to take differentials

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When this is true, the function f(z) is said to be **analytic/holomorphic** 

A function is analytical when the Cauchy-Riemann equations are satisfied

$$f(z) = f(x+iy) = u(x,y) + iv(x,y)$$

$$\frac{\partial u}{\partial x} = \frac{\partial v}{\partial y}$$
 and  $\frac{\partial u}{\partial y} = -\frac{\partial v}{\partial x}$ 

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

Let us now return to the complex-valued estimation problem

Is it analytical so that we can take differentials??

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Let us now return to the complex-valued estimation problem

Is it analytical so that we can take differentials??

No. Since v(x,y) = 0, we have that dv/dx=dv/dy=0 The only real-valued functions that are analytical are constant functions

 $- \cap$ 

f(z) = f(x+iy) = u(x,y) + iv(x,y)  
$$\frac{\partial u}{\partial x} = \frac{\partial v}{\partial y} \quad \text{and} \quad \frac{\partial u}{\partial y} = -\frac{\partial v}{\partial x}$$

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

All estimation (in general: optimization) problems have real-valued cost functions

#### We must abandon the normal definition of differential

$$f'(z_0) = \lim_{z \to z_0} \frac{f(z) - f(z_0)}{z - z_0}$$

More fundamentally, we cannot regard z as being one number. We must consider it to be two numbers: its real-part and its complex-part.

We will do two things next:

- 1. Deal with the complex case as a multivariate optimization problem.
- 2. Show that with clever book-keeping, this multivariate treatment can be written in a form that resembles normal real-valued algebra.

**<u>Recall</u>**: In no cases will the differential be found through the formula

$$f'(z_0) = \lim_{z \to z_0} \frac{f(z) - f(z_0)}{z - z_0}$$

Expressing  $\tilde{A} = A_R + iA_I$  etc, we have

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

$$J'(A_R, A_I) = \sum_{n=0}^{N-1} |x_R[n] + jx_I[n] - (A_R + jA_I)(s_R[n] + js_I[n])|^2$$

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$$= \sum_{n=0}^{N-1} (x_R[n] - A_R s_R[n] + A_I s_I[n])^2 + (x_I[n] - A_R s_I[n] - A_I s_R[n])^2$$

 $|z_R+iz_1|^2 = |z_R|^2 + |z_1|^2$ 

Expressing  $A = A_R + iA_I$  etc, we have

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

$$J'(A_R, A_I) = \sum_{n=0}^{N-1} |x_R[n] + jx_I[n] - (A_R + jA_I)(s_R[n] + js_I[n])|^2$$

$$= \sum_{n=0}^{N-1} (x_R[n] - A_R s_R[n] + A_I s_I[n])^2 + (x_I[n] - A_R s_I[n] - A_I s_R[n])^2$$

This is a quadratic form in  $A_R$  and  $A_I$  Define  $\mathbf{A} = [A_R A_I]^T$ Some algebra will yield something of the form

$$J'(A_R,A_I) = \mathbf{b}^T \mathbf{A} + \mathbf{A}^T \mathbf{C} \mathbf{A}$$

Where **b** is a 2x1 vector and **C** is 2x2

We can now take normal differentials with respect to the real vector A

$$J'(A_{R'}A_{I}) = \mathbf{b}^{T}\mathbf{A} + \mathbf{A}^{T}\mathbf{C}\mathbf{A}$$

We can now take normal differentials with respect to the real vector A

We have seen this repeatedly in the course, and the final result is

$$\hat{\mathbf{A}} = \begin{bmatrix} \frac{\mathbf{s}_R^T \mathbf{x}_R + \mathbf{s}_I^T \mathbf{x}_I}{\mathbf{s}_R^T \mathbf{s}_R + \mathbf{s}_I^T \mathbf{s}_I} \\ \frac{\mathbf{s}_R^T \mathbf{x}_I - \mathbf{s}_I^T \mathbf{x}_R}{\mathbf{s}_R^T \mathbf{s}_R + \mathbf{s}_I^T \mathbf{s}_I} \end{bmatrix}$$

**x** is here a vector containing all x[n] (etc)

$$J'(A_R,A_I) = \mathbf{b}^T \mathbf{A} + \mathbf{A}^T \mathbf{C} \mathbf{A}$$

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$$\hat{\mathbf{A}} = \begin{bmatrix} \frac{\mathbf{s}_{R}^{T} \mathbf{x}_{R} + \mathbf{s}_{I}^{T} \mathbf{x}_{I}}{\mathbf{s}_{R}^{T} \mathbf{s}_{R} + \mathbf{s}_{I}^{T} \mathbf{x}_{I}} \\ \frac{\mathbf{s}_{R}^{T} \mathbf{x}_{I} - \mathbf{s}_{I}^{T} \mathbf{x}_{R}}{\mathbf{s}_{R}^{T} \mathbf{s}_{R} + \mathbf{s}_{I}^{T} \mathbf{s}_{I}} \end{bmatrix}$$
Going back to complex-valued notation, we get
$$\hat{A} = \frac{\mathbf{s}_{R}^{T} \mathbf{x}_{R} + \mathbf{s}_{I}^{T} \mathbf{x}_{I} + j \mathbf{s}_{R}^{T} \mathbf{x}_{I} - j \mathbf{s}_{I}^{T} \mathbf{x}_{R}}{\mathbf{s}_{R}^{T} \mathbf{s}_{R} + \mathbf{s}_{I}^{T} \mathbf{s}_{I}} = \frac{(\mathbf{x}_{R} + j \mathbf{x}_{I})^{T} (\mathbf{s}_{R} - j \mathbf{s}_{I})}{\mathbf{s}_{R}^{T} \mathbf{s}_{R} + \mathbf{s}_{I}^{T} \mathbf{s}_{I}} = \frac{\sum_{n=0}^{N-1} \tilde{x}[n] \tilde{s}^{*}[n]}{\sum_{n=0}^{N-1} |\tilde{s}[n]|^{2}}$$

We can now take normal differentials with respect to the real vector A

We have seen this repeatedly in the course, and the final result is

$$\hat{\mathbf{A}} = \begin{bmatrix} \frac{\mathbf{s}_R^T \mathbf{x}_R + \mathbf{s}_I^T \mathbf{x}_I}{\mathbf{s}_R^T \mathbf{s}_R + \mathbf{s}_I^T \mathbf{s}_I} \\ \frac{\mathbf{s}_R^T \mathbf{x}_I - \mathbf{s}_I^T \mathbf{x}_R}{\mathbf{s}_R^T \mathbf{s}_R + \mathbf{s}_I^T \mathbf{s}_I} \end{bmatrix}$$

This approach is valid, but in-efficient (a notational nightmare)

Going back to complex-valued notation, we get

$$\begin{aligned} \hat{\tilde{A}} &= \frac{\mathbf{s}_R^T \mathbf{x}_R + \mathbf{s}_I^T \mathbf{x}_I + j \mathbf{s}_R^T \mathbf{x}_I - j \mathbf{s}_I^T \mathbf{x}_R}{\mathbf{s}_R^T \mathbf{s}_R + \mathbf{s}_I^T \mathbf{s}_I} \\ &= \frac{(\mathbf{x}_R + j \mathbf{x}_I)^T (\mathbf{s}_R - j \mathbf{s}_I)}{\mathbf{s}_R^T \mathbf{s}_R + \mathbf{s}_I^T \mathbf{s}_I} \\ &= \frac{\sum_{n=0}^{N-1} \tilde{x}[n] \tilde{s}^*[n]}{\sum_{n=0}^{N-1} |\tilde{s}[n]|^2} \end{aligned}$$

Let us now gow back a step 
$$J( ilde{A}) = \sum_{n=0}^{N-1} | ilde{x}[n] - ilde{A} ilde{s}[n]|^2$$

We need to compute 
$$\frac{\partial J}{\partial A_R}$$
 and  $\frac{\partial J}{\partial A_I}$ 

Then, we should set them both to 0

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Then, we should set them both to 0

To keep track of both of them, we can form the quantity

$$\frac{\partial J}{\partial \tilde{A}} = \frac{\partial J}{\partial A_R} + j \frac{\partial J}{\partial A_I}$$

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To keep track of both of them, we can form the quantity



However, to obtain some nice formulas later, Kay (but not all other authors) choose to keep track of the differentials as

$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

Note that we have

$$\frac{\partial J}{\partial \tilde{A}} = 0$$
 if and only if  $\frac{\partial J}{\partial A_R} = 0$  and  $\frac{\partial J}{\partial A_I} = 0$ 

With the new book-keeping, our definition of differential "behaves as normal"

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With the new book-keeping, our definition of differential "behaves as normal"

**RECALL AGAIN** 
$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$
 **IS NOT**  $f'(z_0) = \lim_{z \to z_0} \frac{f(z) - f(z_0)}{z - z_0}$ 

$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

To efficiently use 
$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$
 we

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we need some properties of it

Let us consider first  $\frac{\partial \theta}{\partial \theta}$ 

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we need some properties of it

Let us consider first 
$$\frac{\partial \theta}{\partial \theta} = \frac{1}{2} \left( \frac{\partial}{\partial \alpha} - j \frac{\partial}{\partial \beta} \right) (\alpha + j\beta)$$

To efficiently use

$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

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$$\begin{array}{ll} \displaystyle \frac{\partial \theta}{\partial \theta} & = & \displaystyle \frac{1}{2} \left( \frac{\partial}{\partial \alpha} - j \frac{\partial}{\partial \beta} \right) (\alpha + j\beta) \\ \\ & = & \displaystyle \frac{1}{2} \left( \frac{\partial \alpha}{\partial \alpha} + j \frac{\partial \beta}{\partial \alpha} - j \frac{\partial \alpha}{\partial \beta} + \frac{\partial \beta}{\partial \beta} \right) \end{array}$$

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$$\begin{aligned} \frac{\partial \theta}{\partial \theta} &= \frac{1}{2} \left( \frac{\partial}{\partial \alpha} - j \frac{\partial}{\partial \beta} \right) (\alpha + j\beta) \\ &= \frac{1}{2} \left( \frac{\partial \alpha}{\partial \alpha} + j \frac{\partial \beta}{\partial \alpha} - j \frac{\partial \alpha}{\partial \beta} + \frac{\partial \beta}{\partial \beta} \right) \\ &= \frac{1}{2} (1 + j0 - j0 + 1) \\ &= 1 \end{aligned}$$

To efficiently use

$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

we need some properties of it

Let us now consider  $\frac{\partial \theta^*}{\partial \theta}$ 

$$\frac{*}{2} = \frac{1}{2} \left( \frac{\partial}{\partial \alpha} - j \frac{\partial}{\partial \beta} \right) (\alpha + j\beta)$$

$$= \frac{1}{2} \left( \frac{\partial \alpha}{\partial \alpha} + j \frac{\partial \beta}{\partial \alpha} - j \frac{\partial \alpha}{\partial \beta} + \frac{\partial \beta}{\partial \beta} \right)$$

$$= \frac{1}{2} (1 + j0 - j0 + 1)$$

$$= 1$$

$$\frac{\partial \theta}{\partial \theta} = 1$$

To efficiently use

$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

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$$= \frac{1}{2} (1 + j0 - j0 + 1)$$
Change signs here
$$= 1$$

$$\frac{\partial \theta}{\partial \theta} = 1$$

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$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

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$$\begin{array}{rcl} & \displaystyle \frac{*}{\partial} & \displaystyle = & \displaystyle \frac{1}{2} \left( \frac{\partial}{\partial \alpha} - j \frac{\partial}{\partial \beta} \right) (\alpha - j\beta) \\ & \displaystyle = & \displaystyle \frac{1}{2} \left( \frac{\partial \alpha}{\partial \alpha} - j \frac{\partial \beta}{\partial \alpha} - j \frac{\partial \alpha}{\partial \beta} - \frac{\partial \beta}{\partial \beta} \right) \\ & \displaystyle = & \displaystyle \frac{1}{2} (1 - j0 - j0 - 1) \\ & \displaystyle = & \displaystyle 0 \end{array}$$

$$\frac{\partial \theta}{\partial \theta} = 1 \qquad \qquad \frac{\partial \theta^*}{\partial \theta} = 0$$
To efficiently use 
$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_R} \right)$$

 $\left(\frac{\partial J}{\partial A_I}\right)$  we need some properties of it

From the definition, the chain rule continues to hold

$$\frac{\partial}{\partial \theta} X(\theta) Y(\theta) = Y(\theta) \frac{\partial}{\partial \theta} X(\theta) + X(\theta) \frac{\partial}{\partial \theta} Y(\theta)$$

$$\frac{\partial \theta}{\partial \theta} = 1 \qquad \qquad \frac{\partial \theta^*}{\partial \theta} = 0$$

To efficiently use 
$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

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In particular, this implies that

$$\frac{\partial}{\partial \theta} \,\theta \theta^* = \frac{\partial \theta}{\partial \theta} \theta^* + \theta \frac{\partial \theta^*}{\partial \theta} = 1 \cdot \theta^* + \theta \cdot 0 = \theta^*$$

$$\frac{\partial\theta}{\partial\theta} = 1 \qquad \qquad \frac{\partial\theta^*}{\partial\theta} = 0 \qquad \qquad \frac{\partial}{\partial\theta} X(\theta)Y(\theta) = Y(\theta) \frac{\partial}{\partial\theta} X(\theta) + X(\theta) \frac{\partial}{\partial\theta} Y(\theta)$$

To efficiently use

$$\frac{\partial J}{\partial \tilde{A}} = \frac{1}{2} \left( \frac{\partial J}{\partial A_R} - j \frac{\partial J}{\partial A_I} \right)$$

we need some properties of it

$$\frac{\partial}{\partial\theta} \theta \theta^* = \theta^*$$

$$\frac{\partial\theta^*}{\partial\theta} = 1$$

$$\frac{\partial\theta^*}{\partial\theta} = 0$$

$$\frac{\partial}{\partial\theta} X(\theta)Y(\theta) = Y(\theta) \frac{\partial}{\partial\theta} X(\theta) + X(\theta) \frac{\partial}{\partial\theta} Y(\theta)$$

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

$$\frac{\partial\theta}{\partial\theta} = 1 \qquad \qquad \frac{\partial\theta^*}{\partial\theta} = 0 \qquad \qquad \frac{\partial}{\partial\theta} \theta\theta^* = \theta^*$$

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$$\frac{\partial\theta}{\partial\theta} = 1 \qquad \qquad \frac{\partial\theta^*}{\partial\theta} = 0 \qquad \qquad \frac{\partial}{\partial\theta}\,\theta\theta^* = \theta^*$$

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$

$$\begin{aligned} \frac{\partial J}{\partial \tilde{A}} &= \frac{\partial}{\partial \tilde{A}} \sum_{n=0}^{N-1} \left| \tilde{x}[n] - \tilde{A}\tilde{s}[n] \right|^2 \\ &= \sum_{n=0}^{N-1} \frac{\partial}{\partial \tilde{A}} \left( |\tilde{x}[n]|^2 - \tilde{x}[n] \tilde{A}^* \tilde{s}^*[n] - \tilde{A}\tilde{s}[n] \tilde{x}^*[n] + \tilde{A}\tilde{A}^* |\tilde{s}[n]|^2 \right) \end{aligned}$$

$$\frac{\partial\theta}{\partial\theta} = 1 \qquad \qquad \frac{\partial\theta^*}{\partial\theta} = 0 \qquad \qquad \frac{\partial}{\partial\theta} \theta\theta^* = \theta^*$$

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$
$$\frac{\partial J}{\partial \tilde{A}} = \frac{\partial}{\partial \tilde{A}} \sum_{n=0}^{N-1} \left| \tilde{x}[n] - \tilde{A}\tilde{s}[n] \right|^2$$
$$= \sum_{n=0}^{N-1} \frac{\partial}{\partial \tilde{A}} \left( |\tilde{x}[n]|^2 - \tilde{x}[n] \tilde{A}^* \tilde{s}^*[n] - \tilde{A}\tilde{s}[n] \tilde{x}^*[n] + \tilde{A}\tilde{A}^* |\tilde{s}[n]|^2 \right)$$
$$\frac{\partial \theta}{\partial \theta} = 1$$
$$\frac{\partial \theta^*}{\partial \theta} = 0$$
$$\frac{\partial}{\partial \theta} \theta \theta^* = \theta^*$$

$$J(\tilde{A}) = \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2$$
  
$$\frac{\partial J}{\partial \tilde{A}} = \frac{\partial}{\partial \tilde{A}} \sum_{n=0}^{N-1} \left| \tilde{x}[n] - \tilde{A}\tilde{s}[n] \right|^2$$
  
$$= \sum_{n=0}^{N-1} \frac{\partial}{\partial \tilde{A}} \left( |\tilde{x}[n]|^2 - \tilde{x}[n]\tilde{A}^*\tilde{s}^*[n] - \tilde{A}\tilde{s}[n]\tilde{x}^*[n] + \tilde{A}\tilde{A}^*|\tilde{s}[n]|^2 \right)$$
  
$$= \sum_{n=0}^{N-1} \left( 0 - 0 - \tilde{s}[n]\tilde{x}^*[n] + \tilde{A}^*|\tilde{s}[n]|^2 \right).$$

$$\frac{\partial\theta}{\partial\theta} = 1 \qquad \qquad \frac{\partial\theta^*}{\partial\theta} = 0 \qquad \qquad \frac{\partial}{\partial\theta} \theta\theta^* = \theta^*$$

$$\begin{split} J(\tilde{A}) &= \sum_{n=0}^{N-1} |\tilde{x}[n] - \tilde{A}\tilde{s}[n]|^2 \\ \frac{\partial J}{\partial \tilde{A}} &= \frac{\partial}{\partial \tilde{A}} \sum_{n=0}^{N-1} \left| \tilde{x}[n] - \tilde{A}\tilde{s}[n] \right|^2 \\ &= \sum_{n=0}^{N-1} \frac{\partial}{\partial \tilde{A}} \left( |\tilde{x}[n]|^2 - \tilde{x}[n] \tilde{A}^* \tilde{s}^*[n] - \tilde{A}\tilde{s}[n] \tilde{x}^*[n] + \tilde{A}\tilde{A}^* |\tilde{s}[n]|^2 \right) \\ &= \sum_{n=0}^{N-1} \left( 0 - 0 - \tilde{s}[n] \tilde{x}^*[n] + \tilde{A}^* |\tilde{s}[n]|^2 \right). \qquad \qquad \sum_{n=0}^{N-1} \tilde{x}[n] \tilde{s}^* \end{split}$$

$$\frac{\partial\theta}{\partial\theta} = 1 \qquad \qquad \frac{\partial\theta^*}{\partial\theta} = 0 \qquad \qquad \frac{\partial}{\partial\theta} \theta\theta^* = \theta^*$$

$$\widehat{\tilde{A}} = rac{\sum\limits_{n=0}^{N-1} \widetilde{x}[n] \widetilde{s}^*[n]}{\sum\limits_{n=0}^{N-1} |\widetilde{s}[n]|^2}$$

For vector valued parameters we can, in the same fashion, reach the formulas

$$\frac{\partial \mathbf{b}^{H} \boldsymbol{\theta}}{\partial \boldsymbol{\theta}} = \mathbf{b}^{*} \qquad \frac{\partial \boldsymbol{\theta}^{H} \mathbf{b}}{\partial \boldsymbol{\theta}} = \mathbf{0}$$
$$\frac{\partial \boldsymbol{\theta}^{H} \mathbf{A} \boldsymbol{\theta}}{\partial \boldsymbol{\theta}} = (\mathbf{A} \boldsymbol{\theta})^{*}, \quad \text{where } \mathbf{A}^{H} = \mathbf{A}$$

**Complex Gaussian PDFs** 

Let  $\tilde{x} = u + jv$  where u and v are independent Gaussians  $\mathcal{N}(\mu_u, \sigma^2/2)$  and  $\mathcal{N}(\mu_v, \sigma^2/2)$ 

#### **Complex Gaussian PDFs**

Let  $\tilde{x} = u + jv$  where u and v are independent Gaussians  $\mathcal{N}(\mu_u, \sigma^2/2)$  and  $\mathcal{N}(\mu_v, \sigma^2/2)$ 

Variance is defined as

$$\operatorname{var}(\tilde{x}) = E\left(|\tilde{x} - E(\tilde{x})|^2\right) = E(|\tilde{x}|^2) - |E(\tilde{x})|^2$$

#### **Complex Gaussian PDFs**

Let  $\tilde{x} = u + jv$  where u and v are independent Gaussians  $\mathcal{N}(\mu_u, \sigma^2/2)$  and  $\mathcal{N}(\mu_v, \sigma^2/2)$ 

Variance is defined as

$$var(\tilde{x}) = E(|\tilde{x} - E(\tilde{x})|^2) = E(|\tilde{x}|^2) - |E(\tilde{x})|^2$$
$$= E(|x_R|^2) + E(|x_I|^2) - E(x_R)^2 - E(x_I)^2$$
$$= var(x_R) + var(x_I)$$

**Complex Gaussian PDFs** 

Let  $\tilde{x} = u + jv$  where u and v are independent Gaussians  $\mathcal{N}(\mu_u, \sigma^2/2)$  and  $\mathcal{N}(\mu_v, \sigma^2/2)$ 

The pdf of u,v is 
$$p(u,v) = \frac{1}{\sqrt{2\pi\frac{\sigma^2}{2}}} \exp\left[-\frac{1}{2(\frac{\sigma^2}{2})}(u-\mu_u)^2\right]$$
  
Twice the variance  $\cdot \frac{1}{\sqrt{2\pi\frac{\sigma^2}{2}}} \exp\left[-\frac{1}{2(\frac{\sigma^2}{2})}(v-\mu_v)^2\right]$   
 $= \frac{1}{\pi\sigma^2} \exp\left[-\frac{1}{\sigma^2}\left((u-\mu_u)^2 + (v-\mu_v)^2\right)\right]$ 

**Complex Gaussian PDFs** 

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$$p(u,v) = \frac{1}{\sqrt{2\pi\frac{\sigma^2}{2}}} \exp\left[-\frac{1}{2(\frac{\sigma^2}{2})}(u-\mu_u)^2\right]$$
  
Square root  $\cdot \frac{1}{\sqrt{2\pi\frac{\sigma^2}{2}}} \exp\left[-\frac{1}{2(\frac{\sigma^2}{2})}(v-\mu_v)^2\right]$   
 $= \frac{1}{\pi\sigma^2} \exp\left[-\frac{1}{\sigma^2}\left((u-\mu_u)^2 + (v-\mu_v)^2\right)\right]$ 

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Define  $\tilde{\mu} = E(\tilde{x}) = \mu_u + j\mu_v$ Then  $p(\tilde{x}) = \frac{1}{\pi\sigma^2} \exp\left[-\frac{1}{\sigma^2}|\tilde{x} - \tilde{\mu}|^2\right]$  Variance (s

Variance (since  $var(\tilde{x}) = var(x_R) + var(x_I)$ )

**Complex Gaussian PDFs** 

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Define  $\tilde{\mu} = E(\tilde{x}) = \mu_u + j\mu_v$ Then  $p(\tilde{x}) = \frac{1}{\pi\sigma^2} \exp\left[-\frac{1}{\sigma^2}|\tilde{x} - \tilde{\mu}|^2\right]$  No square root **Theorem 15.1 (Complex Multivariate Gaussian PDF)** If a real random vector **x** of dimension  $2n \times 1$  can be partitioned as

$$\mathbf{x} = \begin{bmatrix} \mathbf{u} \\ \mathbf{v} \end{bmatrix} = \begin{bmatrix} n \times 1 \\ n \times 1 \end{bmatrix}$$

where  $\mathbf{u}, \mathbf{v}$  are real random vectors and  $\mathbf{x}$  has the PDF

$$\mathbf{x} \sim \mathcal{N}\left(\left[\begin{array}{cc}\boldsymbol{\mu}_{u}\\ \boldsymbol{\mu}_{v}\end{array}\right], \left[\begin{array}{cc}\mathbf{C}_{uu} & \mathbf{C}_{uv}\\ \mathbf{C}_{vu} & \mathbf{C}_{vv}\end{array}\right]\right)$$

and  $\mathbf{C}_{uu} = \mathbf{C}_{vv}$  and  $\mathbf{C}_{uv} = -\mathbf{C}_{vu}$ 

then defining the  $n \times 1$  complex random vector  $\tilde{\mathbf{x}} = \mathbf{u} + j\mathbf{v}$ ,  $\tilde{\mathbf{x}}$  has the complex multivariate Gaussian PDF

$$ilde{\mathbf{x}} \sim \mathcal{CN}( ilde{m{\mu}}, \mathbf{C}_{ ilde{m{x}}})$$

where

$$\begin{aligned} \tilde{\boldsymbol{\mu}} &= \boldsymbol{\mu}_u + j \boldsymbol{\mu}_v \\ \mathbf{C}_{\tilde{x}} &= 2(\mathbf{C}_{uu} + j \mathbf{C}_{vu}) \end{aligned}$$

or more explicitly

$$p(\tilde{\mathbf{x}}) = \frac{1}{\pi^n \det(\mathbf{C}_{\bar{x}})} \exp\left[-(\tilde{\mathbf{x}} - \tilde{\boldsymbol{\mu}})^H \mathbf{C}_{\bar{x}}^{-1} (\tilde{\mathbf{x}} - \tilde{\boldsymbol{\mu}})\right].$$
(15.22)

**Theorem 15.1 (Complex Multivariate Gaussian PDF)** If a real random vector **x** of dimension  $2n \times 1$  can be partitioned as

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and  $\mathbf{C}_{uu} = \mathbf{C}_{vv}$  and  $\mathbf{C}_{uv} = -\mathbf{C}_{vu}$ 

Important remark: Not all vectors with complex Gaussian elements are multivariate complex Gaussian. Not even if the underlying real Gaussians are jointly Gaussian

It MUST hold that 
$$\mathbf{C}_{uu} = \mathbf{C}_{vv}$$
 and  $\mathbf{C}_{uv} = -\mathbf{C}_{vu}$ 

Meaning of this is, e.g., that the variance is equal for the real and imag parts (circularily symmetric)

#### Example

Let  $\tilde{\mathbf{x}} = \mathbf{H}\boldsymbol{\theta} + \tilde{\mathbf{w}}$   $\tilde{\mathbf{w}} \sim \mathcal{CN}(\mathbf{0}, \mathbf{C})$ 

Thus,  $\tilde{\mathbf{x}} \sim \mathcal{CN}(\mathbf{H}\boldsymbol{\theta}, \mathbf{C})$ 

#### **Find MLE**

Example

Let  $\tilde{\mathbf{x}} = \mathbf{H}\boldsymbol{\theta} + \tilde{\mathbf{w}}$   $\tilde{\mathbf{w}} \sim \mathcal{CN}(\mathbf{0}, \mathbf{C})$ 

Thus,  $\tilde{\mathbf{x}} \sim \mathcal{CN}(\mathbf{H}\boldsymbol{\theta}, \mathbf{C})$ 

The pdf is 
$$p(\tilde{\mathbf{x}}; \boldsymbol{\theta}) = \frac{1}{\pi^N \det(\mathbf{C})} \exp\left[-(\tilde{\mathbf{x}} - \mathbf{H}\boldsymbol{\theta})^H \mathbf{C}^{-1} (\tilde{\mathbf{x}} - \mathbf{H}\boldsymbol{\theta})\right]$$

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Take log, and differential

$$\frac{\partial \ln p(\tilde{\mathbf{x}}; \boldsymbol{\theta})}{\partial \boldsymbol{\theta}} = -\frac{\partial (\tilde{\mathbf{x}} - \mathbf{H}\boldsymbol{\theta})^H \mathbf{C}^{-1} (\tilde{\mathbf{x}} - \mathbf{H}\boldsymbol{\theta})}{\partial \boldsymbol{\theta}}$$

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Expand

$$J = \tilde{\mathbf{x}}^H \mathbf{C}^{-1} \tilde{\mathbf{x}} - \tilde{\mathbf{x}}^H \mathbf{C}^{-1} \mathbf{H} \boldsymbol{\theta} - \boldsymbol{\theta}^H \mathbf{H}^H \mathbf{C}^{-1} \tilde{\mathbf{x}} + \boldsymbol{\theta}^H \mathbf{H}^H \mathbf{C}^{-1} \mathbf{H} \boldsymbol{\theta}$$

$$\frac{\partial J}{\partial \theta} =$$

#### Example

$$\frac{\partial \mathbf{b}^{H} \boldsymbol{\theta}}{\partial \boldsymbol{\theta}} = \mathbf{b}^{*} \qquad \frac{\partial \boldsymbol{\theta}^{H} \mathbf{b}}{\partial \boldsymbol{\theta}} = \mathbf{0}$$
$$\frac{\partial \boldsymbol{\theta}^{H} \mathbf{A} \boldsymbol{\theta}}{\partial \boldsymbol{\theta}} = (\mathbf{A} \boldsymbol{\theta})^{*}, \text{ where } \mathbf{A}^{H} = \mathbf{A}$$

Expand 
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$$\frac{\partial J}{\partial \boldsymbol{\theta}} = \mathbf{0} - \left(\mathbf{H}^{H}\mathbf{C}^{-1}\tilde{\mathbf{x}}\right)^{*} - \mathbf{0} + \left(\mathbf{H}^{H}\mathbf{C}^{-1}\mathbf{H}\boldsymbol{\theta}\right)^{*}$$
$$\left[\mathbf{H}^{H}\mathbf{C}^{-1}(\tilde{\mathbf{x}} - \mathbf{H}\boldsymbol{\theta})\right]^{*}$$

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$$\left[\mathbf{H}^{H}\mathbf{C}^{-1}(\tilde{\mathbf{x}} - \mathbf{H}\boldsymbol{\theta})\right]^{*}$$
$$\hat{\boldsymbol{\theta}} = (\mathbf{H}^{H}\mathbf{C}^{-1}\mathbf{H})^{-1}\mathbf{H}^{H}\mathbf{C}^{-1}\tilde{\mathbf{x}}$$

I will briefly cover two parts (if time permits)

- 1. Karhunen-Loeve decomposition
- 2. Practical (mathematical) conversion

# **On Bandwidth**

DAVID SLEPIAN, FELLOW, IEEE

#### Abstract-It is easy to argue that real signals must be bandlimited. It is also easy to argue that they cannot be so.

#### THE DILEMMA

RE SIGNALS really bandlimited? They seem to be, and yet they seem not to be.

 $\angle$   $\Delta$  On the one hand, a pair of solid copper wires will not propagate electromagnetic waves at optical frequencies, and so the signals I receive over such a pair must be bandlimited. In fact, it makes little physical sense to talk of energy received over wires at frequencies higher than some finite cutoff W, say 10<sup>20</sup> Hz. It would seem, then, that signals must be bandlimited.

On the other hand, however, signals of limited bandwidth W are finite Fourier transforms,

# **On Bandwidth**

DAVID SLEPIAN, FELLOW, IEEE

$$s(t) = \int_{-W}^{W} e^{2\pi i f t} S(f) df$$

and irrefutable mathematical arguments show them to be extremely smooth. They possess derivatives of all orders. Indeed, such integrals are entire functions of t, completely predictable from any little piece, and they cannot vanish on any t interval unless they vanish everywhere. Such signals cannot start or stop, but must go on forever. Surely *real* signals start and stop, and so they cannot be bandlimited!

#### Karhunen-Loeve decomposition

Recall the operations of the MMSE estimator in discrete time:

#### It starts by whitening the observations

Correlated signals carries little information, so a clear goal of the conversion should be to produce uncorrelated samples

Karhunen-Loeve decomposition

We seek to expand the signal in the form

$$x(t) = \lim_{N \to \infty} \sum_{i=1}^{N} x_i \phi_i(t), \quad 0 \le t \le T,$$
$$x_i \triangleq \int_0^T x(t) \phi_i(t) dt.$$

We work with convergence in the MSE sense, i.e.

$$\lim_{N\to\infty} E\left[\left(x_t-\sum_{i=1}^N x_i\,\phi_i(t)\right)^2\right]=0, \quad 0\leq t\leq T.$$

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The constraint that the x<sub>i</sub>'s should be uncorrelated reads

$$E(x_i x_j) = \lambda_i \delta_{ij}.$$

Karhunen-Loeve decomposition

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$$x_i \triangleq \int_0^T x(t) \phi_i(t) dt.$$

1. The value  $x_i^2$  has a simple physical interpretation. It corresponds to the energy along the coordinate function  $\phi_i(t)$  in a particular sample function.

2. Similarly,  $E(x_i^2) = \lambda_i$  corresponds to the *expected* value of the energy along  $\phi_i(t)$ , assuming that  $m_i = 0$ . Clearly,  $\lambda_i \ge 0$  for all *i*.

#### Karhunen-Loeve decomposition

Let us now examine the requirements on the basis for having uncorrelated coefficients

$$\lambda_i \delta_{ij} = E(x_i x_j)$$

$$= E\left[\int_0^T x(t) \phi_i(t) dt \int_0^T x(u) \phi_j(u) du\right]$$

$$= \int_0^T \phi_i(t) dt \int_0^T K_x(t, u) \phi_j(u) du, \quad \text{for all } i \text{ and } j.$$

$$K_x(t, u) = E(x(t)x(u))$$

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$$= \lambda_j \phi_j(t)$$
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$$\lambda_j \phi_j(t) = \int_0^T K_x(t, u) \phi_j(u) du, \quad 0 \le t \le T.$$

Karhunen-Loeve decomposition

Most important properties:

$$E\left[\int_{0}^{T} x^{2}(t) dt\right] = \int_{0}^{T} K_{x}(t, t) dt = \sum_{i=1}^{\infty} \lambda_{i}$$
  
Mercer's theorem  $K_{x}(t, u) = \sum_{i=1}^{\infty} \lambda_{i} \phi_{i}(t) \phi_{i}(u), \quad 0 \le t, u \le T,$ 

$$\lambda_j \phi_j(t) = \int_0^T K_x(t, u) \phi_j(u) \, du, \qquad 0 \le t \le T.$$

Karhunen-Loeve decomposition

Most important properties:

**Reconstruction** 
$$\xi_N(t) \triangleq E\left[\left(x(t) - \sum_{i=1}^N x_i \phi_i(t)\right)^2\right]$$

for any  $\epsilon > 0$  there exists an  $N_1$  independent of t such that  $\xi_N(t) < \epsilon$  for all  $N > N_1$ 

$$\lambda_j \phi_j(t) = \int_0^T K_x(t, u) \phi_j(u) \, du, \qquad 0 \le t \le T.$$

Karhunen-Loeve decomposition

Most important properties:

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for any  $\epsilon > 0$  there exists an  $N_1$  independent of t such that  $\xi_N(t) < \epsilon$  for all  $N > N_1$ 

Thus, with  $N_1$  basis functions, we can represent x(t) with a small loss  $\varepsilon$ 

$$\lambda_j \phi_j(t) = \int_0^T K_x(t, u) \phi_j(u) \, du, \qquad 0 \le t \le T.$$

Karhunen-Loeve decomposition

A very important application area of Karhunen – Loeve: Time variant Gaussian process

For more details, see Van Trees 1968

Karhunen-Loeve decomposition

**Bandlimited processes** 

Spectrum of process: 
$$S_x(\omega) = \begin{cases} \frac{P}{2W}, & |f| \le W, \\ 0, & |f| > W, \end{cases}$$

**Observation time of the process: T seconds** 

Karhunen-Loeve decomposition

### **Bandlimited processes**

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$$S_x(\omega) = \begin{cases} \frac{P}{2W}, & |f| \le W, \\ 0, & |f| > W, \end{cases}$$

### **Observation time of the process: T seconds**

After some math 
$$K_x(t, u) = P \frac{\sin \alpha (t - u)}{\alpha (t - u)}$$

And we need to solve 
$$\lambda \phi(t) = \int_{-T/2}^{+T/2} P \frac{\sin \alpha(t-u)}{\alpha(t-u)} \phi(u) \, du.$$

Karhunen-Loeve decomposition

#### **Bandlimited processes**

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And we need to solve  $\lambda$ 

$$\phi(t) = \int_{-T/2}^{+T/2} P \frac{\sin \alpha(t-u)}{\alpha(t-u)} \phi(u) \, du$$

#### Solutions are well known: prolate spheroidal wave functions

A process of bandwidth W observed for T seconds

**Karhunen-Loeve decomposition** is 2WT+1 dimensional Interesting discovery 2WT = 2.552WT = 5.10 $\lambda_0 = 1.000 \frac{P}{2W}$  $\lambda_0 = 0.996 \frac{P}{2W}$ Eigenvaluesabove  $\lambda_1 = 0.999 \frac{P}{2W}$  $\lambda_1 = 0.912 \frac{P}{2W}$ 2WT+1 are almost zero  $\lambda_2 = 0.997 \frac{P}{2W}$  $\lambda_2 = 0.519 \frac{P}{2W}$  $\lambda_3 = 0.110 \frac{P}{2W}$  $\lambda_3 = 0.961 \frac{P}{2W}$  $\lambda_4 = 0.748 \frac{P}{2W}$  $\lambda_4 = 0.009 \frac{P}{2W}$  $\lambda_5 = 0.0004 \frac{P}{2W}$  $\lambda_5 = 0.321 \frac{p}{2W}$  $\lambda_6 = 0.061 \frac{P}{2W}$  $\lambda_7 = 0.006 \frac{P}{2W}$  $\lambda_8 = 0.0004 \frac{P}{2W}$ 

Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

A signal s(t) whose Fourier transform S(f) is limited to B (positive) Hz can be perfectly reconstructed from its samples  $s(nT_0)$ , n=...-1,0,1,2,... Where  $T_0 = 1/2B$ 

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Selecting the sampling rate as 2B yields a discrete time spectrum with folding here

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s[n] = s(n/2B) S(w) = ????



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s[n] = s(n/2B) S(w) = ????



Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

A signal s(t) whose Fourier transform S(f) is limited to B (positive) Hz can be perfectly reconstructed from its samples  $s(nT_0)$ , n=...-1,0,1,2,... Where  $T_0 = 1/2B$ 

s[n] = s(n/2B) S(w) = ????



Sampling theorem (  $\approx$  asymptotic version of the prolate spheroidal wave functions)





### Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise



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However, let us first start with noise whitening. This is invertable, so nothing is lost



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### Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

We can now apply the low-pass filter



### Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

Discrete time model will be



Thus, the spectrum of the discrete time model coincides with useful part of the continiuous signal spectrum. Optimal A2D

### Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

However, this LPF is hard to implement



#### Sampling theorem ( $\approx$ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

Let us filter with a realistic filter, and change the sampling rate



#### Sampling theorem ( $\approx$ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

After filtering (but before sampling)



### Sampling theorem ( $\approx$ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

Now sample



We should take the part to the right and fold back

#### Sampling theorem ( ~ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

#### Result



### Sampling theorem ( $\approx$ asymptotic version of the prolate spheroidal wave functions)

To handle this, we must first filter the signal in order to avoid to fold back any noise

### Result



Is it lossy or **not**? The fact that there is "more noise than necessary" can be handled by the discrete-time processing. System is over-sampled