



Detection Theory

Chapter 3. Statistical Decision Theory I.

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Outline



- Neyman-Pearson Theorem
- Detector Performance
- Irrelevant Data
- Minimum Probability of Error
- Bayes Risk
- Multiple Hypothesis Testing

Chapter Goal



- State basic statistical groundwork for detectors design
- Address simple hypothesis , where PDFs are known (Chapter 6 will cover the case of unknown PDFs)
- Introduction to the classical approach based on Neyman-Pearson and Bayes Risk
 - Bayesian methods employ prior knowledge of the hypotheses
 - Neyman-Pearson has no prior knowledge

Bayes vs Newman-Pearson



- The particular method employed depends upon our willingness to incorporate prior knowledge about the probabilities of occurrence of the various hypotheses
- Typical cases
 - Bayes : Communication systems, Pattern recognition
 - NP : Sonar, Radar

Nomenclature



$N(\mu, \sigma^2)$: Gaussian PDF with mean μ and variance σ^2

H_i : Hypothesis i

$p(x | H_i)$: Probability of observing x under assumption H_i

$P(H_i; H_j)$: Probability of detecting H_i , when it is H_j

$P(H_i | H_j)$: Probability of detecting H_i , when H_j was observed

$L(x)$: Likelihood of x

Nomenclature cont...



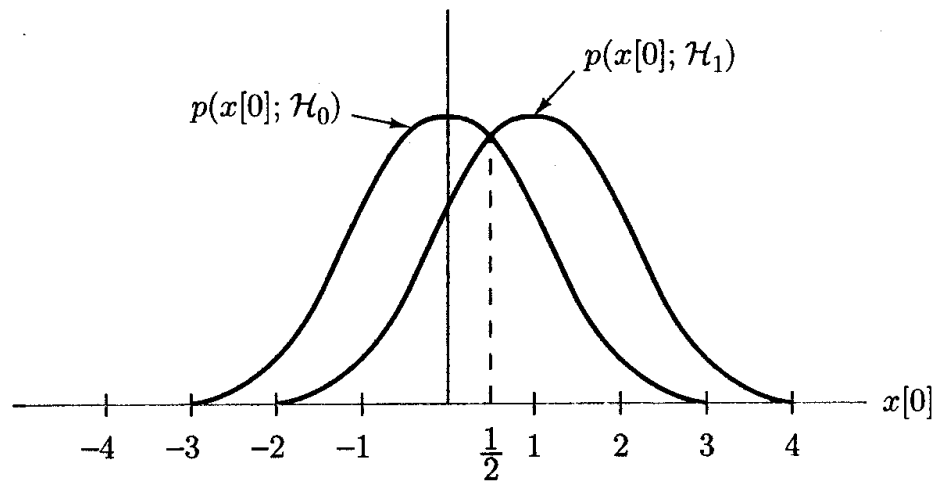
- Note that $P(H_i; H_j)$ is the probability of deciding \mathcal{H}_i , if \mathcal{H}_j is true.
- While $P(H_i | H_j)$ assumes that the outcome of a probabilistic experiment is observed to be \mathcal{H}_j and that the probability of deciding \mathcal{H}_i is conditioned to that outcome.

Neyman-Pearson Theorem



$$\mathcal{H}_0 : \mu = 0$$

$$\mathcal{H}_1 : \mu = 1$$



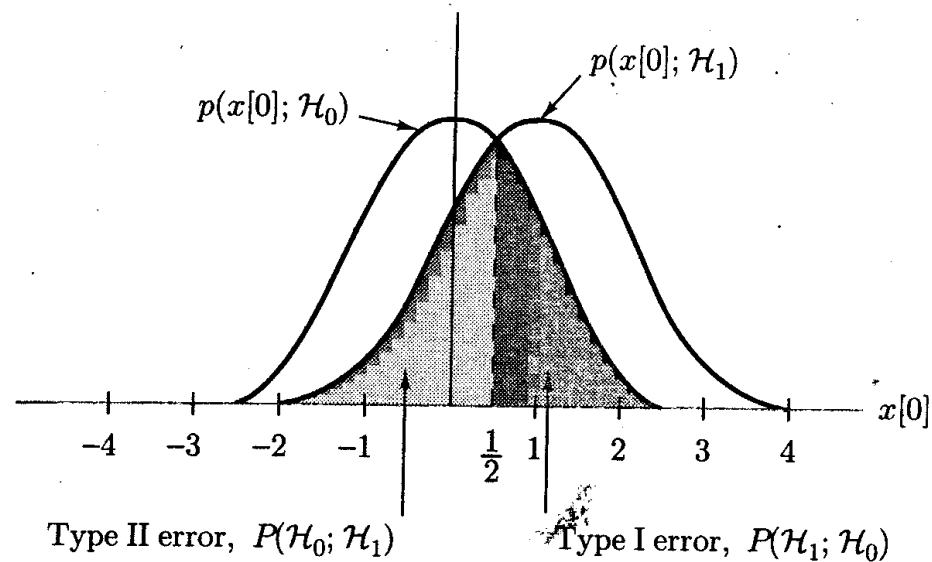
Neyman-Pearson Theorem cont...



$$H_0 : x[0] = w[0]$$

$$H_1 : x[0] = s[0] + w[0]$$

- It is not possible to reduce both error probabilities simultaneously



- In Neyman-Pearson the approach is to hold one error probability while minimizing the other.

Neyman-Pearson Theorem cont...



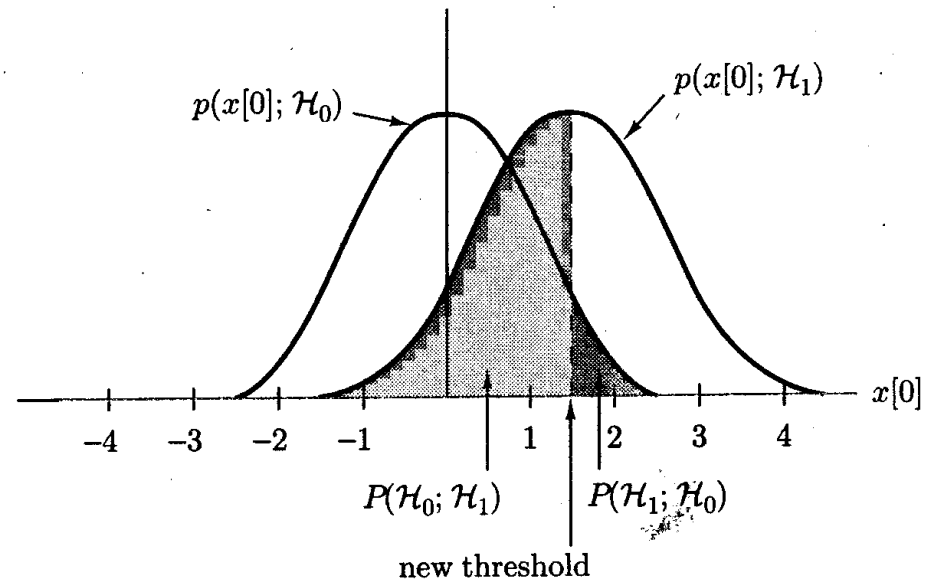
- We wish to maximize the probability of detection

$$P_D = P(H_1; H_1)$$

- And constraint the probability of false alarm

$$P_{FA} = P(H_1; H_0) = \alpha$$

- By choosing the threshold γ



Neyman-Pearson Theorem cont...



- To maximize the probability of detection for a given probability of false alarm decide for H_1 if

$$L(x) = \frac{p(x; H_1)}{p(x; H_0)} > \gamma$$

Likelihood ratio test

- with

$$P_{FA} = \int_{\{x:L(x)>\gamma\}} p(x; H_0) dx = \alpha$$



Likelihood Ratio

- The LR test rejects the null hypothesis if the value of this statistic is too small.
- Intuition: large Pfa \rightarrow small gamma \rightarrow not much larger $p(x, H_1)$ than $p(x; H_0)$
- Limiting case: Pfa \rightarrow 1
- Limiting case: Pfa \rightarrow 0

$$L(x) = \frac{p(x; H_1)}{p(x; H_0)} > \gamma$$



Neyman-Pearson Theorem cont...

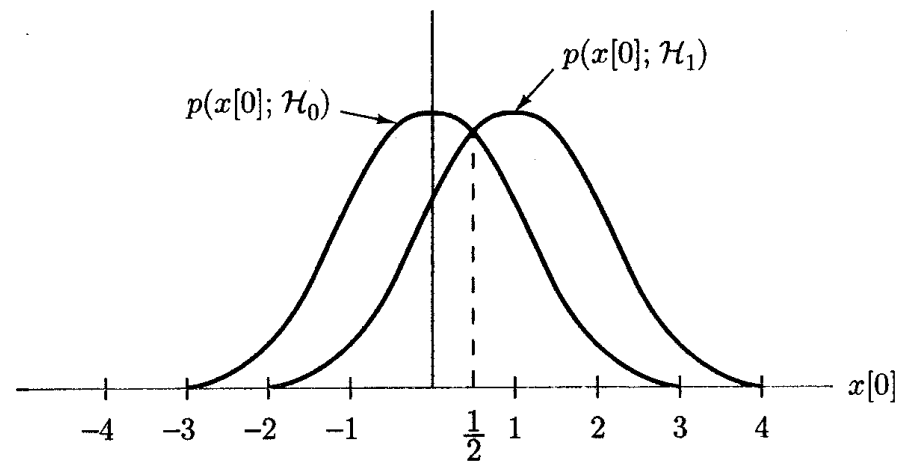
- Example 1
- Maximize

$$P_D = P(H_1; H_1)$$

- Calculate the threshold so that

$$P_{FA} = 10^{-3}$$

$$\begin{aligned} \mathcal{H}_0 &: \mu = 0 \\ \mathcal{H}_1 &: \mu = 1 \end{aligned}$$



Neyman-Pearson Theorem Example 1



$$\frac{p(\mathbf{x}; \mathcal{H}_1)}{p(\mathbf{x}; \mathcal{H}_0)} = \frac{\frac{1}{\sqrt{2\pi}} \exp\left[-\frac{1}{2}(x[0] - 1)^2\right]}{\frac{1}{\sqrt{2\pi}} \exp\left[-\frac{1}{2}x^2[0]\right]} > \gamma \quad (1)$$

$$\exp\left[-\frac{1}{2}(x^2[0] - 2x[0] + 1 - x^2[0])\right] > \gamma \quad (2)$$

$$\exp\left(x[0] - \frac{1}{2}\right) > \gamma. \quad (3)$$

$$\gamma = \exp(\beta) \quad (4)$$

$$\exp\left(x[0] - \frac{1}{2}\right) > \exp(\beta) \quad (5)$$

$$x[0] > \beta + \frac{1}{2} = \boxed{\gamma'} \quad (6)$$

$$P_{FA} = \Pr\{x[0] > \gamma'; \mathcal{H}_0\} = 10^{-3} \quad (7)$$

$$\int_{\gamma'}^{\infty} \frac{1}{\sqrt{2\pi}} \exp\left(-\frac{1}{2}t^2\right) dt = 10^{-3} \quad (8)$$

$$\gamma' = 3$$

$$\begin{aligned} P_D &= \Pr\{x[0] > 3; \mathcal{H}_1\} \\ &= \int_3^{\infty} \frac{1}{\sqrt{2\pi}} \exp\left[-\frac{1}{2}(t - 1)^2\right] dt = 0.023. \end{aligned}$$

Question



How can P_d be improved?

Answer is: Only if we allow for a larger P_{fa} .

NP gives the optimal detector!

By allowing a P_{fa} of 0.5, the P_d is increased to 0.85

System Model



- In the General case

$$\begin{aligned}\mathcal{H}_0 : x[n] &= w[n] & n = 0, 1, \dots, N - 1 \\ \mathcal{H}_1 : x[n] &= A + w[n] & n = 0, 1, \dots, N - 1\end{aligned}$$

- Where the signal $s[n]=A$ for $A>0$ and $w[n]$ is WGN with variance σ^2 .
- It can be shown that the sum of x is a ***sufficient statistic***
- The summation of $x[n]$ is sufficient.

Detector Performance



- In a more general definition

$$P_D = Q \left(Q^{-1}(P_{FA}) - \sqrt{\frac{NA^2}{\sigma^2}} \right).$$

Energy-to-noise-ratio

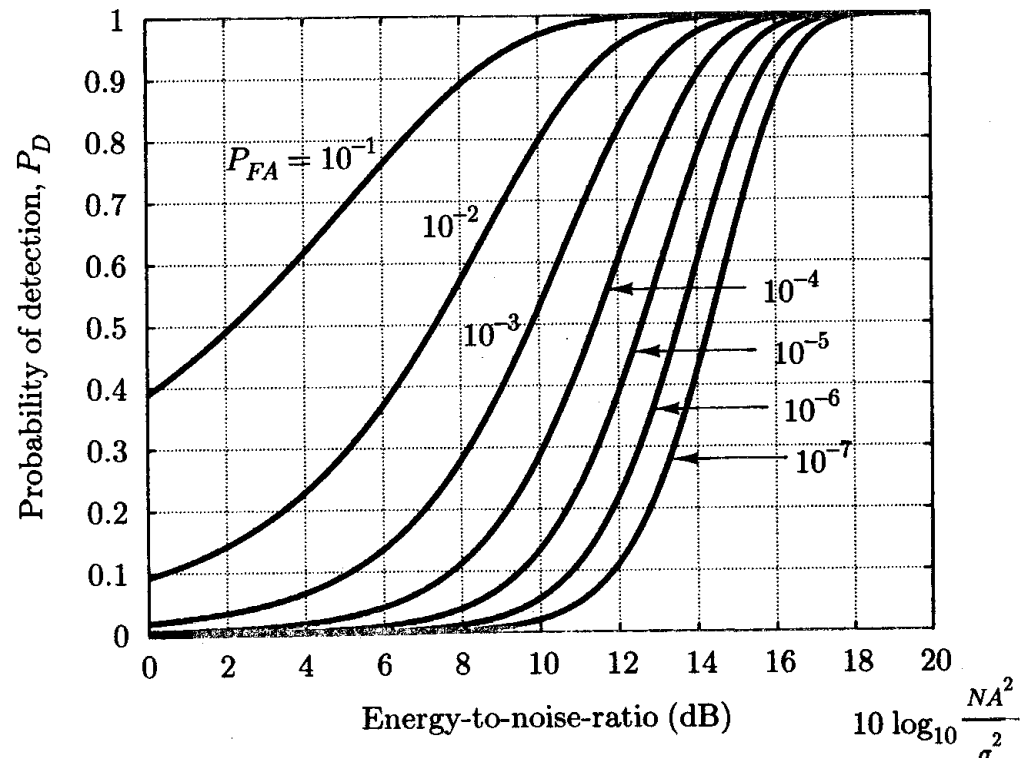
Or

$$P_D = Q \left(Q^{-1}(P_{FA}) - \sqrt{d^2} \right)$$

where

$$d^2 = \frac{(\mu_1 - \mu_0)^2}{\sigma^2}.$$

Deflection Coefficient

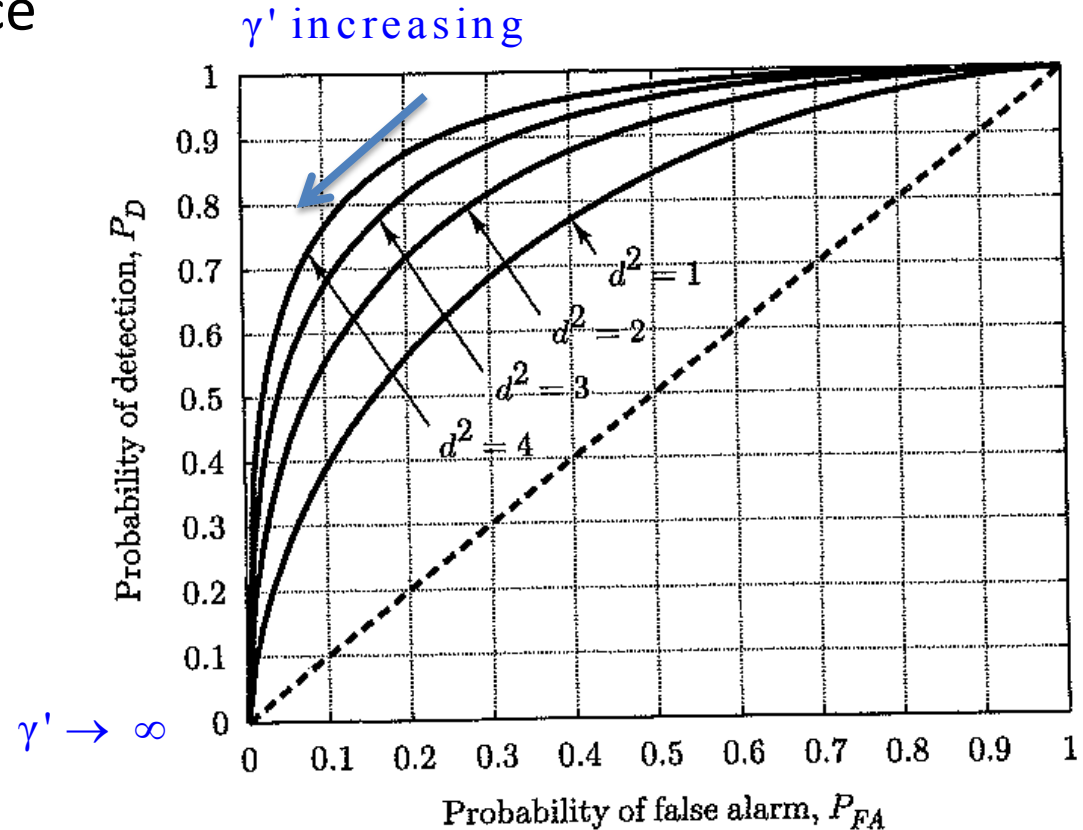


Detector Performance cont...



- An alternative way to present the detection performance

$$P_D = Q\left(Q^{-1}(P_{FA}) - \sqrt{d^2}\right)$$



Receiver Operating Characteristic (ROC)



Irrelevant data

- Some data not belonging to the desired signal might actually be useful.
- The observed data-set is $\{x[0], x[1], \dots, x[N - 1], w_R[0], w_R[1], \dots, w_R[N - 1]\}$
- If

$$H_0 : \begin{array}{l} x[n] = w[n] \\ w_R = w[n] \end{array} \quad H_1 : \begin{array}{l} x[n] = A + w[n] \\ w_R = w[n] \end{array}$$

- Then the reference noise can be used to improve the detection. The perfect detector would become

$$T = x[0] - w_R[0] > \frac{A}{2}$$



Irrelevant data cont...

- Generalizing, using the NP theorem

$$\begin{aligned} L(\mathbf{x}_1, \mathbf{x}_2) &= \frac{p(\mathbf{x}_1, \mathbf{x}_2; \mathcal{H}_1)}{p(\mathbf{x}_1, \mathbf{x}_2; \mathcal{H}_0)} \\ &= \frac{p(\mathbf{x}_2|\mathbf{x}_1; \mathcal{H}_1)p(\mathbf{x}_1; \mathcal{H}_1)}{p(\mathbf{x}_2|\mathbf{x}_1; \mathcal{H}_0)p(\mathbf{x}_1; \mathcal{H}_0)}. \end{aligned}$$

- If they are un-correlated

$$p(\mathbf{x}_2|\mathbf{x}_1; \mathcal{H}_1) = p(\mathbf{x}_2|\mathbf{x}_1; \mathcal{H}_0)$$

- Then, x_2 is irrelevant and

$$L(x_1, x_2) = L(x_1)$$



Minimum Probability Error

- The probability of error is defined with Bayesian paradigm as

$$\begin{aligned} P_e &= \Pr\{\text{decide } \mathcal{H}_0, \mathcal{H}_1 \text{ true}\} + \Pr\{\text{decide } \mathcal{H}_1, \mathcal{H}_0 \text{ true}\} \\ &= P(\mathcal{H}_0|\mathcal{H}_1)P(\mathcal{H}_1) + P(\mathcal{H}_1|\mathcal{H}_0)P(\mathcal{H}_0) \end{aligned}$$

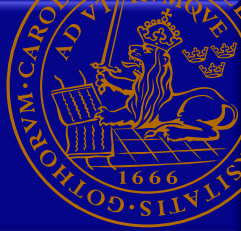
- Minimizing the probability of error, we decide H1 if

$$\frac{p(\mathbf{x}|\mathcal{H}_1)}{p(\mathbf{x}|\mathcal{H}_0)} > \frac{P(\mathcal{H}_0)}{P(\mathcal{H}_1)} = \gamma.$$

- If both prior probabilities are equal we have

$$p(\mathbf{x}|\mathcal{H}_1) > p(\mathbf{x}|\mathcal{H}_0).$$

Minimum Probability Error example



- An On-off key communication system , where we transmit either $S_0[n] = 0$ or $S_1[n] = A$

$$\begin{aligned}\mathcal{H}_0 : x[n] &= w[n] & n = 0, 1, \dots, N - 1 \\ \mathcal{H}_1 : x[n] &= A + w[n] & n = 0, 1, \dots, N - 1\end{aligned}$$

- The probability of transmitting 0 or A is the same $\frac{1}{2}$. Then, in order to minimize the probability of error $\gamma=1$

Minimum Probability Error example ...



$$\frac{\frac{1}{(2\pi\sigma^2)^{\frac{N}{2}}} \exp\left[-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} (x[n] - A)^2\right]}{\frac{1}{(2\pi\sigma^2)^{\frac{N}{2}}} \exp\left[-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} x^2[n]\right]} > 1.$$

$$-\frac{1}{2\sigma^2} \left(-2A \sum_{n=0}^{N-1} x[n] + NA^2 \right) > 0$$

Analogous to NP, we decide H_1 if $\bar{x} > A/2$

Then we calculate the P_e

$$\bar{x} \sim \begin{cases} \mathcal{N}(0, \frac{\sigma^2}{N}) & \text{conditioned on } \mathcal{H}_0 \\ \mathcal{N}(A, \frac{\sigma^2}{N}) & \text{conditioned on } \mathcal{H}_1. \end{cases}$$

$$\begin{aligned} P_e &= \frac{1}{2} [P(\mathcal{H}_0|\mathcal{H}_1) + P(\mathcal{H}_1|\mathcal{H}_0)] \\ &= \frac{1}{2} [\Pr\{\bar{x} < A/2|\mathcal{H}_1\} + \Pr\{\bar{x} > A/2|\mathcal{H}_0\}] \\ &= \frac{1}{2} \left[\left(1 - Q\left(\frac{A/2 - A}{\sqrt{\sigma^2/N}}\right) \right) + Q\left(\frac{A/2}{\sqrt{\sigma^2/N}}\right) \right] \end{aligned}$$

$$P_e = Q\left(\sqrt{\frac{NA^2}{4\sigma^2}}\right).$$

Deflection Coefficient

The probability error decreases monotonically with respect of the deflection coefficient

Bayes Risk



- Bayes Risk assigns costs to the type of errors, C_{ij} denotes the cost if we decide H_i but H_j is true. If no error is made, no cost is assigned.

$$\mathcal{R} = \sum_{i=0}^{M-1} \sum_{j=0}^{M-1} C_{ij} P(\mathcal{H}_i | \mathcal{H}_j) P(\mathcal{H}_j).$$

- The detector that minimizes Bayes risk is the following

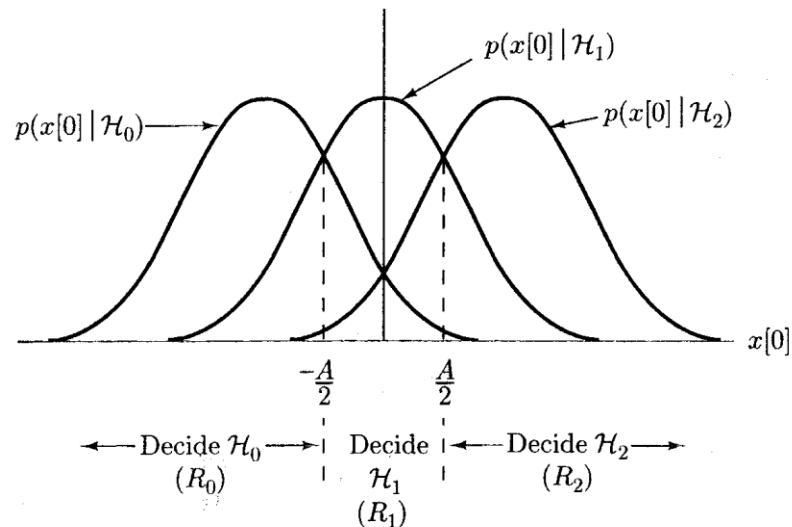
$$\frac{p(\mathbf{x} | \mathcal{H}_1)}{p(\mathbf{x} | \mathcal{H}_0)} > \frac{(C_{10} - C_{00})P(\mathcal{H}_0)}{(C_{01} - C_{11})P(\mathcal{H}_1)} = \gamma.$$

Multiple Hypotheses testing



- If we have a communication system, M-ary PAM (pulse amplitude modulation) with $M=3$.

$$\begin{aligned}\mathcal{H}_0 : x[n] &= -A + w[n] & n = 0, 1, \dots, N-1 \\ \mathcal{H}_1 : x[n] &= w[n] & n = 0, 1, \dots, N-1 \\ \mathcal{H}_2 : x[n] &= A + w[n] & n = 0, 1, \dots, N-1\end{aligned}$$



Multiple Hypotheses testing cont...



- The conditional PDF describing the communication system is given by

$$p(\mathbf{x}|\mathcal{H}_i) = \frac{1}{(2\pi\sigma^2)^{\frac{N}{2}}} \exp \left[-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} (x[n] - A_i)^2 \right]$$

- In order to maximize $p(x/H_i)$, we can minimize

$$D_i^2 = \sum_{n=0}^{N-1} (x[n] - A_i)^2.$$

Multiple Hypotheses testing cont...



- Then

$$\begin{aligned} D_i^2 &= \sum_{n=0}^{N-1} (x[n] - \bar{x} + \bar{x} - A_i)^2 \\ &= \sum_{n=0}^{N-1} (x[n] - \bar{x})^2 + 2(\bar{x} - A_i) \sum_{n=0}^{N-1} (x[n] - \bar{x}) + N(\bar{x} - A_i)^2 \\ &= \sum_{n=0}^{N-1} (x[n] - \bar{x})^2 + N(\bar{x} - A_i)^2. \end{aligned}$$

- Therefore

$$\begin{aligned} \mathcal{H}_0 &\text{ if } \bar{x} < -A/2 \\ \mathcal{H}_1 &\text{ if } -A/2 < \bar{x} < A/2 \\ \mathcal{H}_2 &\text{ if } \bar{x} > A/2. \end{aligned}$$

Multiple Hypotheses testing cont...



- There are six types of \mathcal{P}_e , so instead of calculating \mathcal{P}_e , we calculate $\mathcal{P}_c=1-\mathcal{P}_e$, the probability of correct decision

$$\begin{aligned} P_c &= \sum_{i=0}^2 P(\mathcal{H}_i|\mathcal{H}_i)P(\mathcal{H}_i) \\ &= \frac{1}{3} \sum_{i=0}^2 P(\mathcal{H}_i|\mathcal{H}_i) \\ &= \frac{1}{3} [\Pr\{\bar{x} < -A/2|\mathcal{H}_0\} + \Pr\{-A/2 < \bar{x} < A/2|\mathcal{H}_1\} + \Pr\{\bar{x} > A/2|\mathcal{H}_2\}]. \end{aligned}$$

- And finally

$$\begin{aligned} P_c &= \frac{1}{3} \left[1 - Q\left(\frac{-\frac{A}{2} + A}{\sqrt{\sigma^2/N}}\right) + Q\left(\frac{-\frac{A}{2}}{\sqrt{\sigma^2/N}}\right) - Q\left(\frac{\frac{A}{2}}{\sqrt{\sigma^2/N}}\right) + Q\left(\frac{\frac{A}{2} - A}{\sqrt{\sigma^2/N}}\right) \right] \\ &= 1 - \frac{4}{3} Q\left(\sqrt{\frac{NA^2}{4\sigma^2}}\right) \end{aligned}$$

Multiple Hypotheses testing cont...



Binary Case

$$P_e = Q \left(\sqrt{\frac{NA^2}{4\sigma^2}} \right)$$

PAM Case

$$P_e = \frac{4}{3} Q \left(\sqrt{\frac{NA^2}{4\sigma^2}} \right)$$

In the digital communications world, there is a difference in the energy per symbol, which is not considered in this example.

Chapter Problems



- Warm up problems: 3.1, 3.2
- Problems to be solved in class: 3.4, 3.8, 3.15, 3.18, 3.21