



Lecture no: **4**

## Channel models and antennas

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# WHY DO WE NEED CHANNEL MODELS?

# Why do we need channel models?



During system design, testing and type approval:

Simple models reflecting the important properties of important channels (best, average, worst case)

Models used to make sure that the system design behaves well in typical situations.

During network design:

More detailed models appropriate for certain geographical areas

Models used to obtain an efficient network in terms of base station locations and other parameters



# NARROWBAND MODELS

# Narrowband models

## Review of properties



Narrowband models contain "only one" attenuation, which is modeled as a propagation loss, plus large- and small-scale fading.

Path loss: Often proportional to  $1/d^n$ , where  $n$  is the propagation exponent. ( $n$  may be different at different distances)

Large-scale fading: Log-normal distribution (normal distr. in dB scale)

Small-scale fading: Rayleigh, Rice, Nakagami distributions ... (**not** in dB-scale)

NOTE: Several of these models are found in an on-line appendix of the textbook which can be downloaded from the publisher's website (see "Literature" on course web).

Printed copies of textbook appendices are allowed during Part B of the written exam.

# Okumura's measurements Background



Extensive measurement campaign in Japan in the 1960's.

Parameters varied during measurements:

Frequency	100 - 3000 MHz
Distance	1 - 100 km
Mobile station height	1 - 10 m
Base station height	20 - 1000 m
Environment	medium-size city, large city, etc.

Propagation loss is given as a **median** value (50% of the time and 50% of the area).

Results from these measurements are displayed in figures 7.12 - 7.14.

# Okumura's measurements

## How to calculate the prop. loss



1. We start by calculating the free-space attenuation
2. Apply a frequency and distance dependent correction
3. Apply a BS-height and distance dependent correction
4. Apply a MS-height, frequency and environment dependent correction

Free space  
attenuation



$L_{Oku}$

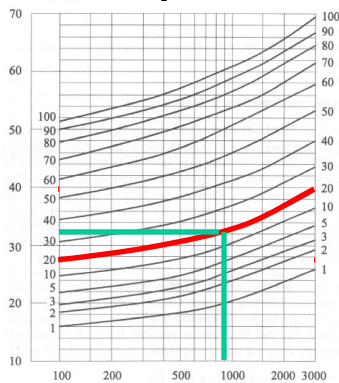


Fig. 7.12

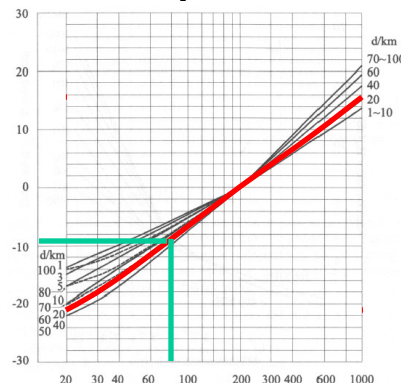


Fig. 7.13

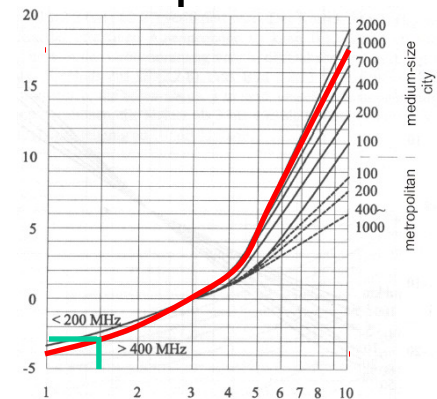


Fig. 7.14



# Okumura's measurements

## Example



## Example

Propagation at 900 MHz in medium-size city with 40 m base station antenna height and 1.5 m mobile station antenna height.

Use Okumura's curves to calculate the propagation loss at a distance of 30 km between base station and mobile station.

# Okumura's measurements

## 1. Calculate free-space loss



## Example

Attenuation between two isotropic antennas in free space is (free-space loss):

$$L_{\text{free|dB}}(d) = 20 \log \left( \frac{4 \pi d}{\lambda} \right)$$

The obtained value does not depend on antenna heights.

900 MHz and  
30 km distance

=> **121 dB**

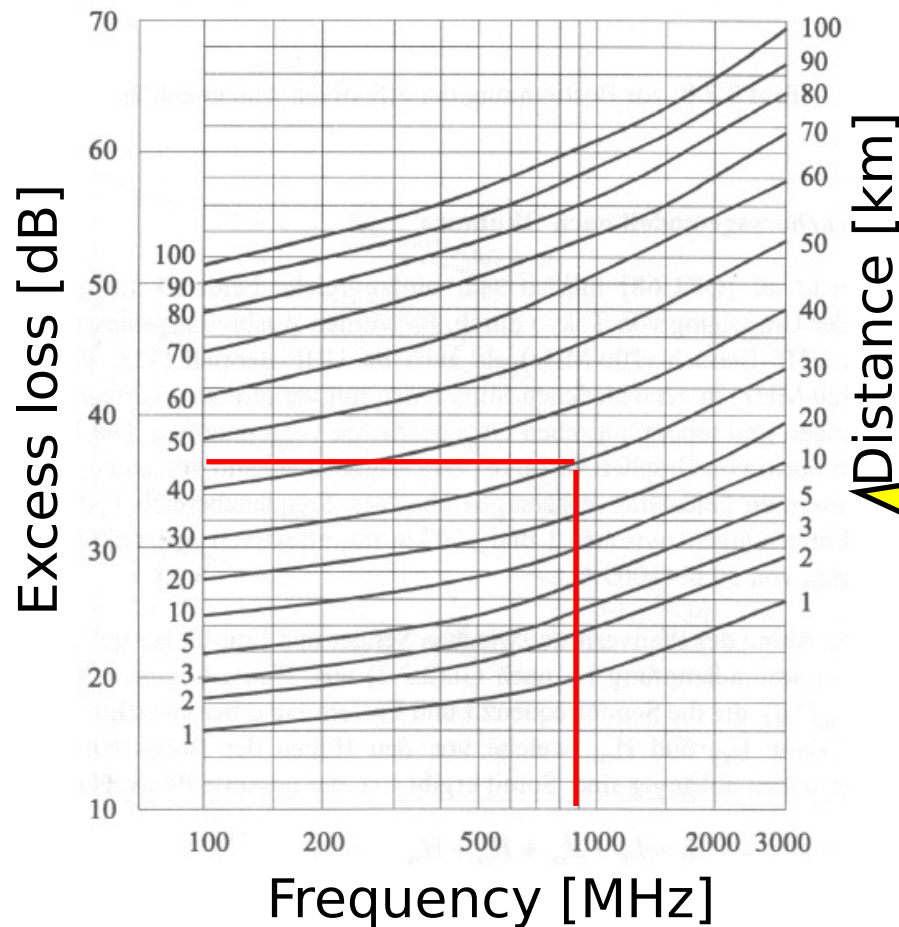
# Okumura's measurements

## 2. Apply correction for excess loss



# Example

**FIGURE 7.12**



These curves are only for  $h_b=200$  m and  $h_m=3$  m

900 MHz and  
30 km distance

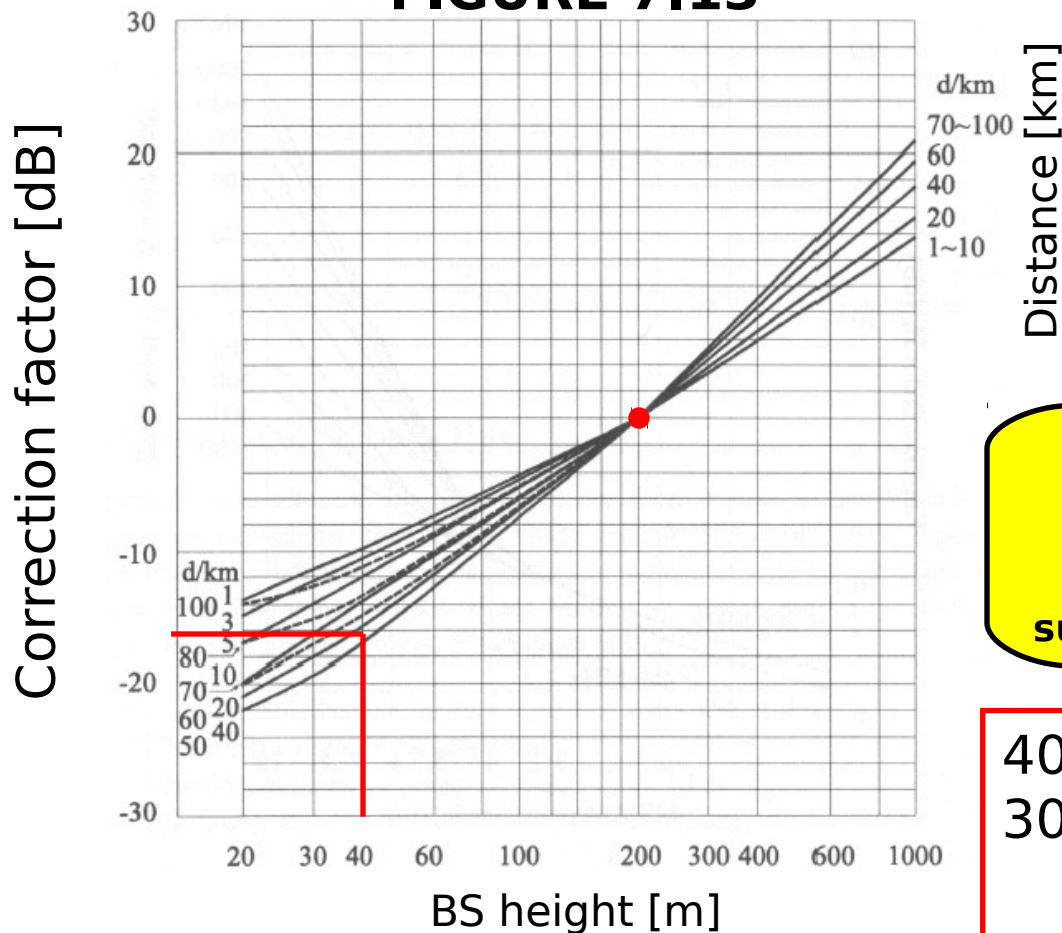
=> **36.5 dB**

# Okumura's measurements

## 3. Apply correction of BS height



FIGURE 7.13



# Example

**Note:** Lower base station means INCREASING attenuation => **subtract** this number.

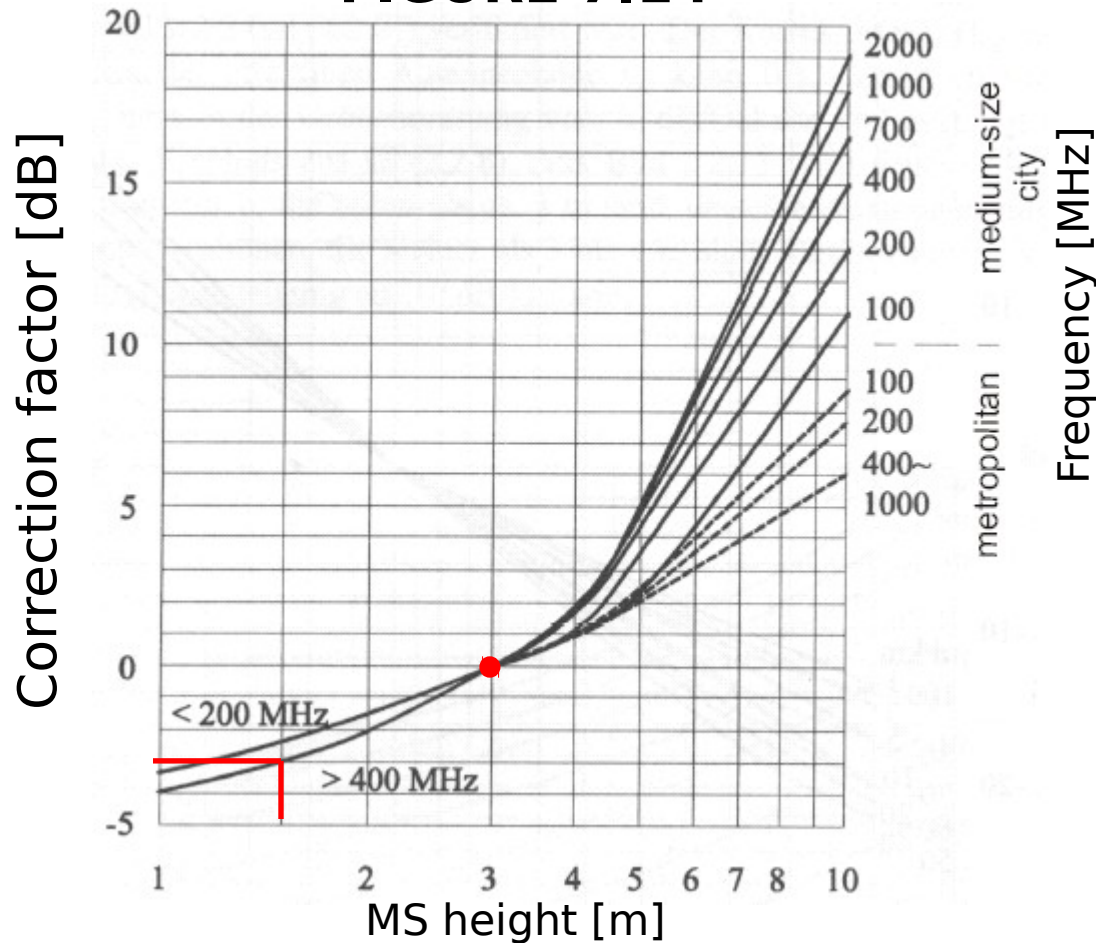
40 m BS and  
30 km distance  
=> **-16 dB**

# Okumura's measurements

## 4. Apply correction of MS height



FIGURE 7.14



# Example

**Note:** Lower mobile station means INCREASING attenuation => **subtract** this number.

1.5 m MS and 900 MHz in medium-size city => **-3 dB**

# Okumura's measurements

## Summary of example



# Example

Propagation loss (between isotropic antennas) using Okumura's measurements:

$$L_{Oku|dB} = 121 + 36.5 - (-16) - (-3) = 176.5 \text{ dB}$$

Calc. step:    1            2            3            4

# The Okumura-Hata model Background



In 1980 Hata published a parameterized model, based on Okumura's measurements.

The parameterized model has a *smaller range of validity* than the measurements by Okumura:

Frequency	150 – 1500 MHz
Distance	1 – 20 km
Mobile station height	1 – 10 m
Base station height	30 – 200 m

# The Okumura-Hata model

## How to calculate prop. loss



$$L_{O-H} = A + B \log(d_{|km}) + C$$

$$A = 69.55 + 26.16 \log(f_{0|MHz}) - 13.82 \log(h_b) - a(h_m)$$

$$B = 44.9 - 6.55 \log(h_b)$$

$h_b$  and  $h_m$   
in **meter**

	$a(h_m) =$	$C =$
Metropolitan areas	$8.29 (\log(1.54 h_m))^2 - 1.1$ for $f_0 \leq 200$ MHz $3.2 (\log(11.75 h_m))^2 - 4.97$ for $f_0 \geq 400$ MHz	0
Small/medium-size cities	$(1.1 \log(f_{0 MHz}) - 0.7) h_m -$ $(1.56 \log(f_{0 MHz}) - 0.8)$	0
Suburban environments		$-2 (\log(f_{0 MHz}/28))^2 - 5.4$
Rural areas		$-4.78 (\log(f_{0 MHz}))^2 + 18.33 \log(f_{0 MHz}) - 40.94$



# COST 231-Walfish-Ikegami model Background



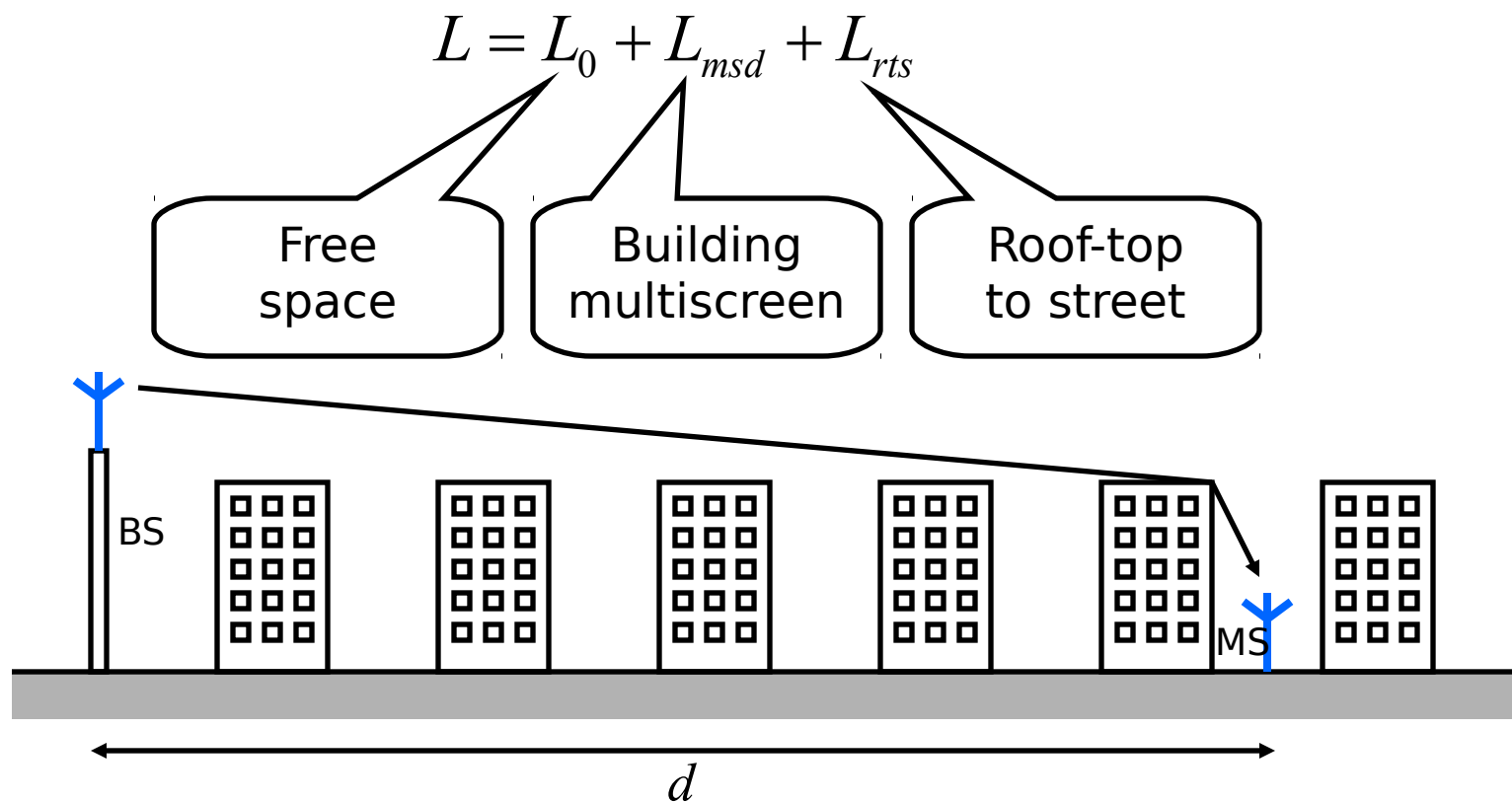
The Okumura-Hata model is not suitable for micro cells or small macro cells, due to its restrictions on distance ( $d > 1$  km).

The COST 231-Walfish-Ikegami model covers much smaller distances and is better suited for calculations on small cells.

Frequency	800 – 2000 MHz
Distance	0.02 – 5 km
Mobile station height	1 – 3 m
Base station height	4 – 50 m

# COST 231-Walfish-Ikegami model

## How to calculate prop. loss



Details about calculations can be found in Appendix 7.B.



# WIDEBAND MODELS

# Wideband models

## Review of properties



Let's assume the tapped delay-line model

$$h(t, \tau) = \sum_{i=1}^N \alpha_i(t) \exp(j \theta_i(t)) \delta(\tau - \tau_i)$$

The **power-delay profile** tells us how much energy the channel has at a certain delay  $\tau$  (essentially the rms values of the  $\alpha_i(t)$ 's).

The **Doppler spectrum** tells us how fast the channel changes in time (essentially how fast the  $\alpha_i(t)$ 's and  $\theta_i(t)$ 's change). There can be one Doppler spectrum for each delay.

# Wideband models

## COST 207 model for GSM



The COST 207 model specifies:

FOUR power-delay profiles for different environments.

FOUR Doppler spectra used for different delays.

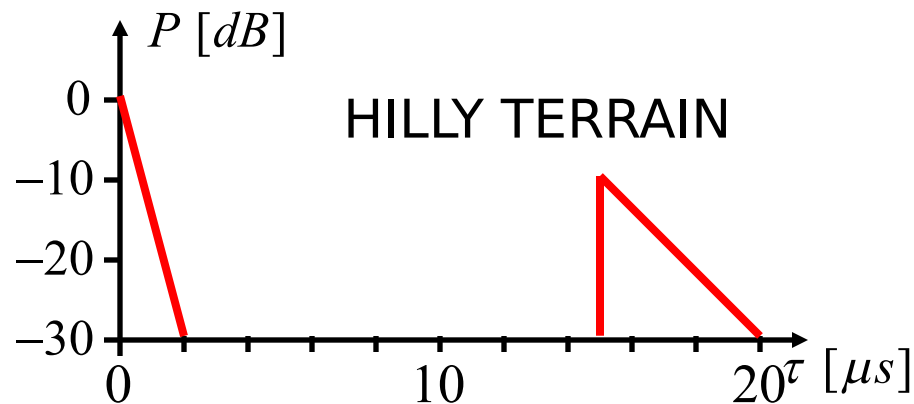
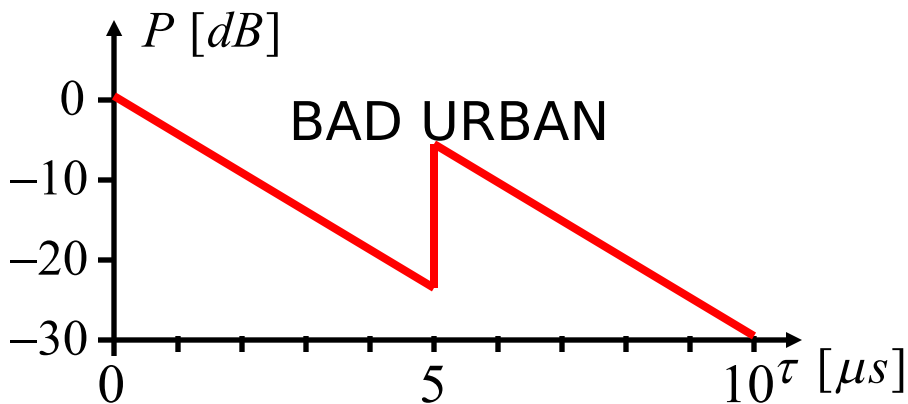
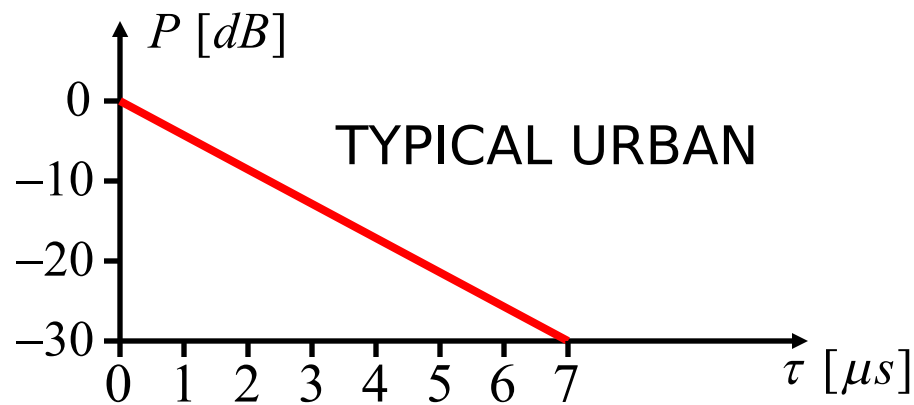
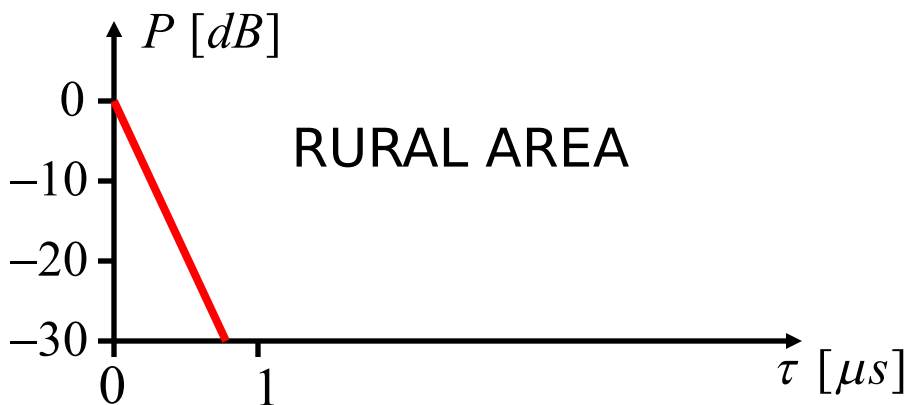
**IT DOES NOT SPECIFY PROAGATION LOSSES FOR THE DIFFERENT ENVIRONMENTS!**

# Wideband models

## COST 207 model for GSM



### Four specified power-delay profiles



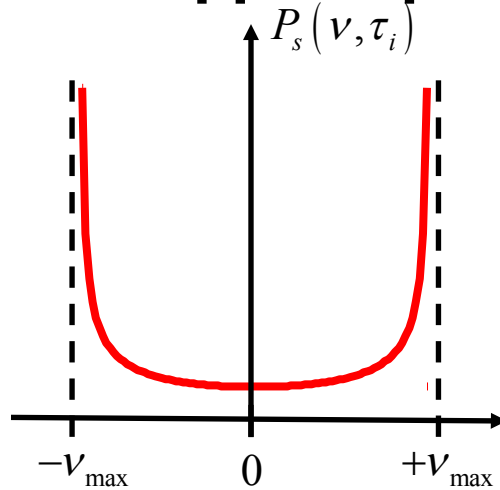
# Wideband models

## COST 207 model for GSM

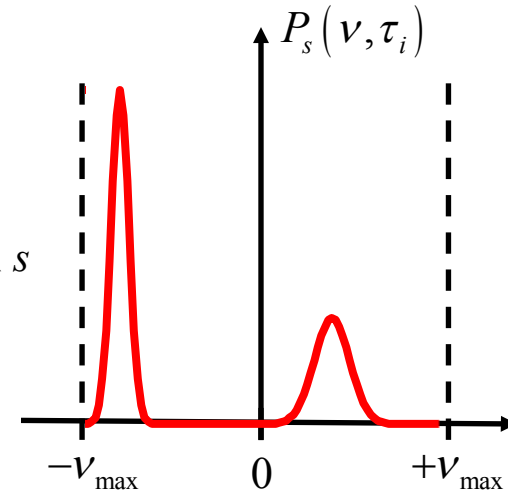


### Four specified Doppler spectra

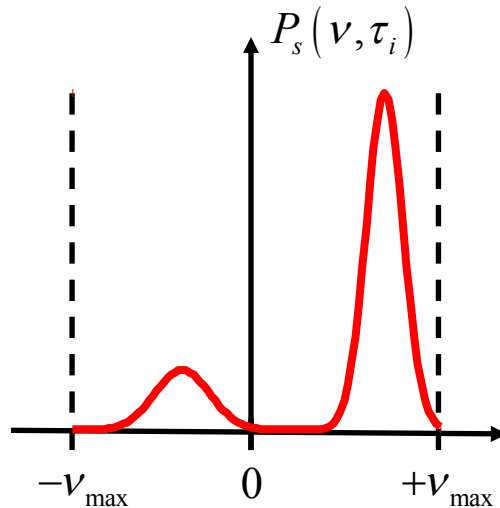
CLASS  
 $\tau_i \leq 0.5 \mu s$



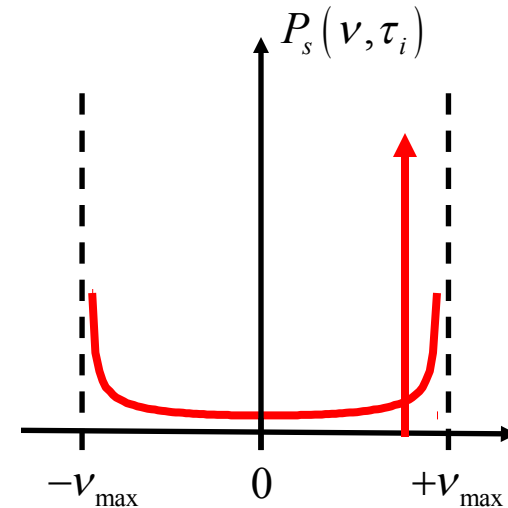
GAUS1  
 $0.5 \mu s < \tau_i \leq 2 \mu s$



GAUS2  
 $\tau_i > 2 \mu s$



RICE  
Shortest  
path in  
rural areas

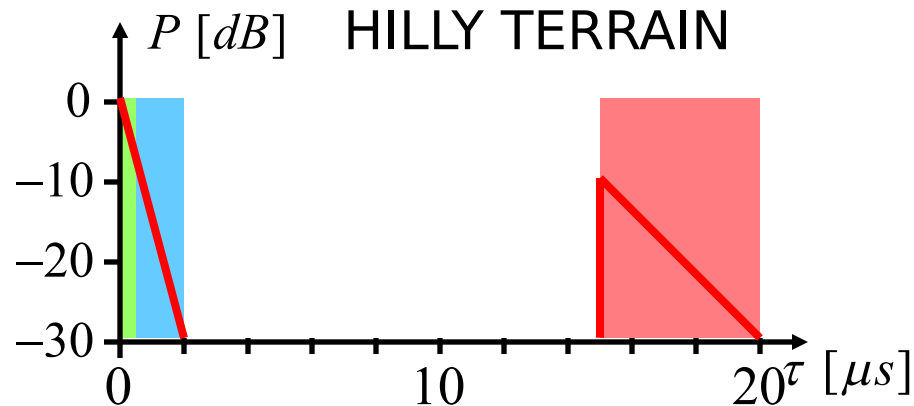
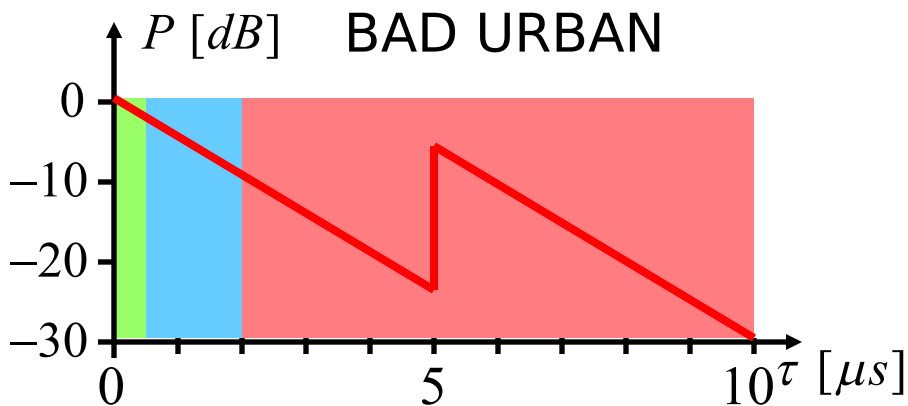
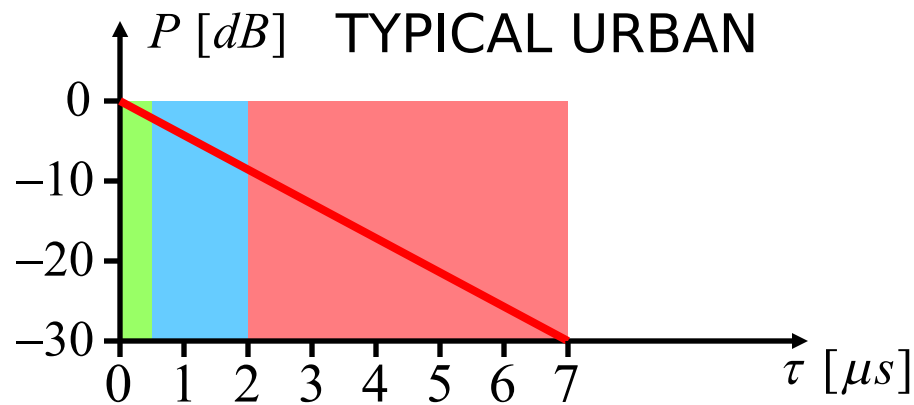
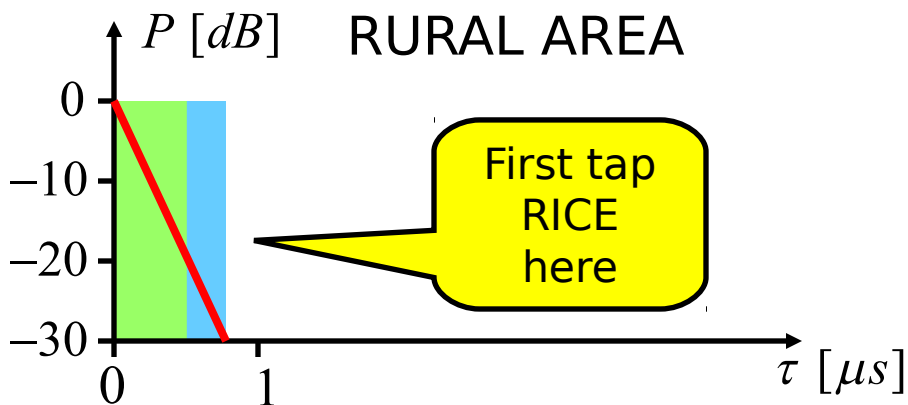


# Wideband models

## COST 207 model for GSM



Doppler spectra: CLASS GAUS1 GAUS2





# Wideband models

## COST 207 model for GSM



There are also suggested tapped delay-line implementations, with six Rayleigh-fading taps per channel. See Appendix 7.C (on-line).

**QUICK QUIZ:** The system bit-rate of GSM is 271 kbit/s.

How long is one bit in time?

How long are the different COST 207 channels, measured in bit-times?

# Wideband models

## ITU-R model for 3G



The ITU-R model specifies:

SIX different tapped delay-line channels for three different scenarios (indoor, pedestrian, vehicular).

TWO channels per scenario (one short and one long delay spread).

TWO different Doppler spectra (uniform & classical), depending on scenario.

THREE different models for propagation loss (one for each scenario).

The standard deviation of the log-normal shadow fading is specified for each scenario.

The autocorrelation of the log-normal shadow fading is specified for the vehicular scenario.

# Wideband models

## ITU-R model for 3G



Tap No.	delay/ns	power/dB	delay/ns	power/dB
<b>INDOOR</b>	<b>CHANNEL A (50%)</b>		<b>CHANNEL B (45%)</b>	
1	0	0	0	0
2	50	-3	100	-3.6
3	110	-10	200	-7.2
4	170	-18	300	-10.8
5	290	-26	500	-18.0
6	310	-32	700	-25.2
<b>PEDESTRIAN</b>	<b>CHANNEL A (40%)</b>		<b>CHANNEL B (55%)</b>	
1	0	0	0	0
2	110	-9.7	200	-0.9
3	190	-19.2	800	-4.9
4	410	-22.8	1200	-8.0
5			2300	-7.8
6			3700	-23.9
<b>VEHICULAR</b>	<b>CHANNEL A (40%)</b>		<b>CHANNEL B (55%)</b>	
1	0	0	0	-2.5
2	310	-1	300	0
3	710	-9	8900	-12.8
4	1090	-10	12900	-10.0
5	1730	-15	17100	-25.2
6	2510	-20	20000	-16.0



# ANTENNAS

# Antennas Efficiency



The antenna efficiency measures “how efficiently” an antenna converts the input power into radiation. This translates directly into power consumption and battery life.

Antenna efficiency of mobiles has **decreased** mainly due to cosmetic restrictions.

What cosmetic restrictions?

# Antennas Bandwidth



We can say that the bandwidth of an antenna is the width of the frequency range over which it fulfills some specification.

Most cellular systems have a bandwidth requirement in the range of 10% of the carrier frequency.

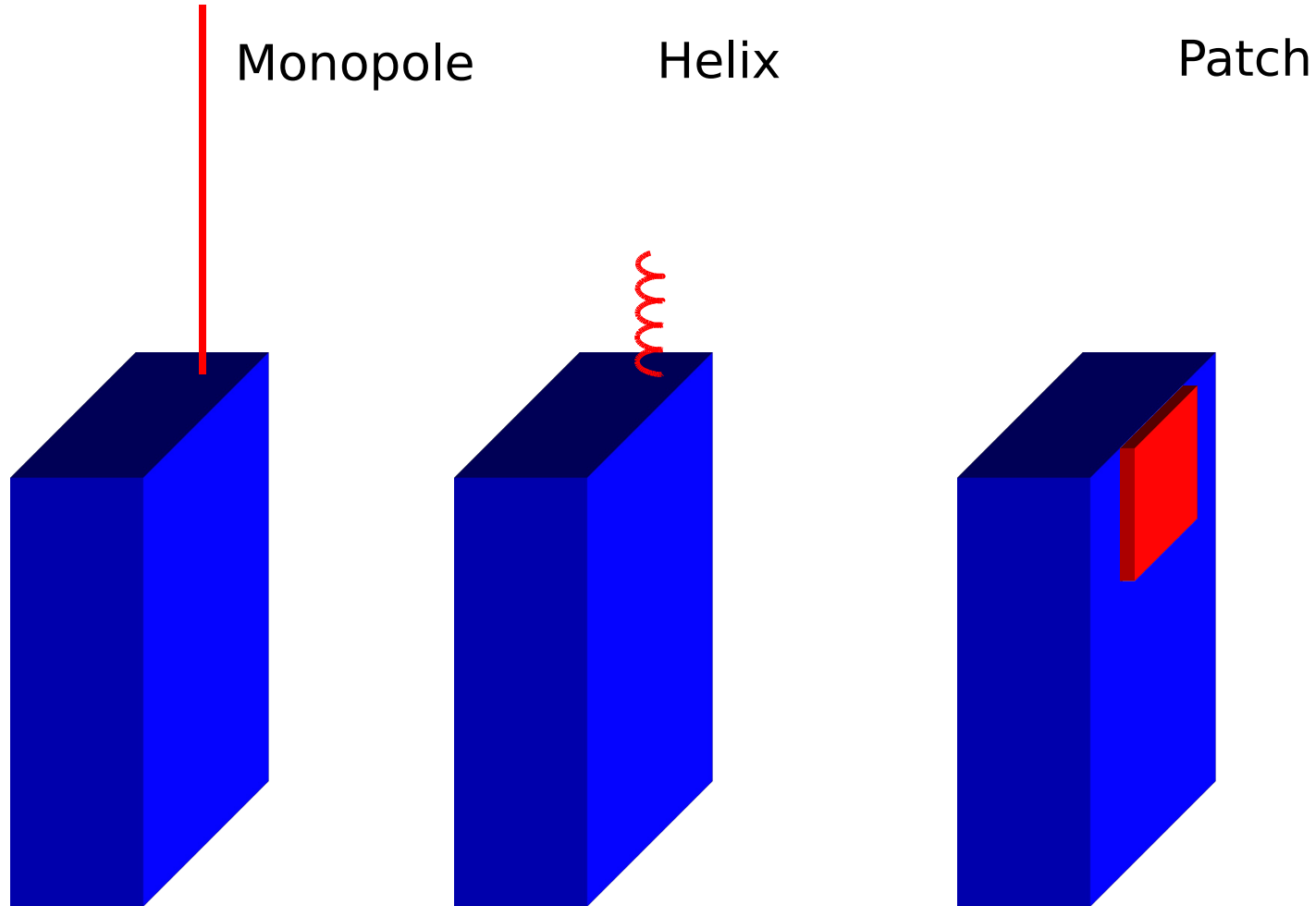
**Example:** 900 MHz GSM needs an antenna that can transmit/receive well in a total bandwidth of about 100 MHz.

It is difficult to make small and efficient broadband antennas!

What happens when we have dual- (900/1800) or triple-band (900/1800/1900) GSM phones ... or phones with 3G and Bluetooth (2.4 GHz) as well?

# Antennas

## Mobile station antennas

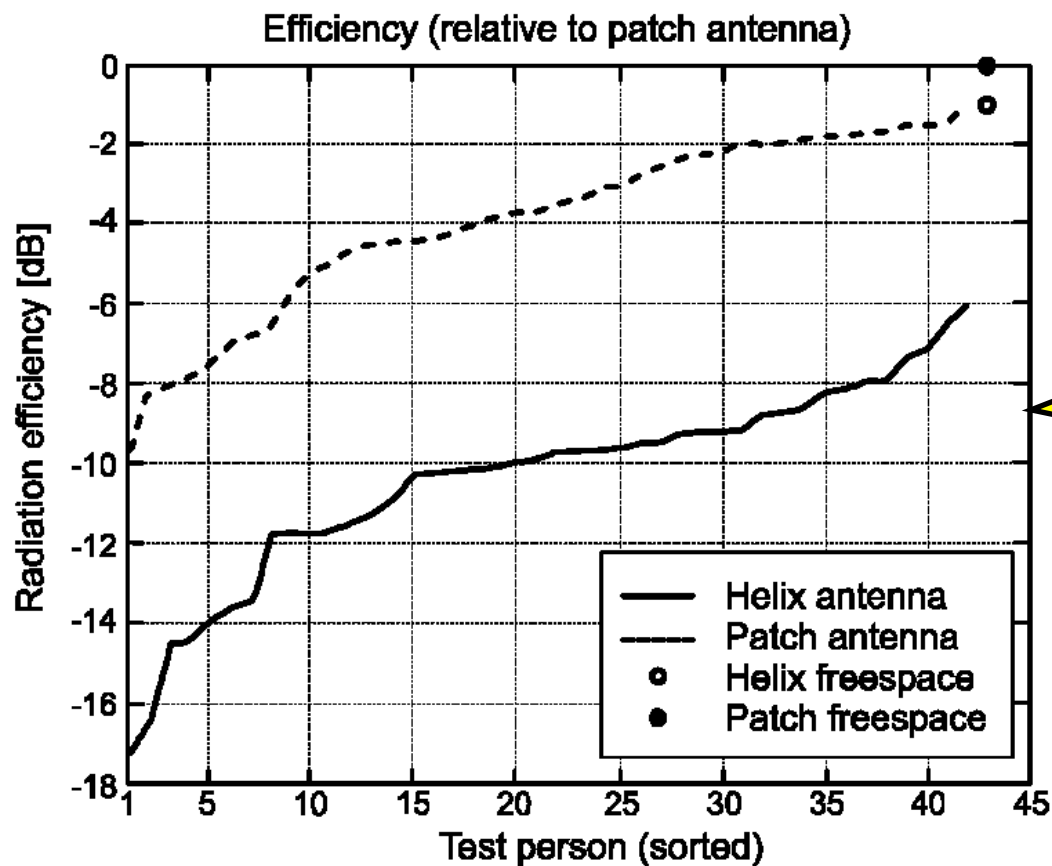


# Antennas

## Mobile station antennas



The efficiency depends on many parameters, but a very important one is its environment. Below you can see differences in antenna efficiency for 42 test persons holding the mobile.



Up to around 10 dB difference, depending on person.

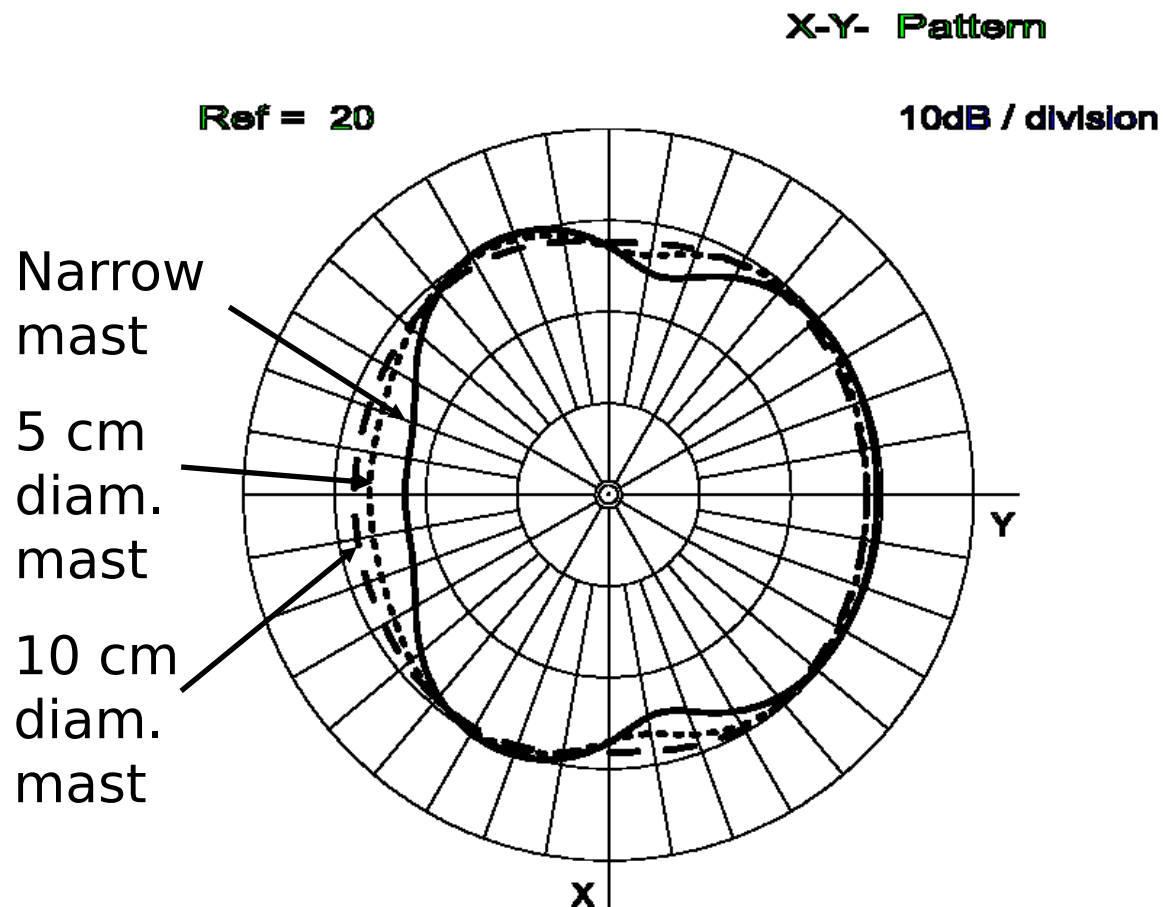


# Antennas

## Base station antennas



Base station antenna pattern affected by the mast (30 cm from antenna).

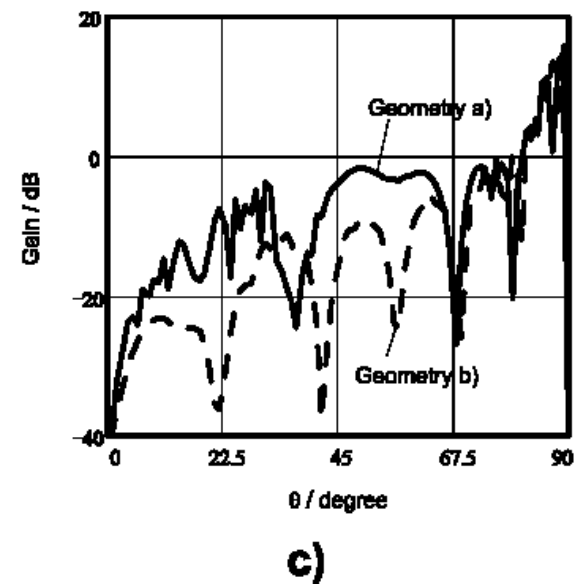
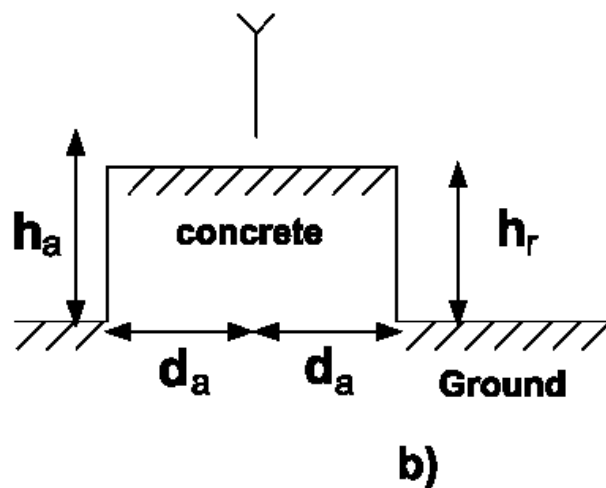
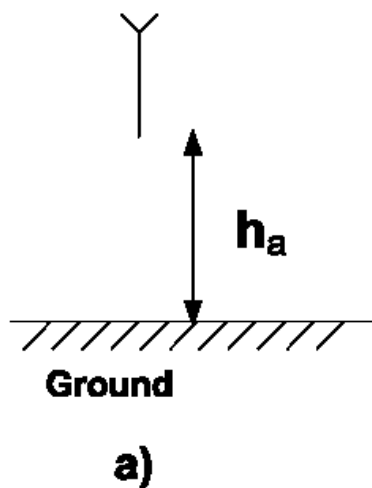


# Antennas

## Base station antennas



Base station antenna pattern affected by a concrete foundation.



# Antennas

## The dipole antenna



Short dipole  
( $L \ll \lambda/2$ )

$$R_r \approx 20\pi^2 2 \left(\frac{L}{\lambda}\right)^2$$

$$G_a = 1,5 = 1,8 \text{ dB}$$

$$L_{\text{eff}} = \frac{L}{2}$$

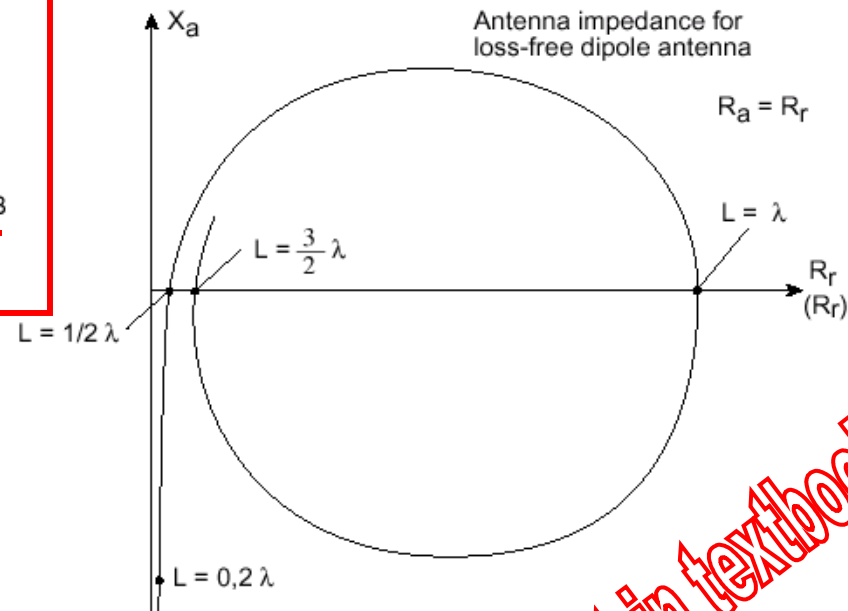
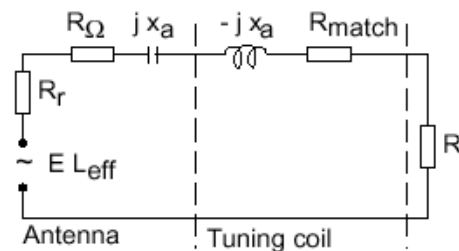
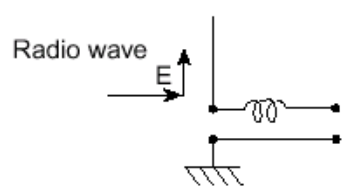
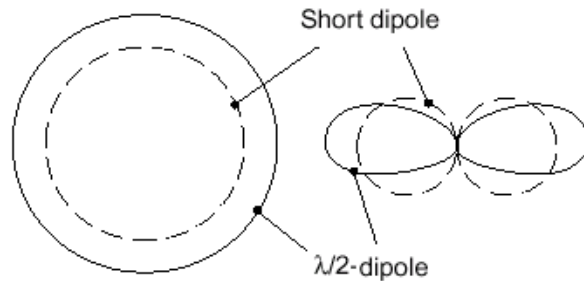
( $R_r$  = Radiation resistance)

$\lambda/2$ -dipole  
( $L = \lambda/2$ )

$$R_r = 73 \Omega$$

$$G_a = 1,64 = 2,15 \text{ dB}$$

$$L_{\text{eff}} = \frac{2}{\pi} L$$



**Not in textbook!**

For a short dipole ( $L/\lambda \ll 1/2$ )  
 $R_r$  will be very low and  $\frac{|X_a|}{R_r}$   
 very high. Difficult to avoid  
 ohmic losses ( $R_{\Omega}$ )  
 and losses in the tuning coil ( $R_{\text{match}}$ )

$L_{\text{eff}}$ : Effective length of antenna

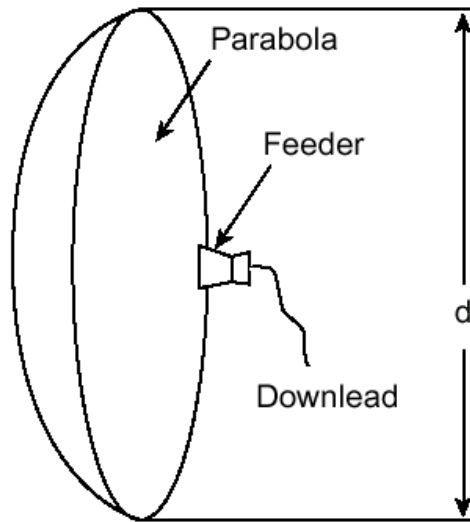
Matching condition:

$$R_l = R_r + R_{\Omega} + R_{\text{match}}$$

[Figure from Ericsson Radio School documentation]

# Antennas

## The parabolic antenna

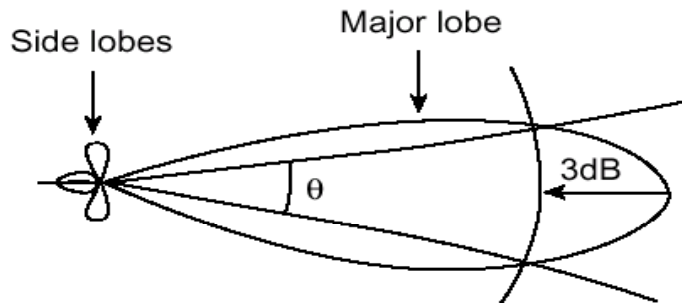


Opening area:  $A = \frac{\pi d^2}{4}$

Effective area:  $A_{\text{eff}} \approx 0.55 A$

Antenna gain:  $G_a = \frac{4\pi}{\lambda^2} A_{\text{eff}} \approx 0.55 \frac{\pi^2 d^2}{\lambda^2}$

**Not in textbook!**



3dB beamwidth:  $\theta \approx \frac{200}{\sqrt{G_a}}$  [degrees] ( $\theta < 25^\circ$ )

[Figure from Ericsson Radio School documentation]



# Summary

- Narrowband models: **Okumura's measurements, Okumura-Hata, COST 231-Ikegami-Walfish**. Mainly models for **propagation** loss. Fading has to be added.
- Wideband models: **COST 207 for GSM & ITU-R for 3G**. Mainly specification of **power-delay profile** and **doppler spectrum** (ITU-R also gives e.g. path loss).
- Antennas: **Efficiency** has decreased for mobile antennas. Antenna **environment** changes their properties. Some specific properties for **dipole** and **parabolic** antennas.