



Lecture no: 4

Channel models and antennas

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WHY DO WE NEED CHANNEL MODELS?

Why do we need channel models?



During system design, testing and type approval:

- Simple models reflecting the important properties of important channels (best, average, worst case)

- Models used to make sure that the system design behaves well in typical situations.

During network design:

- More detailed models appropriate for certain geographical areas

- Models used to obtain an efficient network in terms of base station locations and other parameters



NARROWBAND MODELS

Narrowband models

Review of properties



Narrowband models contain "only one" attenuation, which is modeled as a propagation loss, plus large- and small-scale fading.

Path loss: Often proportional to $1/d^n$, where n is the propagation exponent. (n may be different at different distances)

Large-scale fading: Log-normal distribution (normal distr. in dB scale)

Small-scale fading: Rayleigh, Rice, Nakagami distributions ... (**not** in dB-scale)

NOTE: Several of these models are found in an on-line appendix of the textbook which can be downloaded from the publisher's website (see "Literature" on course web).

Printed copies of textbook appendices are allowed during Part B of the written exam.

Okumura's measurements Background



Extensive measurement campaign in Japan in the 1960's.

Parameters varied during measurements:

Frequency	100 – 3000 MHz
Distance	1 – 100 km
Mobile station height	1 – 10 m
Base station height	20 – 1000 m
Environment	medium-size city, large city, etc.

Propagation loss is given as a **median** value (50% of the time and 50% of the area).

Results from these measurements are displayed in figures 7.12 – 7.14.

Okumura's measurements

How to calculate the prop. loss



1. We start by calculating the free-space attenuation
2. Apply a frequency and distance dependent correction
3. Apply a BS-height and distance dependent correction
4. Apply a MS-height, frequency and environment dependent correction

Free space
attenuation



L_{Oku}

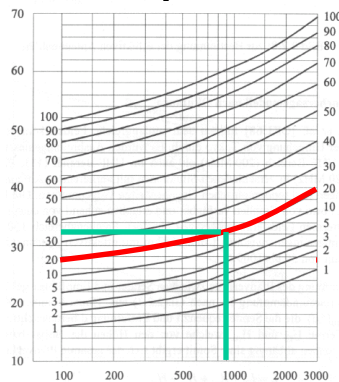


Fig. 7.12

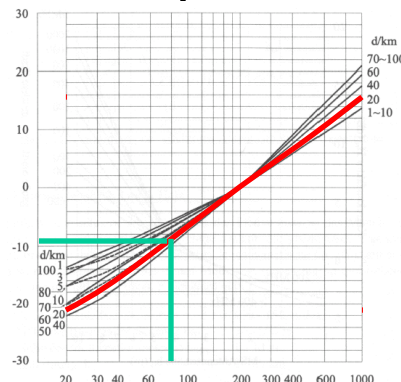


Fig. 7.13

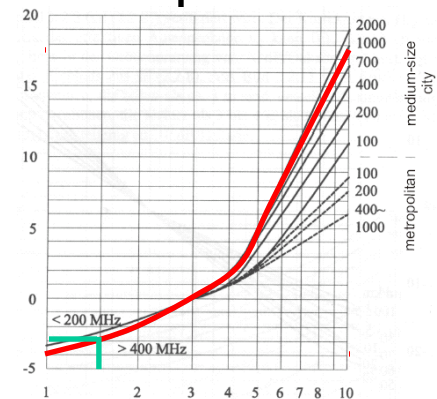


Fig. 7.14

Okumura's measurements

Example



Example

Propagation at 900 MHz in medium-size city with 40 m base station antenna height and 1.5 m mobile station antenna height.

Use Okumura's curves to calculate the propagation loss at a distance of 30 km between base station and mobile station.

Okumura's measurements

1. Calculate free-space loss



Example

Attenuation between two isotropic antennas in free space is (free-space loss):

$$L_{\text{free|dB}}(d) = 20 \log \left(\frac{4 \pi d}{\lambda} \right)$$

The obtained value does not depend on antenna heights.

900 MHz and
30 km distance

=> **121 dB**

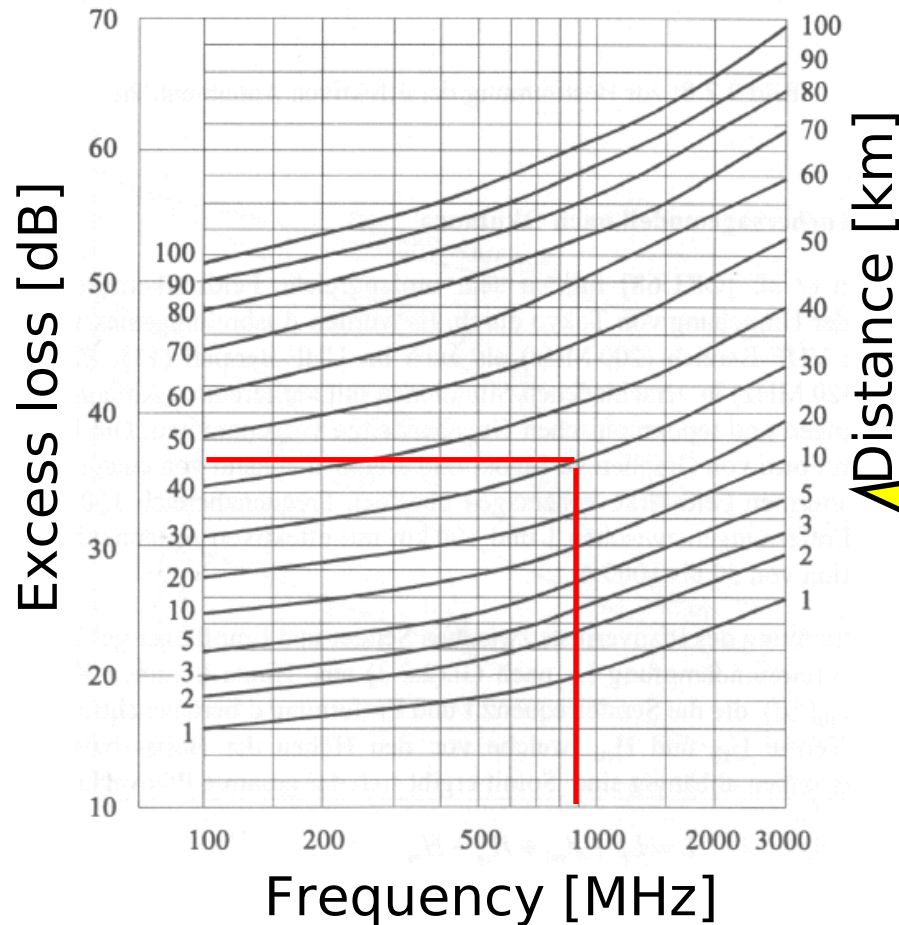
Okumura's measurements

2. Apply correction for excess loss



Example

FIGURE 7.12



These curves
are only for
 $h_b = 200$ m and
 $h_m = 3$ m

900 MHz and
30 km distance

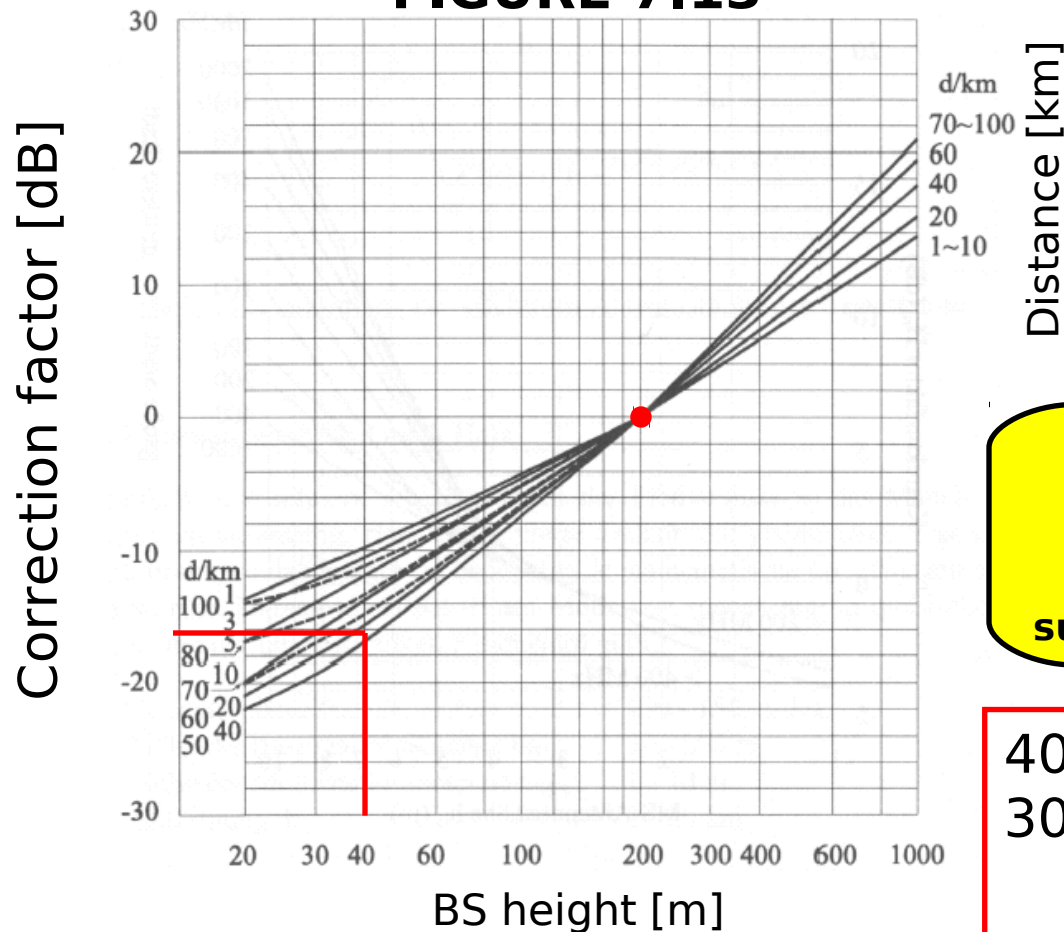
=> **36.5 dB**

Okumura's measurements

3. Apply correction of BS height



FIGURE 7.13



Example

Note: Lower base station means INCREASING attenuation => **subtract** this number.

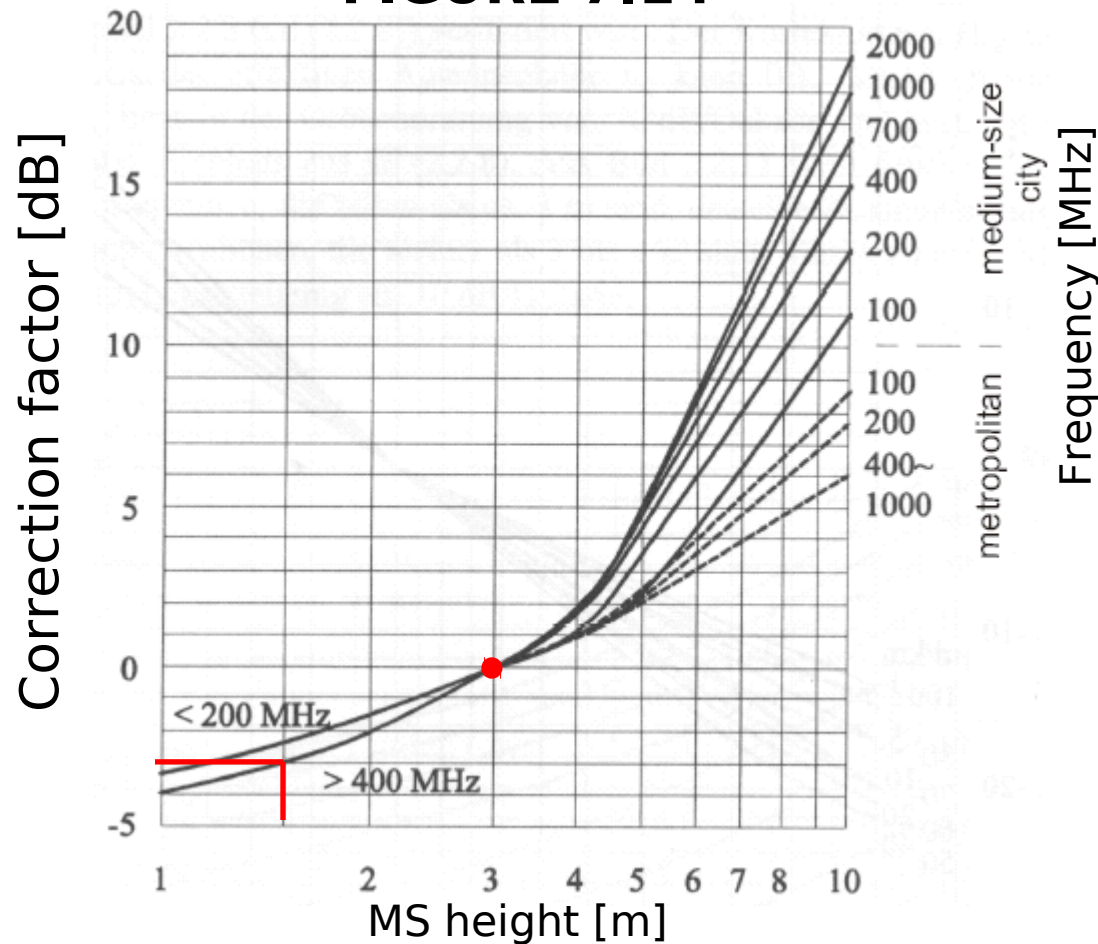
40 m BS and
30 km distance
=> **-16 dB**

Okumura's measurements

4. Apply correction of MS height



FIGURE 7.14



Example

Note: Lower mobile station means INCREASING attenuation => **subtract** this number.

1.5 m MS and 900 MHz in medium-size city => **-3 dB**

Okumura's measurements

Summary of example



Example

Propagation loss (between isotropic antennas) using Okumura's measurements:

$$L_{Oku|dB} = 121 + 36.5 - (-16) - (-3) = 176.5 \text{ dB}$$

Calc. step: 1 2 3 4

The Okumura-Hata model Background



In 1980 Hata published a parameterized model, based on Okumura's measurements.

The parameterized model has a *smaller range of validity* than the measurements by Okumura:

Frequency	150 – 1500 MHz
Distance	1 – 20 km
Mobile station height	1 – 10 m
Base station height	30 – 200 m

The Okumura-Hata model

How to calculate prop. loss



$$L_{O-H} = A + B \log(d_{|km}) + C$$

h_b and h_m
in **meter**

$$A = 69.55 + 26.16 \log(f_{0|MHz}) - 13.82 \log(h_b) - a(h_m)$$

$$B = 44.9 - 6.55 \log(h_b)$$

	$a(h_m) =$	$C =$
Metropolitan areas	$8.29 (\log(1.54 h_m))^2 - 1.1$ for $f_0 \leq 200$ MHz $3.2 (\log(11.75 h_m))^2 - 4.97$ for $f_0 \geq 400$ MHz	0
Small/medium-size cities	$(1.1 \log(f_{0 MHz}) - 0.7) h_m -$ $(1.56 \log(f_{0 MHz}) - 0.8)$	0
Suburban environments		$-2 (\log(f_{0 MHz}/28))^2 - 5.4$
Rural areas		$-4.78 (\log(f_{0 MHz}))^2 + 18.33 \log(f_{0 MHz}) - 40.94$

COST 231-Walfish-Ikegami model Background



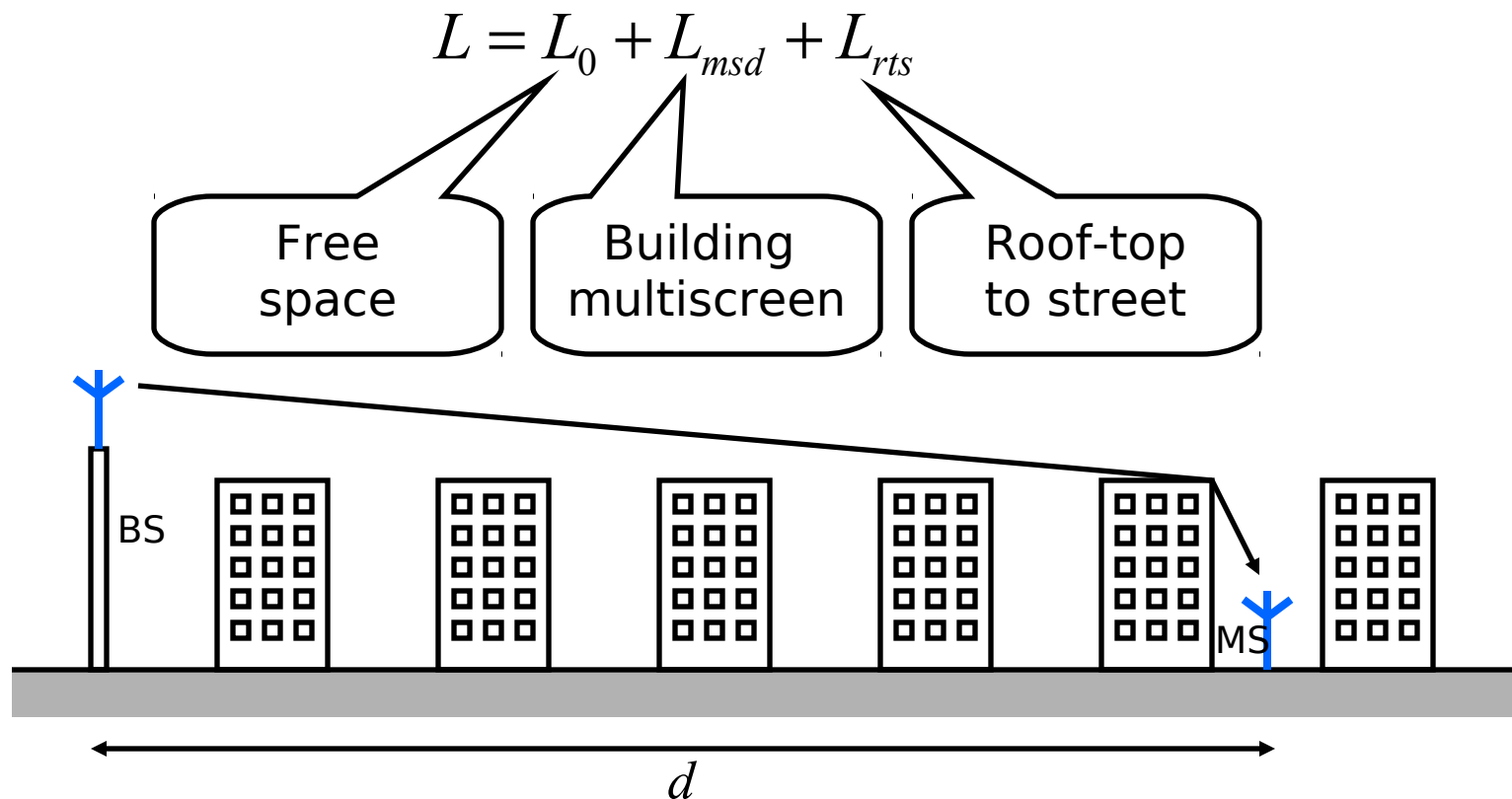
The Okumura-Hata model is not suitable for micro cells or small macro cells, due to its restrictions on distance ($d > 1$ km).

The COST 231-Walfish-Ikegami model covers much smaller distances and is better suited for calculations on small cells.

Frequency	800 – 2000 MHz
Distance	0.02 – 5 km
Mobile station height	1 – 3 m
Base station height	4 – 50 m

COST 231-Walfish-Ikegami model

How to calculate prop. loss



Details about calculations can be found in Appendix 7.B.



WIDEBAND MODELS

Wideband models

Review of properties



Let's assume the tapped delay-line model

$$h(t, \tau) = \sum_{i=1}^N \alpha_i(t) \exp(j \theta_i(t)) \delta(\tau - \tau_i)$$

The **power-delay profile** tells us how much energy the channel has at a certain delay τ (essentially the rms values of the $\alpha_i(t)$'s).

The **Doppler spectrum** tells us how fast the channel changes in time (essentially how fast the $\alpha_i(t)$'s and $\theta_i(t)$'s change). There can be one Doppler spectrum for each delay.

Wideband models

COST 207 model for GSM



The COST 207 model specifies:

FOUR power-delay profiles for different environments.

FOUR Doppler spectra used for different delays.

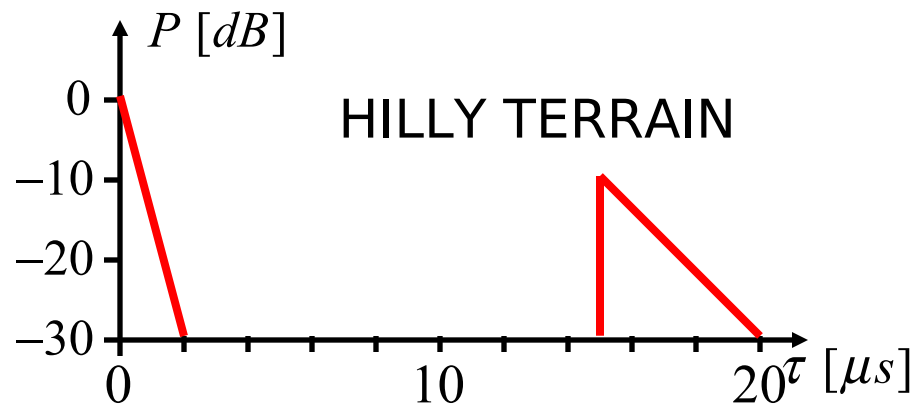
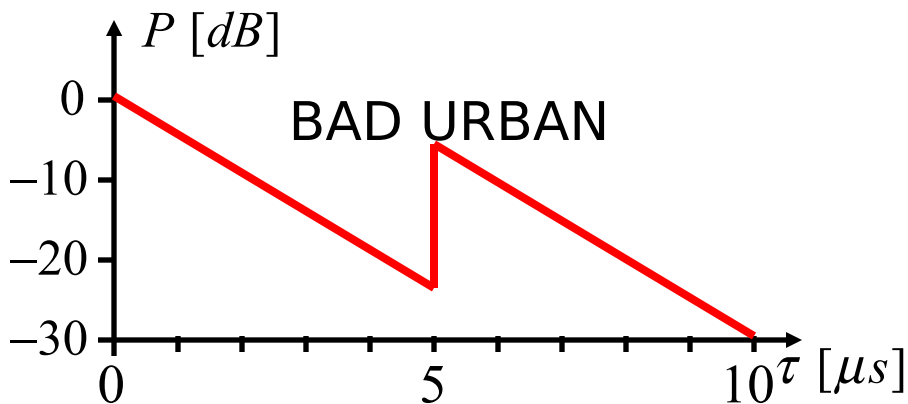
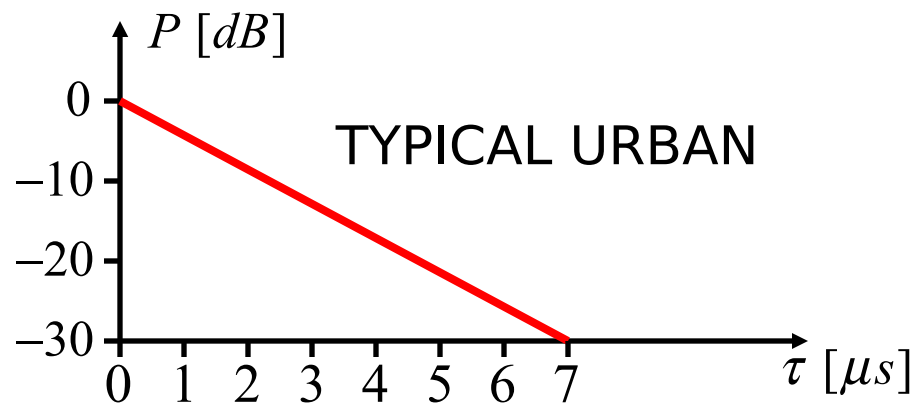
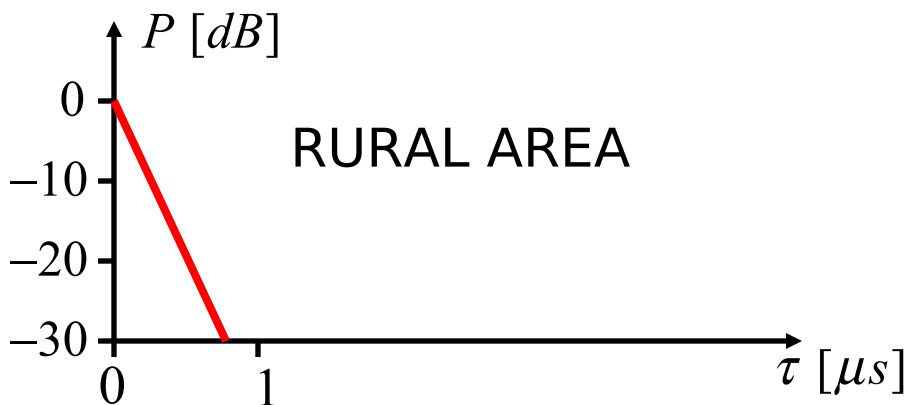
IT **DOES NOT** SPECIFY PROPAGATION LOSSES FOR THE DIFFERENT ENVIRONMENTS!

Wideband models

COST 207 model for GSM



Four specified power-delay profiles



Wideband models

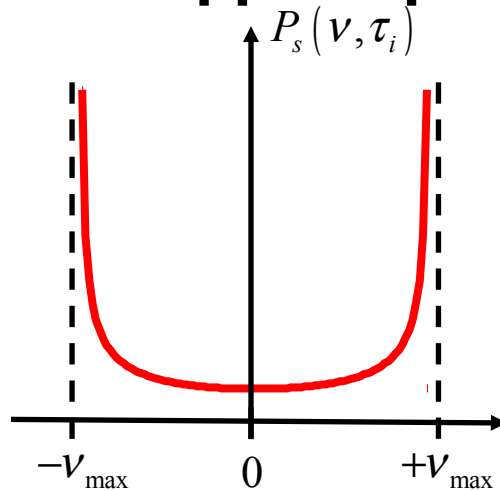
COST 207 model for GSM



Four specified Doppler spectra

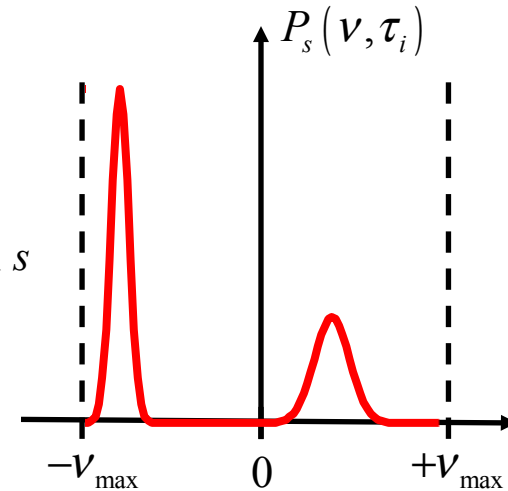
CLASS

$$\tau_i \leq 0.5 \mu s$$



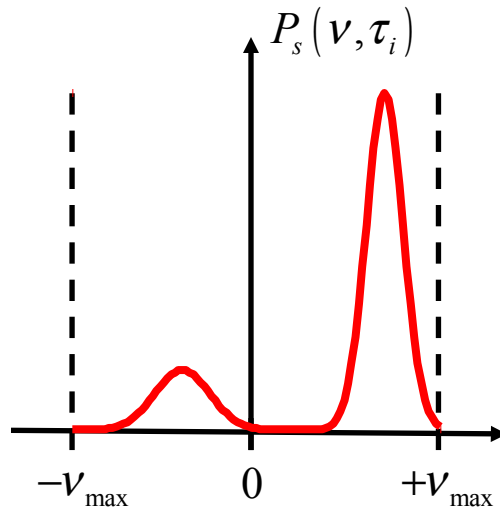
GAUS1

$$0.5 \mu s < \tau_i \leq 2 \mu s$$



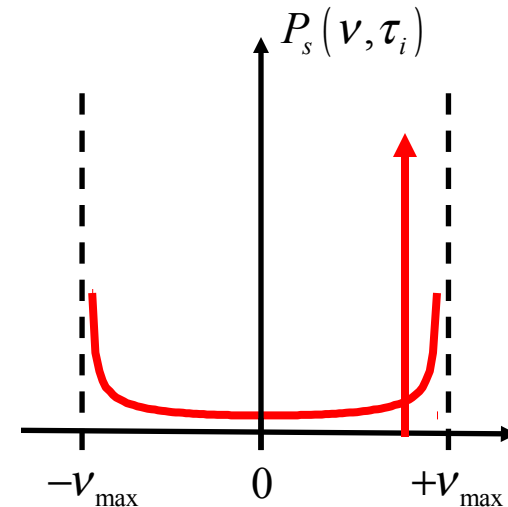
GAUS2

$$\tau_i > 2 \mu s$$



RICE

Shortest
path in
rural areas

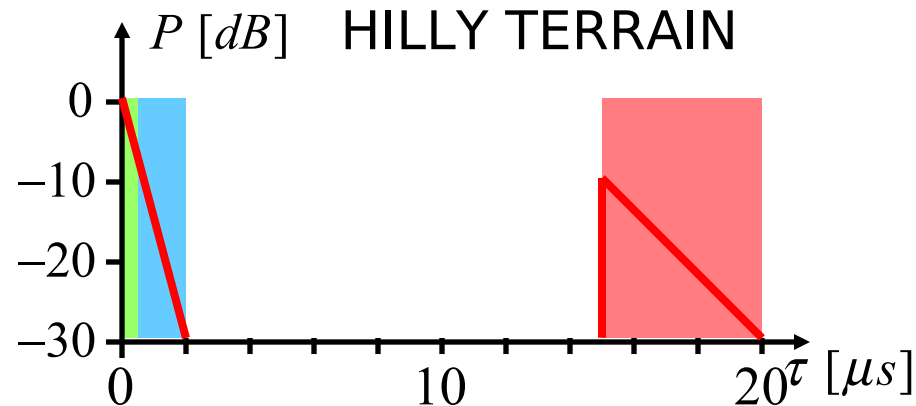
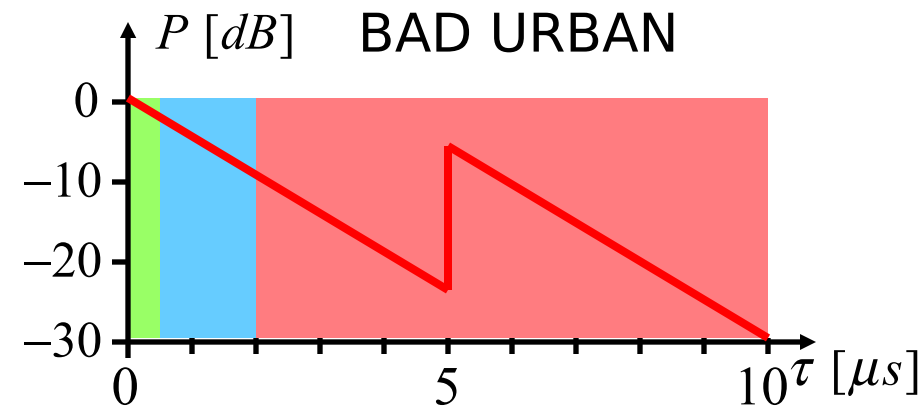
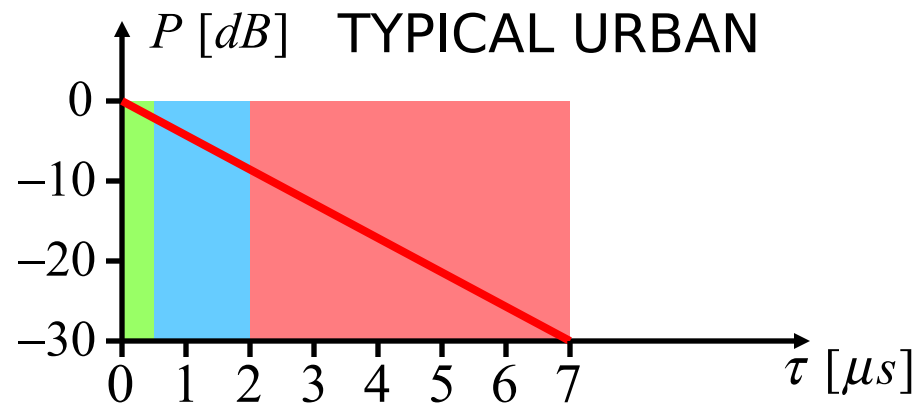
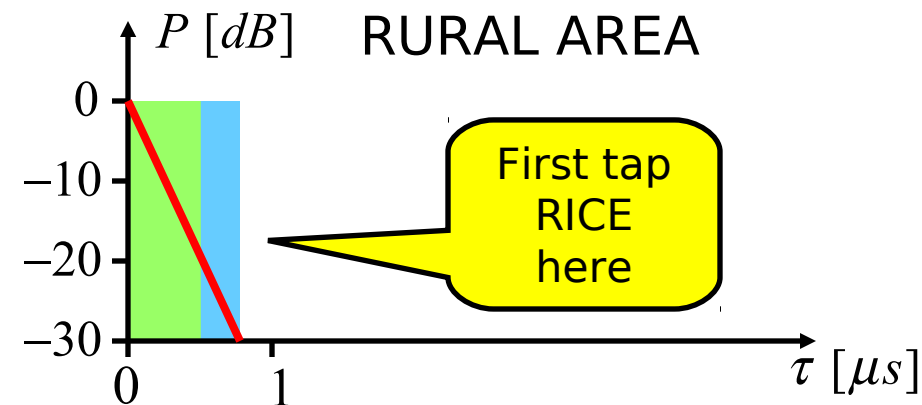


Wideband models

COST 207 model for GSM



Doppler spectra: CLASS GAUS1 GAUS2



Wideband models

COST 207 model for GSM



There are also suggested tapped delay-line implementations, with six Rayleigh-fading taps per channel. See Appendix 7.C (on-line).

QUICK QUIZ: The system bit-rate of GSM is 271 kbit/s.

How long is one bit in time?

How long are the different COST 207 channels, measured in bit-times?

Wideband models

ITU-R model for 3G



The ITU-R model specifies:

SIX different tapped delay-line channels for three different scenarios (indoor, pedestrian, vehicular).

TWO channels per scenario (one short and one long delay spread).

TWO different Doppler spectra (uniform & classical), depending on scenario.

THREE different models for propagation loss (one for each scenario).

The standard deviation of the log-normal shadow fading is specified for each scenario.

The autocorrelation of the log-normal shadow fading is specified for the vehicular scenario.

Wideband models

ITU-R model for 3G



Tap No.	delay/ns	power/dB	delay/ns	power/dB
INDOOR	CHANNEL A (50%)		CHANNEL B (45%)	
1	0	0	0	0
2	50	-3	100	-3.6
3	110	-10	200	-7.2
4	170	-18	300	-10.8
5	290	-26	500	-18.0
6	310	-32	700	-25.2
PEDESTRIAN	CHANNEL A (40%)		CHANNEL B (55%)	
1	0	0	0	0
2	110	-9.7	200	-0.9
3	190	-19.2	800	-4.9
4	410	-22.8	1200	-8.0
5			2300	-7.8
6			3700	-23.9
VEHICULAR	CHANNEL A (40%)		CHANNEL B (55%)	
1	0	0	0	-2.5
2	310	-1	300	0
3	710	-9	8900	-12.8
4	1090	-10	12900	-10.0
5	1730	-15	17100	-25.2
6	2510	-20	20000	-16.0



ANTENNAS

Antennas Efficiency



The antenna efficiency measures “how efficiently” an antenna converts the input power into radiation. This translates directly into power consumption and battery life.

Antenna efficiency of mobiles has **decreased** mainly due to cosmetic restrictions.

What cosmetic restrictions?

Antennas Bandwidth



We can say that the bandwidth of an antenna is the width of the frequency range over which it fulfills some specification.

Most cellular systems have a bandwidth requirement in the range of 10% of the carrier frequency.

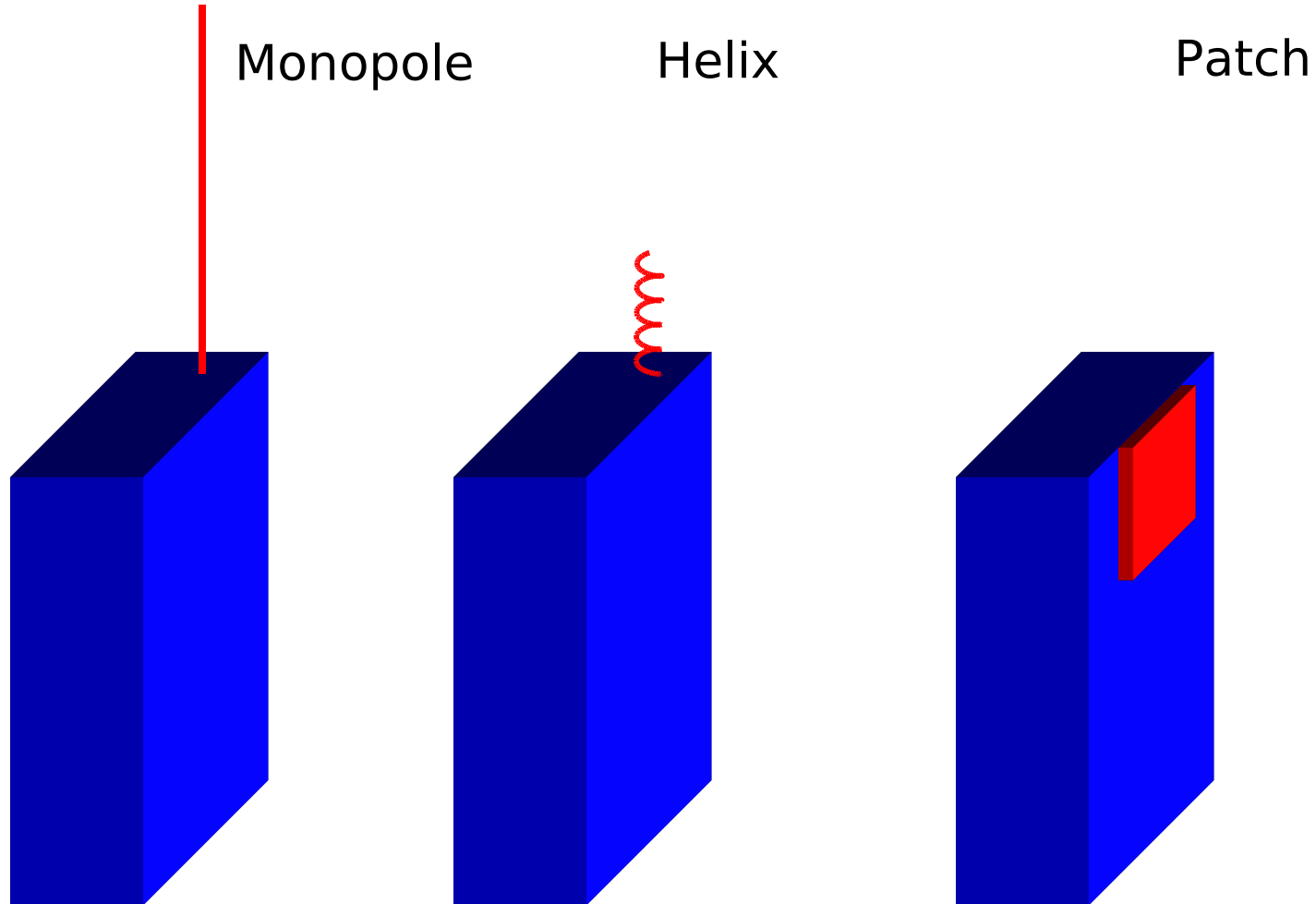
Example: 900 MHz GSM needs an antenna that can transmit/receive well in a total bandwidth of about 100 MHz.

It is difficult to make small and efficient broadband antennas!

What happens when we have dual- (900/1800) or triple-band (900/1800/1900) GSM phones ... or phones with 3G and Bluetooth (2.4 GHz) as well?

Antennas

Mobile station antennas

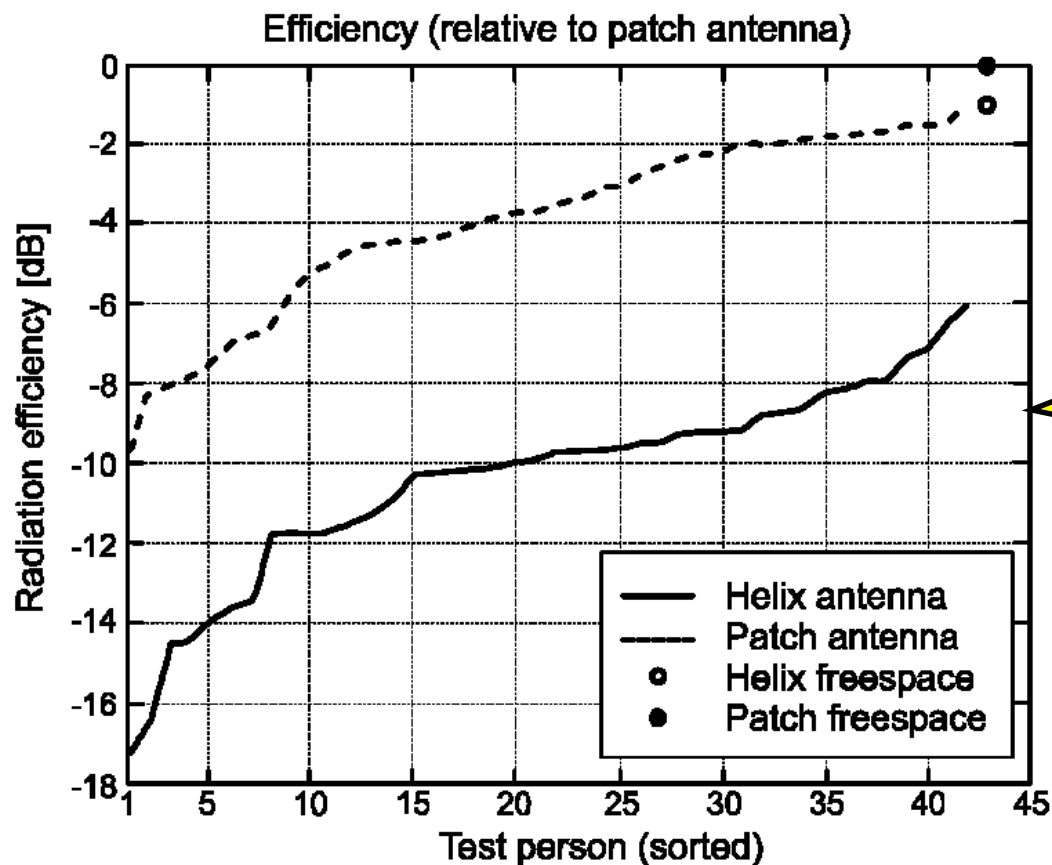


Antennas

Mobile station antennas



The efficiency depends on many parameters, but a very important one is its environment. Below you can see differences in antenna efficiency for 42 test persons holding the mobile.



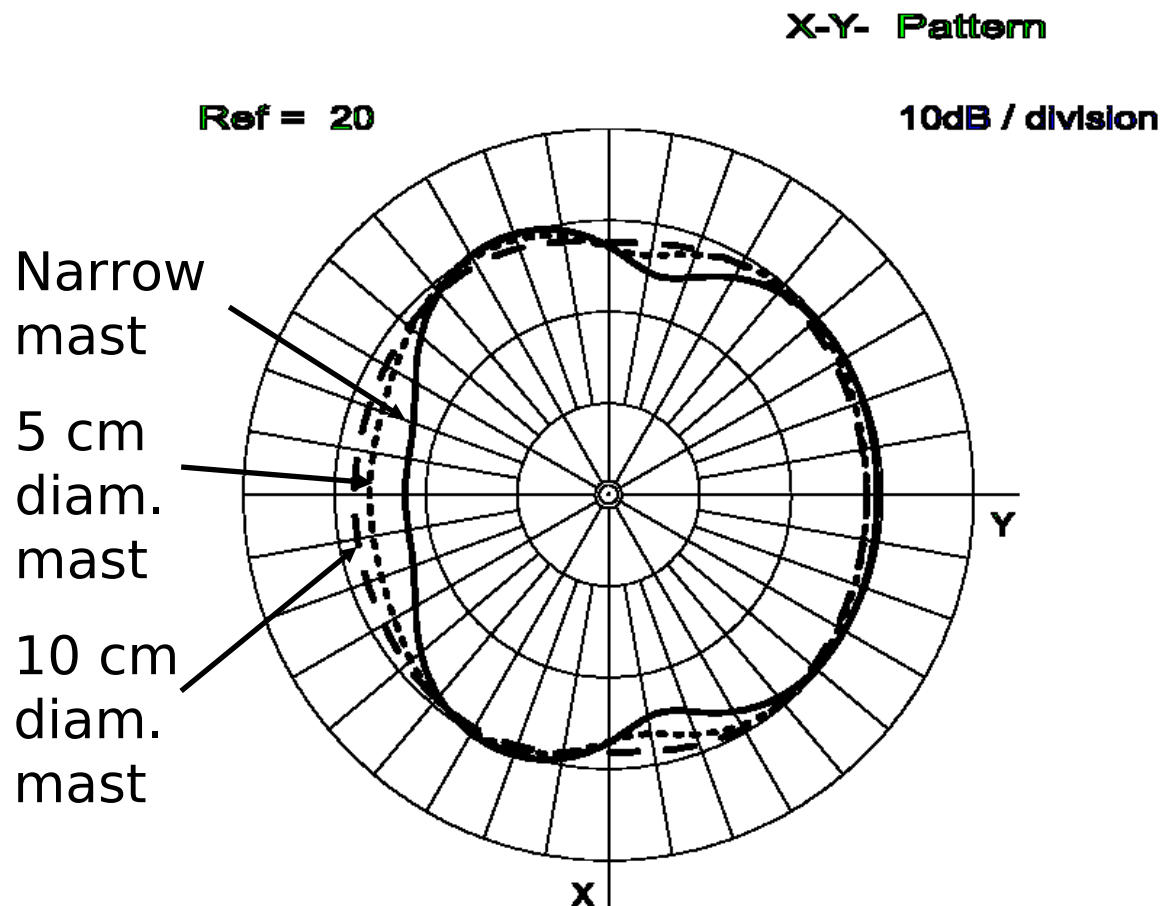
Up to around 10 dB difference, depending on person.

Antennas

Base station antennas



Base station antenna pattern affected by the mast (30 cm from antenna).

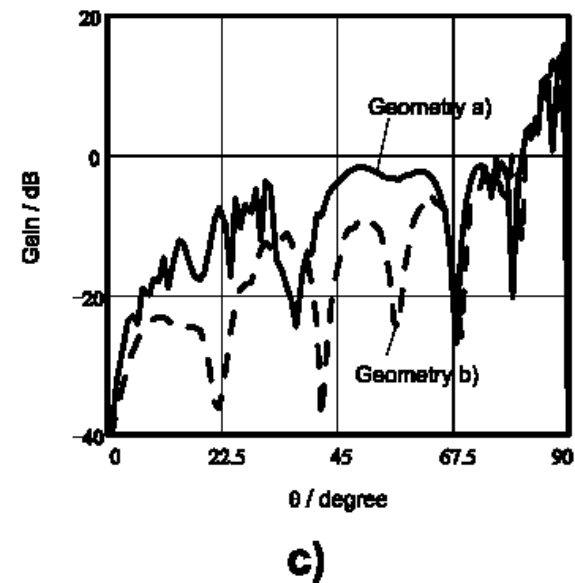
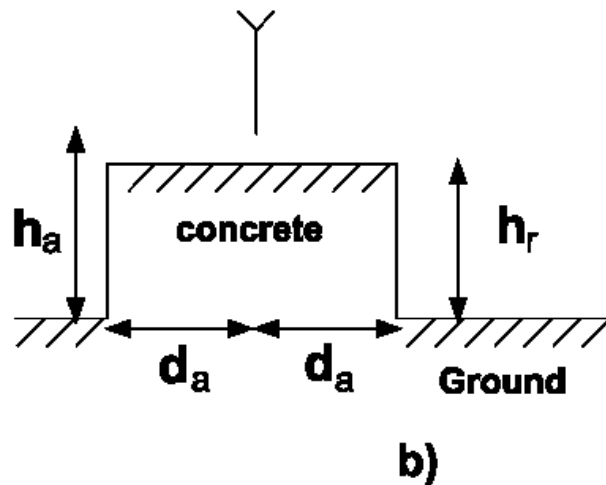
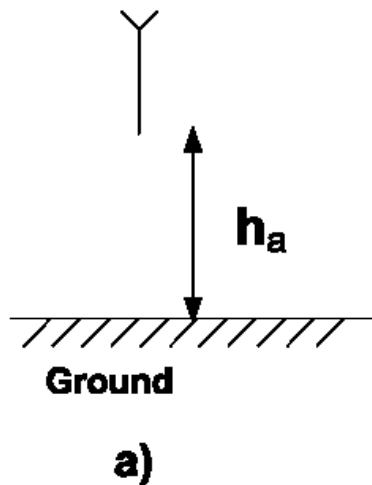


Antennas

Base station antennas



Base station antenna pattern affected by a concrete foundation.



Antennas

The dipole antenna



Short dipole
($L \ll \lambda/2$)

$$R_r \approx 20\pi^2 2 \left(\frac{L}{\lambda}\right)^2$$

$$G_a = 1,5 = 1,8 \text{ dB}$$

$$L_{\text{eff}} = \frac{L}{2}$$

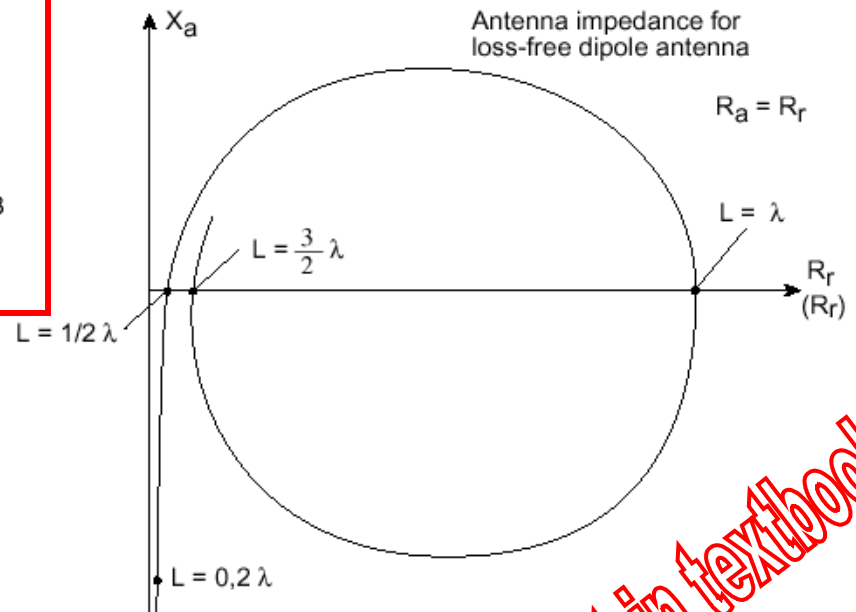
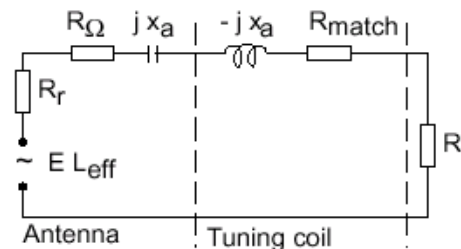
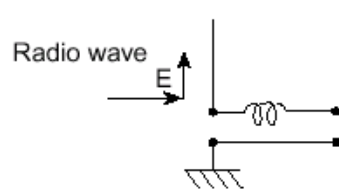
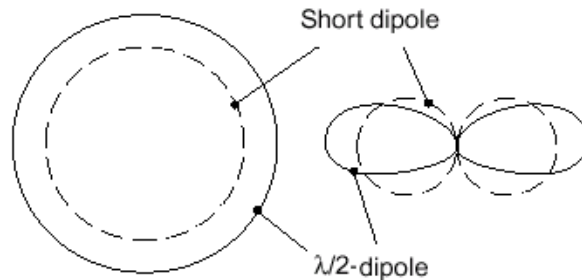
(R_r = Radiation resistance)

$\lambda/2$ -dipole
($L = \lambda/2$)

$$R_r = 73 \Omega$$

$$G_a = 1,64 = 2,15 \text{ dB}$$

$$L_{\text{eff}} = \frac{2}{\pi} L$$



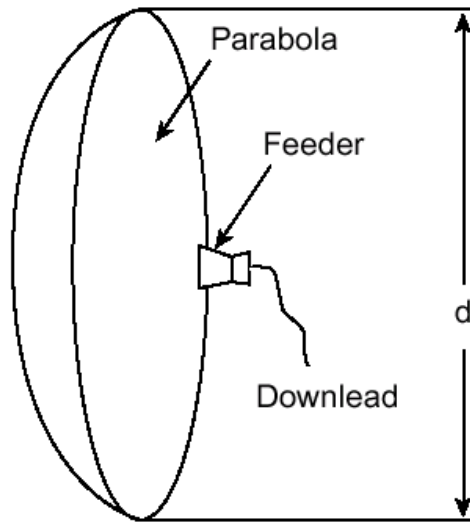
Not in textbook!

For a short dipole ($L/\lambda \ll 1/2$)
 R_r will be very low and $\frac{|X_a|}{R_r}$
 very high. Difficult to avoid
 ohmic losses (R_Ω)
 and losses in the tuning coil (R_{match})
 L_{eff} : Effective length of antenna
 Matching condition:
 $R_l = R_r + R_\Omega + R_{\text{match}}$

[Figure from Ericsson Radio School documentation]

Antennas

The parabolic antenna

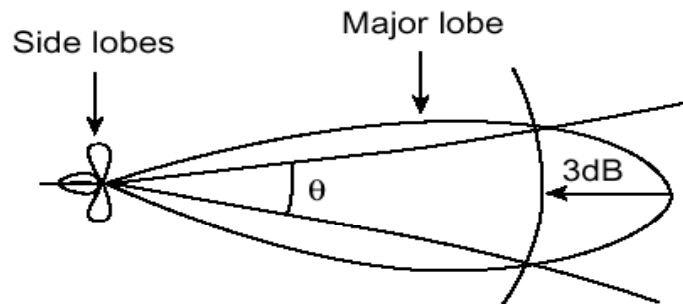


Opening area: $A = \frac{\pi d^2}{4}$

Effective area: $A_{\text{eff}} \approx 0.55 A$

Antenna gain: $G_a = \frac{4\pi}{\lambda^2} A_{\text{eff}} \approx 0.55 \frac{\pi^2 d^2}{\lambda^2}$

Not in textbook!



3dB beamwidth: $\theta \approx \frac{200}{\sqrt{G_a}} [\text{degrees}] (\theta < 25^\circ)$

[Figure from Ericsson Radio School documentation]



Summary

- Narrowband models: **Okumura's measurements, Okumura-Hata, COST 231-Ikegami-Walfish**. Mainly models for **propagation** loss. Fading has to be added.
- Wideband models: **COST 207 for GSM & ITU-R for 3G**. Mainly specification of **power-delay profile** and **doppler spectrum** (ITU-R also gives e.g. path loss).
- Antennas: **Efficiency** has decreased for mobile antennas. Antenna **environment** changes their properties. Some specific properties for **dipole** and **parabolic** antennas.