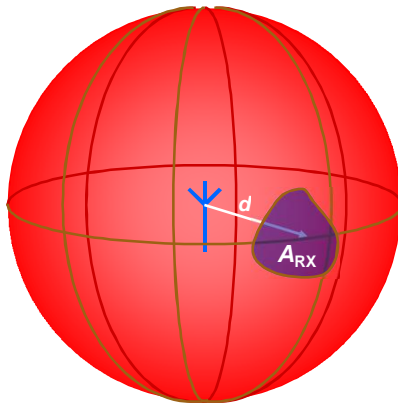


## Contents

---

- Free space loss
- Propagation mechanisms
  - Transmission
  - Reflection
  - Diffraction
  - Scattering
  - Waveguiding
- Examples from propagation scenarios

## Free-space loss



If we assume RX antenna to be isotropic:

$$P_{RX} = \frac{\rho}{4\pi d^2} P_{TX}$$

Attenuation between two isotropic antennas in free space is (free-space loss):

$$L_{free}(d) = \frac{4\pi d^2}{\rho}$$

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## Free-space loss

### Friis' law

Received power, with antenna gains  $G_{TX}$  and  $G_{RX}$ :

$$P_{RX}(d) = \frac{G_{RX} G_{TX}}{L_{free}(d)} P_{TX} = P_{TX} \frac{\rho}{4\pi d^2} G_{RX} G_{TX}$$



Valid in the far field only!

$$\begin{aligned} P_{RX|dB}(d) &= P_{TX|dB} + G_{TX|dB} - L_{free|dB}(d) + G_{RX|dB} \\ &= P_{TX|dB} + G_{TX|dB} - 10 \log_{10} \frac{4\pi d^2}{\rho} + G_{RX|dB} \end{aligned}$$

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## Free-space loss

### What is far field?

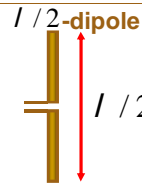
The free-space loss calculations are only valid in the far field of the antennas.

Far-field conditions are assumed "far beyond" the Rayleigh distance:

$$d_R = \frac{2L_a^2}{\lambda}$$

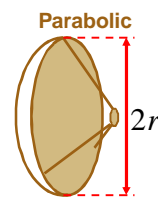
where  $L_a$  is the largest dimension of the antenna.

Another rule of thumb is:  
"At least 10 wavelengths"



$$L_a = l / 2$$

$$d_R = l / 2$$



$$L_a = 2r$$

$$d_R = \frac{8r^2}{\lambda}$$



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## Example

- Cellular phone, height 10 cm,  $f_c=900$  MHz  
Rayleigh distance  $d_R=2*0.1^2/0.333=6$  cm  
For this device the limit is the 10-lambda rule of thumb.
- Microwave link, antenna diameter 1.2 m,  $f_c=26$  GHz  
Rayleigh distance  $d_R=2*1.2^2/0.011=250$  m  
For this device the limit is the Rayleigh distance



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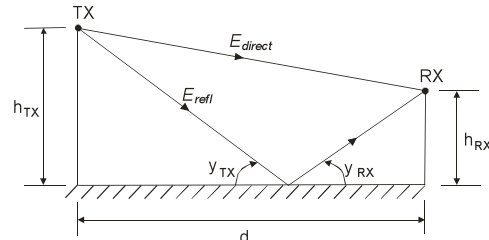
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## The $d^{-4}$ law (1)

For the following scenario



the power goes like

$$P_{RX}(d) \approx P_{TX} G_{TX} G_{RX} \left( \frac{h_{TX} h_{RX}}{d^2} \right)^2$$

for distances greater than

$$d_{\text{break}} \gtrsim 4h_{TX}h_{RX}/\lambda$$

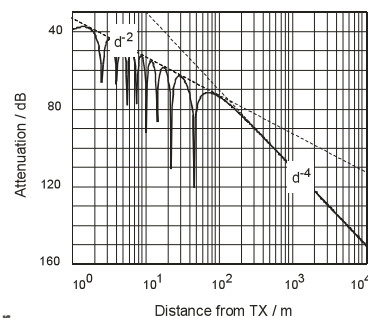
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## The $d^{-4}$ law (2)



• However .....

- $n=4$  is not a universal decay exponent
- Theoretical model is not fulfilled in practice
- Breakpoint is rarely where theoretically predicted
- Second breakpoint at the radio horizon

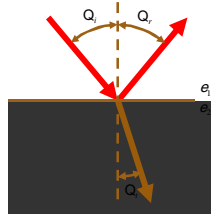
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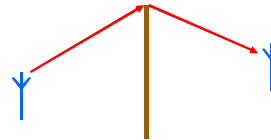


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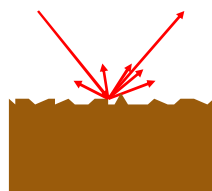
## Propagation mechanisms



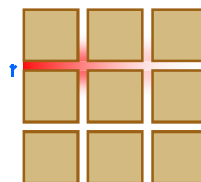
Reflection and transmission



Diffraction



Scattering



Waveguiding

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## Complex dielectric constant

$$\delta_i = \epsilon_i - j \frac{\sigma_{e,i}}{2\pi f_c}$$

↖ conductivity  
↙ dielectric constant, permittivity

Describes the dielectric material in one single parameter

Examples	Rel. permittivity	conductivity
Concrete	6	10 <sup>-2</sup>
Gypsum	6.5	10 <sup>-2</sup>
Wood	23	10 <sup>-11</sup>
Glass	5	10 <sup>-12</sup>
Air	1	

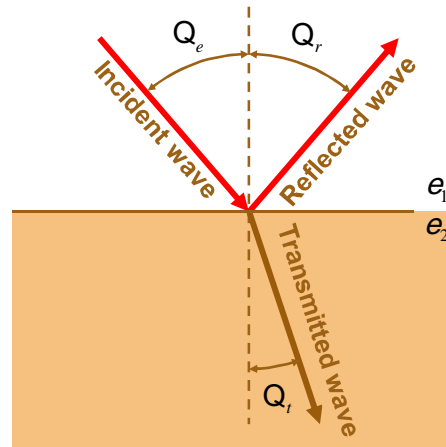
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## Reflection and transmission



reflected angle

$$\Theta_e = \Theta_r$$

transmitted angle

$$\frac{\sin \Theta_t}{\sin \Theta_e} = \frac{\sqrt{\epsilon_1}}{\sqrt{\epsilon_2}}$$

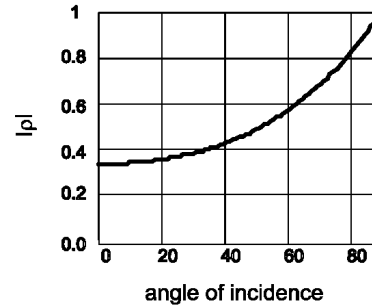
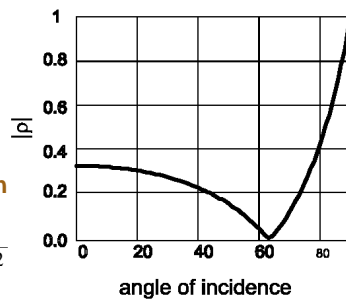
## TM and TE waves behave differently

Reflection coefficient

$$\rho_{TM} = -\frac{\sqrt{\delta_2} \cos \Theta_e - \sqrt{\delta_1} \cos(\Theta_t)}{\sqrt{\delta_2} \cos \Theta_e + \sqrt{\delta_1} \cos(\Theta_t)} \quad \rho_{TE} = \frac{\sqrt{\delta_1} \cos(\Theta_e) - \sqrt{\delta_2} \cos(\Theta_t)}{\sqrt{\delta_1} \cos(\Theta_e) + \sqrt{\delta_2} \cos(\Theta_t)}$$

TM-waves

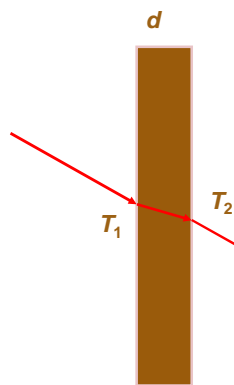
TE-waves



Transmission coefficient

$$T = \sqrt{1 - r^2}$$

## Transmission through walls – layered structures



**Total transmission coefficient**

$$T = \frac{T_1 T_2 e^{-j\alpha}}{1 + R_1 R_2 e^{-2j\alpha}}$$

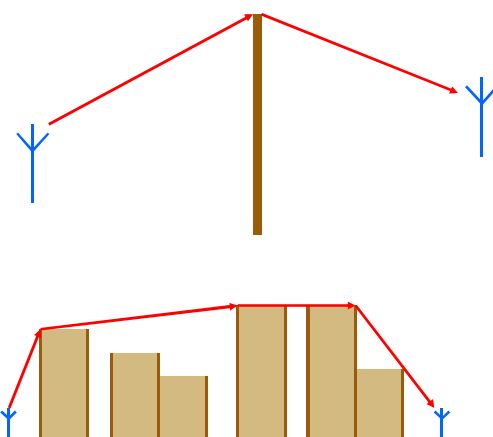
**total reflection coefficient**

$$\rho = \frac{\rho_1 + \rho_2 e^{-j2\alpha}}{1 + \rho_1 \rho_2 e^{-2j\alpha}}$$

**with the electrical length in the wall**

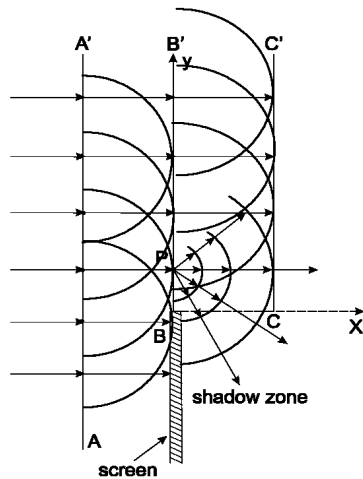
$$\alpha = \frac{2\pi}{\lambda} \sqrt{\epsilon_1} d_{\text{layer}} \cos(\Theta_t)$$

## Diffraction



- Single or multiple edges
- makes it possible to go behind corners
- less pronounced when the wavelength is small compared to objects

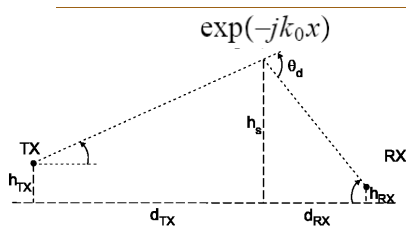
## Diffraction, Huygen's principle



Each point of a wavefront can be considered as a source of a spherical wave

➔ Bending around corners and edges

## Diffraction coefficient



The Fresnel integral is defined

$$F(v_F) = \int_0^{v_F} \exp(-j\pi \frac{t^2}{2}) dt.$$

with the Fresnel parameter

$$v_F = \alpha_k \sqrt{\frac{2d_1 d_2}{\lambda(d_1 + d_2)}}$$

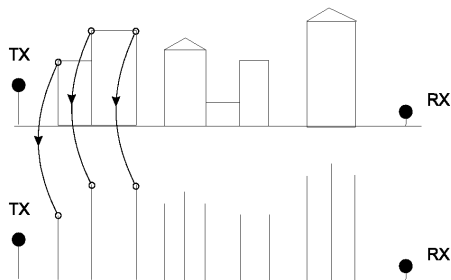
Total field

$$E_{total} = \exp(-jk_0 x) \left( \frac{1}{2} - \frac{\exp(-j\pi/4)}{\sqrt{2}} F(v_F) \right)$$

Fresnel integral

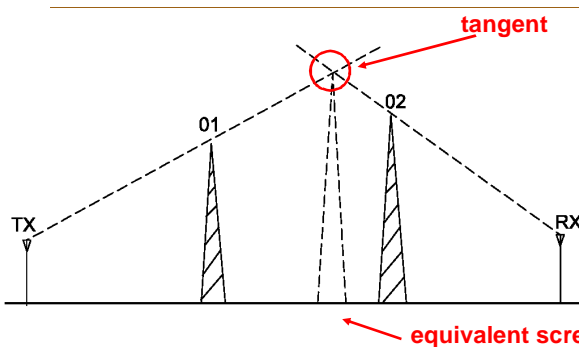


## Diffraction in real environments



For real environments we can represent buildings and objects as multiple screens

## Diffraction – Bullington’s method



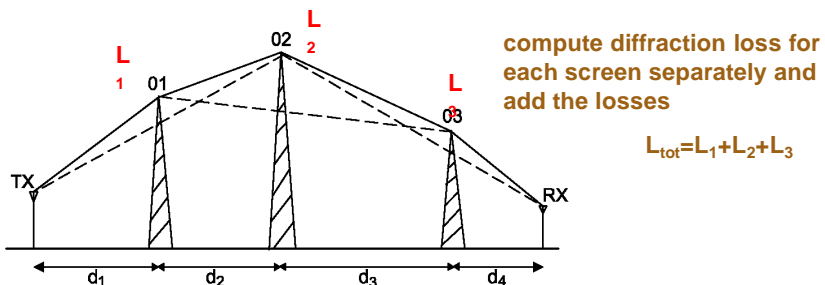
Replace all screens with one equivalent screen

Height determined by the steepest angle

Simple but a bit optimistic

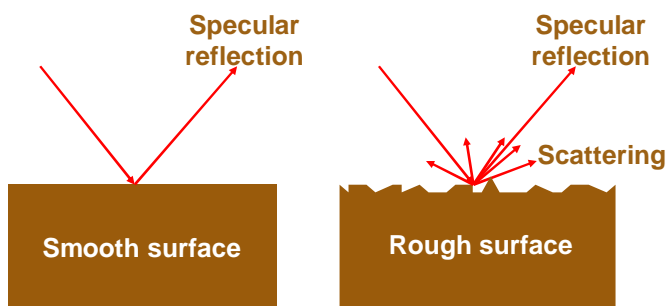
$$E_{\text{total}} = \exp(-jk_0x) \left( \frac{1}{2} - \frac{\exp(-j\pi/4)}{\sqrt{2}} F(v_F) \right) \quad v_F = \alpha_k \sqrt{\frac{2d_1d_2}{\lambda(d_1+d_2)}}$$

## Diffraction – Epstein-Petersen Method

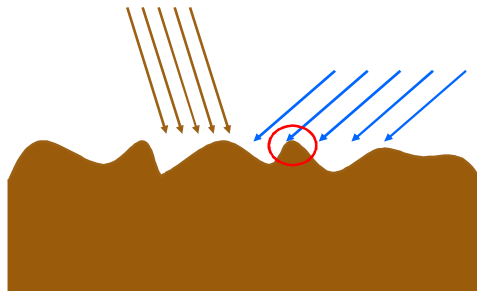


The same approach is used also for the ITU model, but with an empirical correction factor

## Scattering



### Kirchhoff theory – scattering by rough surfaces



calculate distribution of the surface amplitude

assume no “shadowing” from surface

calculate a new reflection coefficient

for Gaussian surface distribution **angle of incidence**

$$\rho_{\text{rough}} = \rho_{\text{smooth}} \exp\left[-2\left(k_0 \sigma_h \sin \psi\right)^2\right]$$

**standard deviation of height**

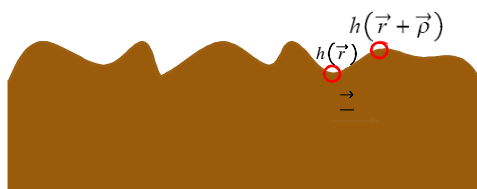


### Perturbation theory – scattering by rough surfaces

$$\sigma_h^2 W(\vec{\rho}) = E_{\vec{r}} \{ h(\vec{r}) h(\vec{r} + \vec{\rho}) \}$$

Include shadowing effects by the surface

includes spatial correlation of surface – how fast are the changes in height

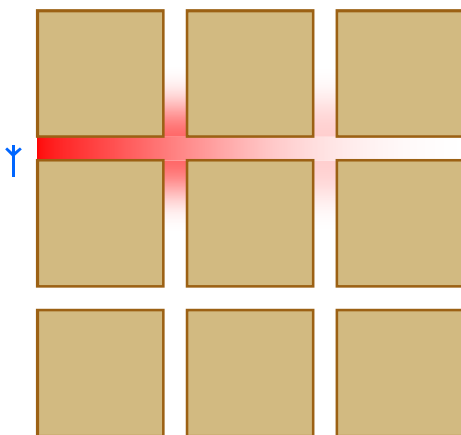


based on calculation of an “effective” dielectric constant

More accurate than Kirchhoff theory, especially for large angles of incidence and “rougher” surfaces



## Waveguiding

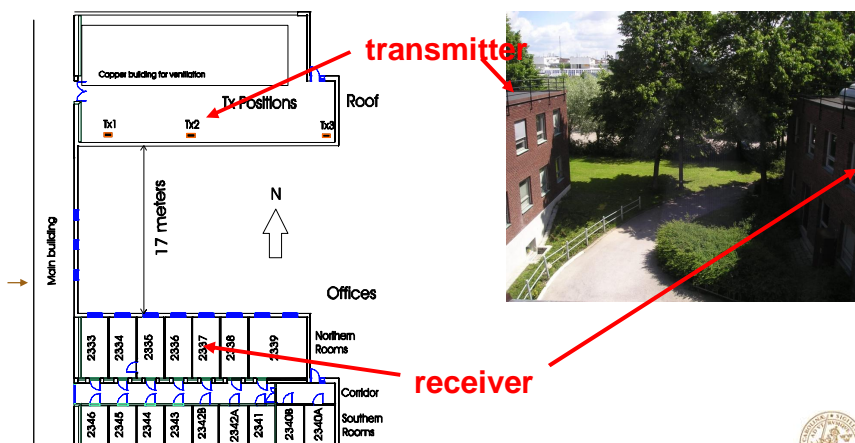


Waveguiding effects often result in lower propagation exponents

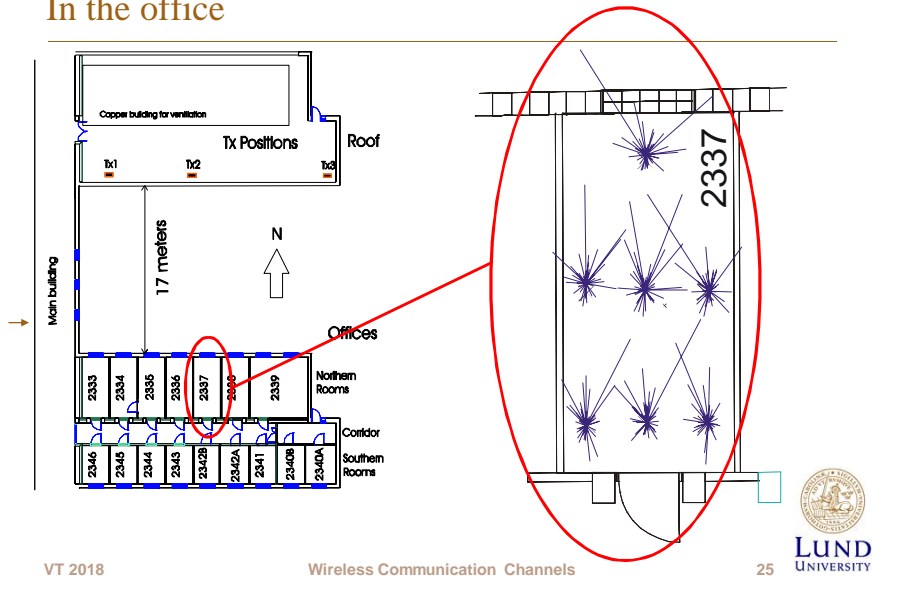
$$n = 1.5-5$$

This means lower path loss along certain street corridors

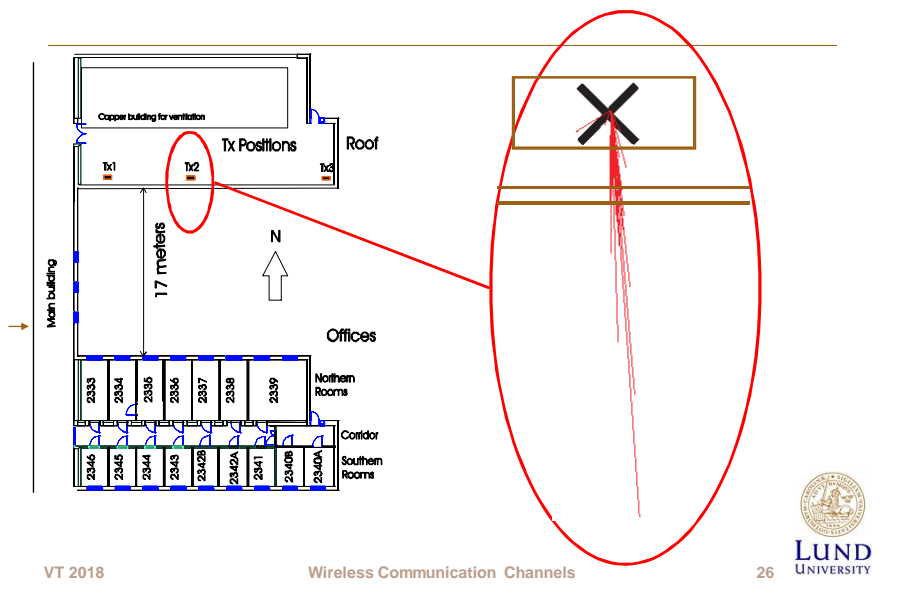
## How does the signal reach the receiver Outdoor-to-indoor



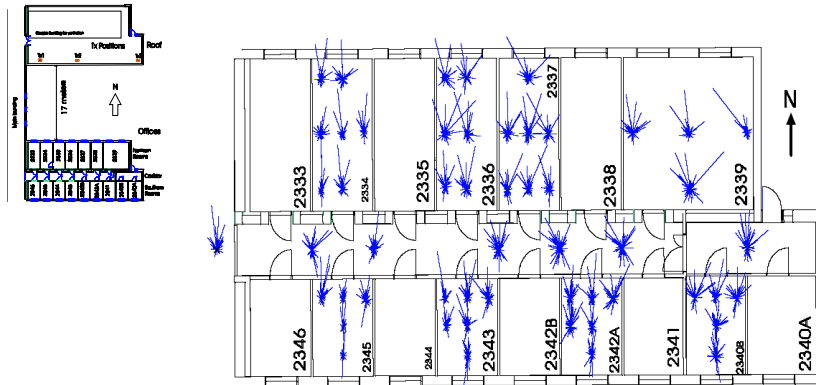
### How does the signal reach the receiver In the office



### How does the signal leave the transmitter at the roof



In all offices



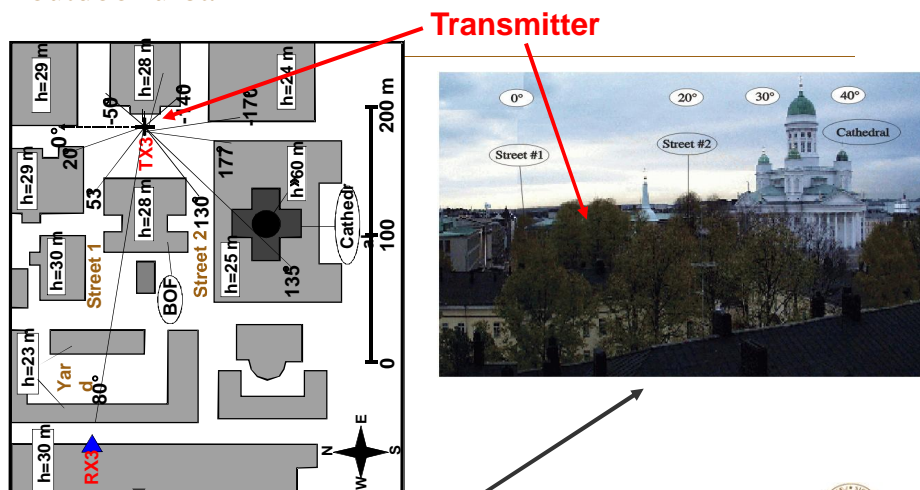
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How does the signal reach the receiver outdoor urban



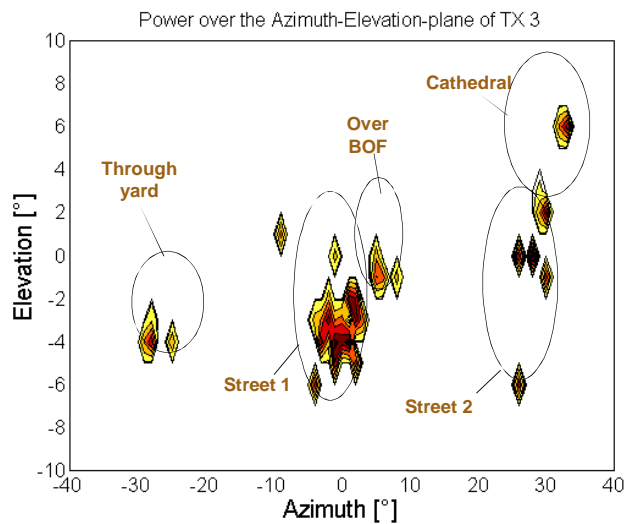
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Signal arrives from some specific areas



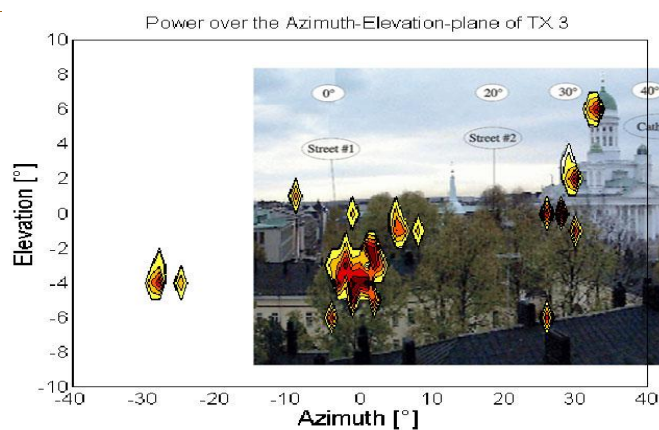
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Diffraction, reflection, scattering, transmission



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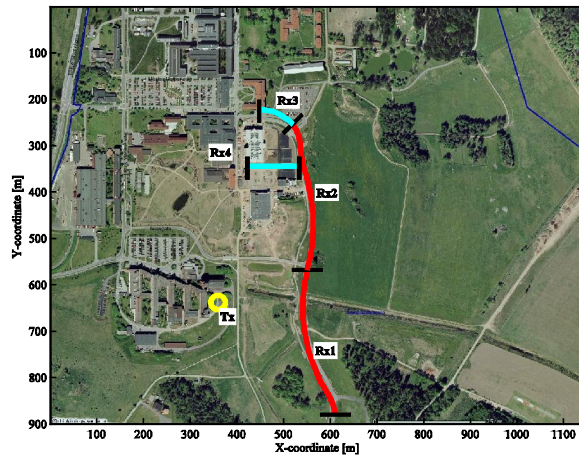
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## Outdoor 300 MHz peer-to-peer scenario

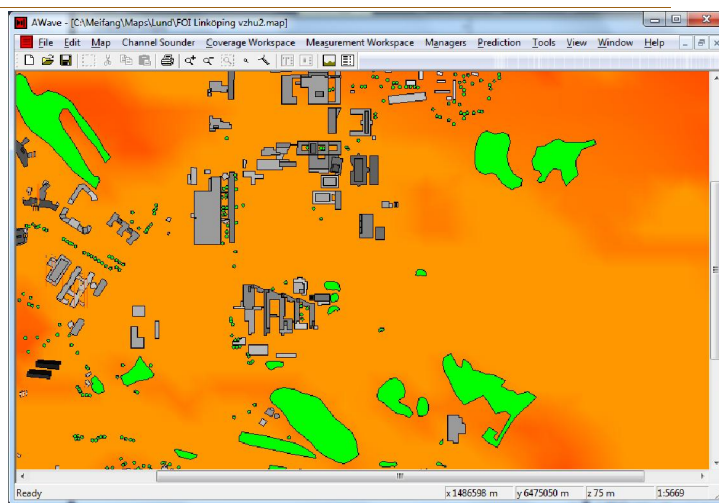
- Center frequency:
  - 285 MHz, 20MHz bandwidth
- Peer-to-peer measurement:
  - TX, 1.8m (BS)
  - RX, 2.1m (MS)
- Four routes:
  - 322, 320, 80, 110 m
  - semi-rural scenario
  - sub-urban scenario



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## Digital 3D map of the environment

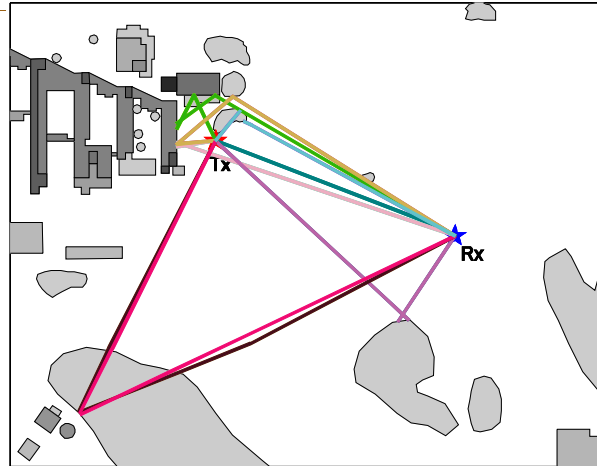


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## Visualized paths for a particular Tx/Rx position



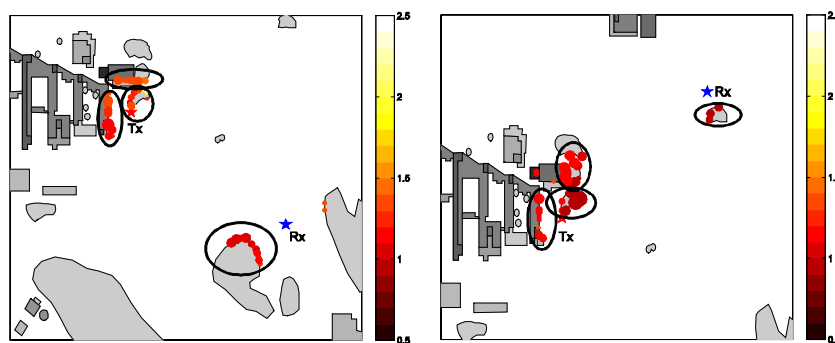
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## Interaction points, 20 strongest MPCs



LOS

NLOS

Radius of circles reflects power, color reflects delay

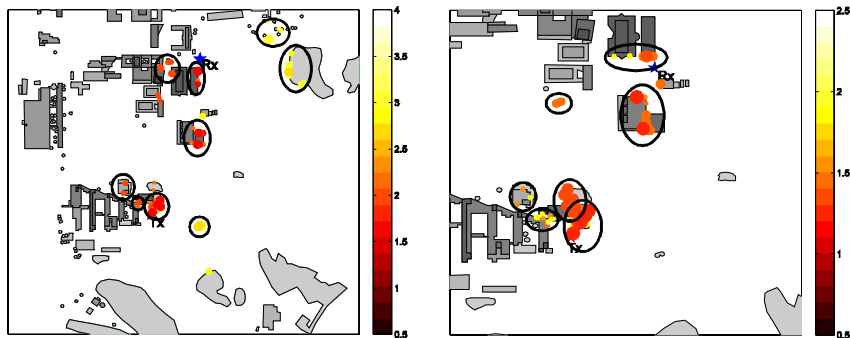
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## Interaction points, 20 strongest MPCs



NLOS

NLOS

Radius of circles reflects power, color reflects delay

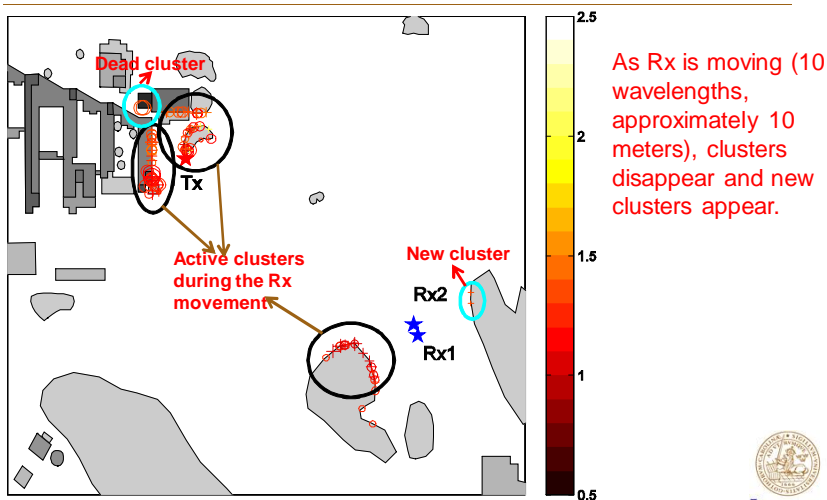
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## Multipath components tend to appear in clusters, moving Rx



As Rx is moving (10 wavelengths, approximately 10 meters), clusters disappear and new clusters appear.

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