

**Problem 1**

The channel probabilities are given as  $P(Y|X)$ . With  $P(X = 0) = p$  and  $P(X = 1) = 1 - p$ , the probabilities for  $Y$  are given by

$$P(Y = y) = \begin{cases} \alpha p + \alpha(1 - p) = \alpha, & y = 0 \\ \beta p + \beta(1 - p) = \beta, & y = \Delta \\ \gamma p + \gamma(1 - p) = \gamma, & y = 1 \end{cases} \quad (1.1)$$

Hence,

$$H(Y) = H(\alpha, \beta, \gamma)$$

$$H(Y|X) = \sum_x H(Y|X = x)P(X = x) = H(\alpha, \beta, \gamma)p + H(\alpha, \beta, \gamma)(1 - p) = H(\alpha, \beta, \gamma)$$

and the mutual information is  $I(X;Y) = H(Y) - H(Y|X) = 0$ . Since that imply  $I(X;Y) = H(X) - H(X|Y) = 0$  it means  $H(X) = H(X|Y)$ .

An alternativ solution is to observe that (1.1) means  $P(Y = y) = P(Y = y|X = x)$ , which gives

$$P(X = x)P(Y = y) = P(X = x)P(Y = y|X = x) = P(X = x, Y = y)$$

i.e.  $X$  and  $Y$  are independent and hence,  $H(X) = H(X|Y)$ .

**Problem 2**

Let  $X$  be the outcome of the die and  $Y$  the number of heads tossed with the coin. Then  $P(X = n) = 1/6$  for  $n = 1, \dots, 6$  and  $P(Y = k|X = n) = \binom{n}{k}2^{-n}$ ,  $k = 0, \dots, n$ . Thus the joint probability is  $P(X = n, Y = k) = \frac{\binom{n}{k}}{6 \cdot 2^n}$ , which is shown in the matrix

$$P = \begin{pmatrix} 1/12 & 1/12 & 0 & 0 & 0 & 0 & 0 \\ 1/24 & 1/12 & 1/24 & 0 & 0 & 0 & 0 \\ 1/48 & 1/16 & 1/16 & 1/48 & 0 & 0 & 0 \\ 1/96 & 1/24 & 1/16 & 1/24 & 1/96 & 0 & 0 \\ 1/192 & 5/192 & 5/96 & 5/96 & 5/192 & 1/192 & 0 \\ 1/384 & 1/64 & 5/128 & 5/96 & 5/128 & 1/64 & 1/384 \end{pmatrix}$$

The probability for the number of heads can be derived as

$$P(Y = k) = \sum_n P(X = n, Y = k) = (21/128 \quad 5/16 \quad 33/128 \quad 1/6 \quad 29/384 \quad 1/48 \quad 1/384)$$

for  $k = 0, \dots, 6$ . With these the entropies can be derived as  $H(X) = \log 6 \approx 2.5850$ ,  $H(Y) \approx 2.3074$  and  $H(X, Y) \approx 4.3972$ . Hence, the mutual information is  $I(X;Y) = H(X) + H(Y) - H(XY) \approx 0.4952$ .

**Problem 3**

The bandwidth for each channel is  $W_\Delta = 2$  MHz and the channel parameters expressed in linear scale becomes

$$|H_i|^2 = (0.1 \quad 0.05 \quad 0.025 \quad 0.0126 \quad 0.0063)$$

$$N_0 = 10^{-15} \text{ mW/Hz}$$

$$P = 10^{-6} \text{ mW}$$

- (a) To maximise the capacity the power is distributed according to the water-filling argument

$$P_i = \left( B - \frac{N_0 W_\Delta}{|H_i|^2} \right)^+ \quad \text{where} \quad \sum_i P_i = P$$

Summing the powers gives

$$\sum_i P_i = 5B - \sum_i \frac{N_0 W_\Delta}{|H_i|^2} = P$$

which gives

$$B = \frac{1}{5} \left( P + \frac{N_0 W_\Delta}{|H_i|^2} \right) = 3.23 \cdot 10^{-7}$$

and

$$P_i = (0.3031 \quad 0.2832 \quad 0.2435 \quad 0.1642 \quad 0.0061) \cdot 10^{-6}$$

The capacity is given by

$$\begin{aligned} C &= \sum_i W_\Delta \log \left( 1 + \frac{P_i |H_i|^2}{N_0 W_\Delta} \right) \\ &= 8.0276 + 6.0344 + 4.0413 + 2.0481 + 0.055 = 20.2 \text{ Mbps} \end{aligned}$$

- (b) For an M-PAM communication scheme we can estimate the total bit rate by

$$R = \sum_i R_i = \sum_i W_\Delta \log \left( 1 + \frac{P_i |H_i|^2}{\Gamma N_0 W_\Delta} \right) \quad \text{where} \quad \sum_i P_i = P$$

and  $\Gamma = 9 \text{ dB} = 10^{0.9} = 7.94$ . To maximise the bit rate the following maximisation function is used

$$J = \sum_i W_\Delta \log \left( 1 + \frac{P_i |H_i|^2}{\Gamma N_0 W_\Delta} \right) + \lambda \left( \sum_i P_i - P \right)$$

By setting the derivative equal to zero, the water-filling function can be derived as

$$P_i = \left( B - \frac{\Gamma N_0 W_\Delta}{|H_i|^2} \right)^+$$

After succesiv cancelations of sub-channels with negative power, i.e. sub-channel 4 and 5, the optimal power distribution is given by

$$P_i = (0.5439 \quad 0.3858 \quad 0.0703 \quad 0 \quad 0)$$

and the corresponding bit rate

$$R = \sum_i W_\Delta \log \left( 1 + \frac{P_i |H_i|^2}{\Gamma N_0 W_\Delta} \right) = 4.2905 + 2.2973 + 0.3042 = 6.89 \text{ Mbps}$$

**Problem 4**

The transition probability matrix is

$$P = \begin{pmatrix} 1/3 & 2/3 & 0 \\ 1/3 & 1/3 & 1/3 \\ 0 & 2/3 & 1/3 \end{pmatrix}$$

- (a) Through the equation system  $\pi P = \pi$  and  $\sum_i \pi_i = 1$  the stationary distribution is given by  $\pi = (1/4 \ 1/2 \ 1/4)$ . The entropy rate is given by

$$H_\infty(X) = \pi_1 h\left(\frac{1}{3}\right) + \pi_2 \log(3) + \pi_3 h\left(\frac{1}{3}\right) \approx 1.2516 \text{ bit}$$

- (b) The sequence  $\mathbf{Y}$  is given by taking two steps at a time in the graph. That corresponds to the transition probability matrix

$$P_b = P^2 = \begin{pmatrix} 1/3 & 4/9 & 2/9 \\ 2/9 & 5/9 & 2/9 \\ 2/9 & 4/9 & 1/3 \end{pmatrix}$$

Deriving the stationary distribution, by observing that  $\pi P^2 = \pi P$ , will give the same equation as in a), i.e.  $\pi = (1/4 \ 1/2 \ 1/4)$ . The entropy rate is given by

$$H_\infty(X) = \pi_1 H\left(\frac{1}{3}, \frac{4}{9}, \frac{2}{9}\right) + \pi_2 H\left(\frac{2}{9}, \frac{5}{9}, \frac{2}{9}\right) + \pi_3 H\left(\frac{2}{9}, \frac{4}{9}, \frac{1}{3}\right) \approx 1.4830 \text{ bit}$$

**Problem 5**

- (a) To form the new code replace  $x_k$  and  $x_l$  be replaced by their parent node  $z_m$  with probability  $q_m = p_k + p_l$ . The new code tree has the leaf probabilities  $\tilde{p}_j$ ,  $j = 1, \dots, n-1$ , which are equal to  $p_i$ ,  $i \neq k, l$  and  $q_m$ . Then

$$\begin{aligned} H(X) &= - \sum_{i=1}^n p_i \log p_i = - \sum_{i \neq k, l} p_i \log p_i - p_k \log p_k - p_l \log p_l \\ &= - \sum_{j=1}^{n-1} \tilde{p}_j \log \tilde{p}_j + q_m \log q_m - p_k \log p_k - p_l \log p_l \\ &= H(\tilde{X}) + (p_k + p_l) \left( \frac{p_k + p_l}{p_k + p_l} \log(p_k + p_l) - \frac{p_k}{p_k + p_l} \log p_k - \frac{p_l}{p_k + p_l} \log p_l \right) \\ &= H(\tilde{X}) + (p_k + p_l) \left( - \frac{p_k}{p_k + p_l} \log \frac{p_k}{p_k + p_l} - \frac{p_l}{p_k + p_l} \log \frac{p_l}{p_k + p_l} \right) \\ &= H(\tilde{X}) + (p_k + p_l) h\left(\frac{p_k}{p_k + p_l}\right) \end{aligned}$$

and, hence  $\alpha = \frac{p_k}{p_k + p_l}$ .

- (b) From the structure of the formula in a), assume that  $H(X) = \sum_{i=1}^{n-1} q_i h(\alpha_i)$ , where  $\alpha_i$  is the probability for choosing branch 0 from the inner node  $z_i$ . That is, If  $z_k$  and  $z_l$  are the children nodes of  $z_i$ , then  $\alpha_i = \frac{q_k}{q_k + q_l}$ .

- For  $n = 2$  there is one inner node, the root with  $q_1 = 1$  and two leaves with probabilities  $p_1$  and  $p_2$ . Then  $\alpha_1 = p_k$  and, hence,  $H(X) = h(p_1) = q_1 h(\alpha_1)$ . That means our assumption is true for this case.

- For  $n > 2$ , assume the formula is true for a tree with  $n - 1$  leaves. Then construct a code according to a),

$$\sum_{i=1}^{n-1} q_i h(\alpha_i) = \sum_{i \neq m} q_i h(\alpha_i) + q_m h(\alpha_m) = H(\tilde{X})(p_k + p_l) h\left(\frac{p_k}{p_k + p_l}\right) = H(X)$$

- (c) From the path length lemma and the fact that the entropy lower bounds the codeword length, we get

$$L = \sum_{i=1}^{n-1} q_i \geq H(X) = \sum_{i=1}^{n-1} q_i h(\alpha_i)$$

with equality if and only if  $h(\alpha_i) = 1$  for all  $i$ . This is equivalent to saying that  $\alpha_i = \frac{1}{2}$ , for all  $i$ . Since the leaf node  $x_i$  uniquely determines the path from the root to the leaf,  $z_{i_0}, z_{i_1}, \dots, z_{i_{\ell-1}}$ , where  $z_{i_0}$  is the root node, we get

$$\begin{aligned} p(x_i) &= P(z_{i_0}, \dots, z_{i_{\ell-1}}, x_i) \\ &= P(z_{i_0}) \prod_{j=1}^{\ell-1} P(z_{i_j} | z_{i_0}, \dots, z_{i_{j-1}}) P(x_i | z_{i_0}, \dots, z_{i_{\ell-1}}) \\ &= P(z_{i_0}) \prod_{j=1}^{\ell-1} P(z_{i_j} | z_{i_{j-1}}) P(x_i | z_{i_{\ell-1}}) \\ &= 1 \prod_{j=1}^{\ell} \alpha_{i_j} = 1 \prod_{j=1}^{\ell} \frac{1}{2} = \frac{1}{2^\ell} \end{aligned}$$

That is, to get  $L = H(X)$  it is required that the leaf probabilities are powers of 2. (This fact was also seen from the optimal codeword length, where  $\log p_i$  must be an integer to fulfil the bound.)